



STD95NH02L

N-CHANNEL 24V - 0.0039Ω - 80A DPAK ULTRA LOW GATE CHARGE STripFET™ MOSFET

Table 1: General Features

TYPE	V _{DSS}	R _{DS(on)}	I _D
STD95NH02L	24 V	< 0.005Ω	80(*) A

- TYPICAL R_{DS(on)} = 0.0039Ω @ 10 V
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED

DESCRIPTION

The **STD95NH02L** is based on the latest generation of ST's proprietary STripFET™ technology. An innovative layout enables the device to also exhibit extremely low gate charge for the most demanding requirements in high-frequency DC-DC converters. It's therefore ideal for high-density converters in Telecom and Computer applications.

APPLICATIONS

- SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTERS

Figure 1: Package



Figure 2: Internal Schematic Diagram

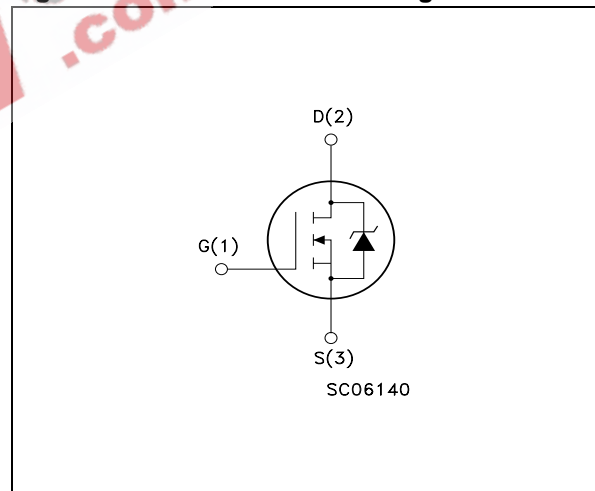


Table 2: Order Codes

PART NUMBER	MARKING	PACKAGE	PACKAGING
STD95NH02LT4	D95NH02L	DPAK	TAPE & REEL

STD95NH02L

Table 3: Absolute Maximum ratings

Symbol	Parameter	Value	Unit
$V_{spike(1)}$	Drain-source Voltage Rating	30	V
V_{DS}	Drain-source Voltage ($V_{GS} = 0$)	24	V
V_{DGR}	Drain-gate Voltage ($R_{GS} = 20\text{ k}\Omega$)	24	V
V_{GS}	Gate- source Voltage	± 20	V
$I_D (*)$	Drain Current (continuous) at $T_C = 25^\circ\text{C}$	80	A
I_D	Drain Current (continuous) at $T_C = 100^\circ\text{C}$	68	A
$I_{DM} (2)$	Drain Current (pulsed)	320	A
P_{TOT}	Total Dissipation at $T_C = 25^\circ\text{C}$	100	W
	Derating Factor	0.67	W/ $^\circ\text{C}$
$E_{AS} (3)$	Single Pulse Avalanche Energy	600	mJ
T_{stg}	Storage Temperature	-55 to 175	$^\circ\text{C}$
T_j	Max. Operating Junction Temperature		

(1) Guaranteed when external $R_g = 4.7\ \Omega$ and $t_f < t_r$ max.

(2) Pulse width limited by safe operating area.

(3) Starting $T_j = 25^\circ\text{C}$, $I_D = 40\text{A}$, $V_{DD} = 22\text{V}$

(*) Value limited by wires

Table 4: Thermal Data

$R_{thj-case}$	Thermal Resistance Junction-case Max	1.5	$^\circ\text{C}/\text{W}$
$R_{thj-amb}$	Thermal Resistance Junction-ambient Max	100	$^\circ\text{C}/\text{W}$
T_l	Maximum Lead Temperature For Soldering Purpose	275	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS ($T_{CASE} = 25^\circ\text{C}$ UNLESS OTHERWISE SPECIFIED)

Table 5: On/Off

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$V_{(BR)DSS}$	Drain-source Breakdown Voltage	$I_D = 250\ \mu\text{A}$, $V_{GS} = 0$	24			V
I_{DSS}	Zero Gate Voltage Drain Current ($V_{GS} = 0$)	$V_{DS} = \text{Max Rating}$ $V_{DS} = \text{Max Rating}$, $T_C = 125^\circ\text{C}$			1 10	μA μA
I_{GSS}	Gate-body Leakage Current ($V_{DS} = 0$)	$V_{GS} = \pm 20\text{V}$			± 100	nA
$V_{GS(th)}$	Gate Threshold Voltage	$V_{DS} = V_{GS}$, $I_D = 250\ \mu\text{A}$	1			V
$R_{DS(on)}$	Static Drain-source On Resistance	$V_{GS} = 10\text{V}$, $I_D = 40\text{A}$ $V_{GS} = 5\text{V}$, $I_D = 40\text{A}$		0.0039 0.0055	0.005 0.009	Ω Ω

ELECTRICAL CHARACTERISTICS (CONTINUED)

Table 6: Dynamic

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
g_{fs} (4)	Forward Transconductance	$V_{DS} = 10\text{ V}$, $I_D = 10\text{ A}$		30		S
C_{iss} C_{oss} C_{rss}	Input Capacitance Output Capacitance Reverse Transfer Capacitance	$V_{DS} = 15\text{ V}$, $f = 1\text{ MHz}$, $V_{GS} = 0$		2070 990 90		pF pF pF
$t_{d(on)}$ t_r $t_{d(off)}$ t_f	Turn-on Delay Time Rise Time Turn-off Delay Time Fall Time	$V_{DD} = 12\text{ V}$, $I_D = 40\text{ A}$, $R_G = 4.7\ \Omega$, $V_{GS} = 10\text{ V}$ (see Figure 16)		20 110 47 20		ns ns ns ns
Q_g Q_{gs} Q_{gd}	Total Gate Charge Gate-Source Charge Gate-Drain Charge	$V_{DD} = 12\text{ V}$, $I_D = 80\text{ A}$, $V_{GS} = 5\text{ V}$ (see Figure 19)		17 7.6 6.8		nC nC nC
Q_{oss} (5)	Output Charge	$V_{DS} = 19\text{ V}$, $V_{GS} = 0\text{ V}$		22.6		nC
Q_{gls} (6)	Third-Quadrant Gate Charge	$V_{DS} < 0\text{ V}$, $V_{GS} = 5\text{ V}$		15		nC
R_G	Gate Input Resistance	$f = 1\text{ MHz}$ Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.8		Ω

Table 7: Source Drain Diode

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
I_{SD}	Source-drain Current				80	A
I_{SDM}	Source-drain Current (pulsed)				320	A
V_{SD} (4)	Forward On Voltage	$I_{SD} = 40\text{ A}$, $V_{GS} = 0$			1.3	V
t_{rr} Q_{rr} I_{RRM}	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 80\text{ A}$, $di/dt = 100\text{ A}/\mu\text{s}$, $V_{DD} = 20\text{ V}$, $T_j = 150^\circ\text{C}$ (see Figure 16)		42 50.4 2.4		ns nC A

(4). Pulsed: Pulse duration = 300 μs , duty cycle 1.5 %.

(5). $Q_{oss} = C_{oss} \cdot \Delta V_{in}$, $C_{oss} = C_{gd} + C_{ds}$. See Appendix A.

(6). Gate charge for Synchronous Operation.

Figure 3: Safe Operating Area

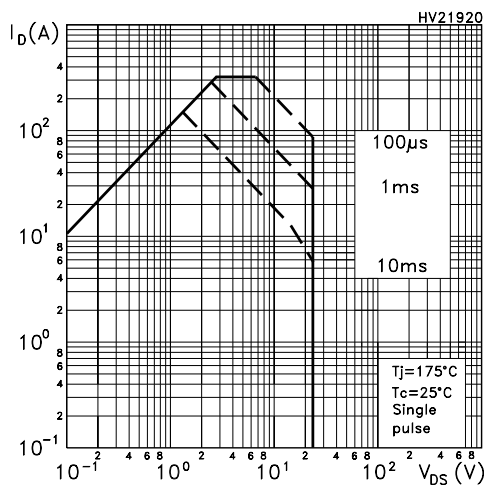


Figure 4: Output Characteristics

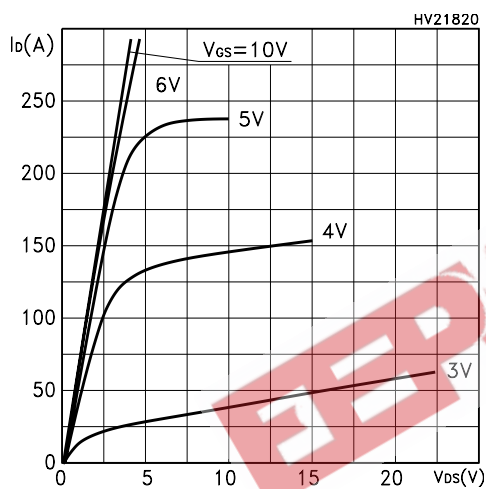


Figure 5: Transconductance

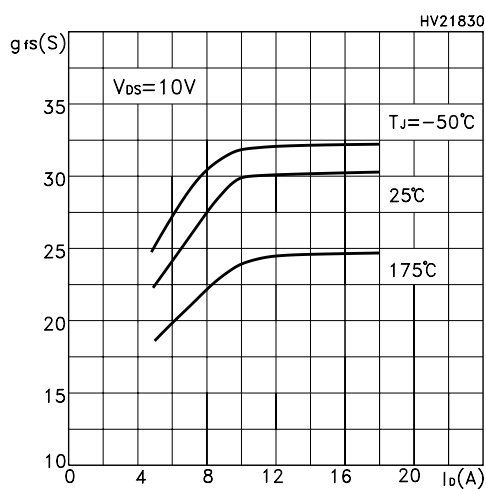


Figure 6: Thermal Impedance

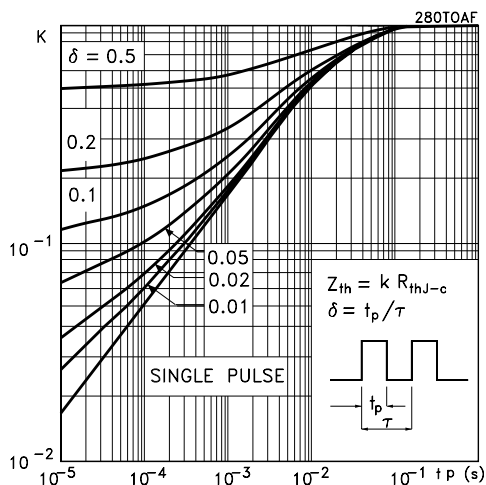


Figure 7: Transfer Characteristics

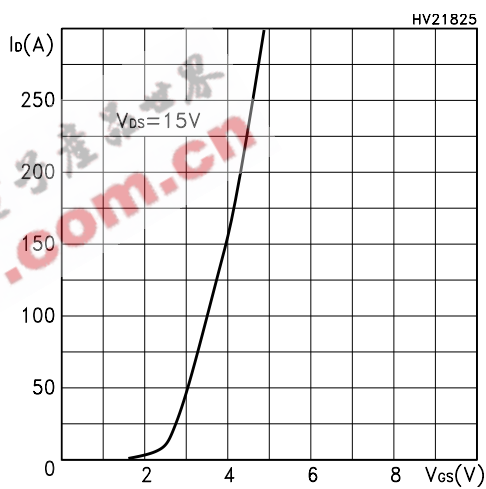


Figure 8: Static Drain-source On Resistance

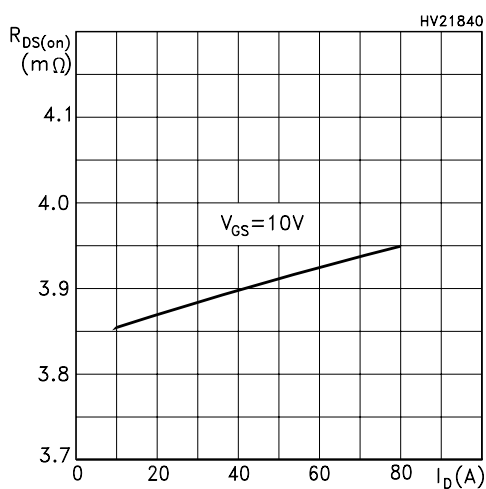


Figure 9: Gate Charge vs Gate-source Voltage

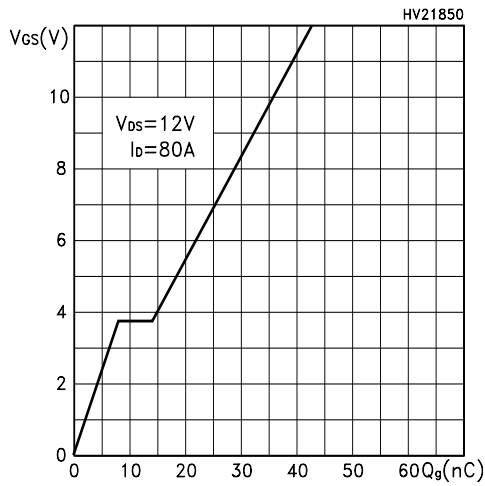


Figure 10: Normalized Gate Threshold Voltage vs Temperature

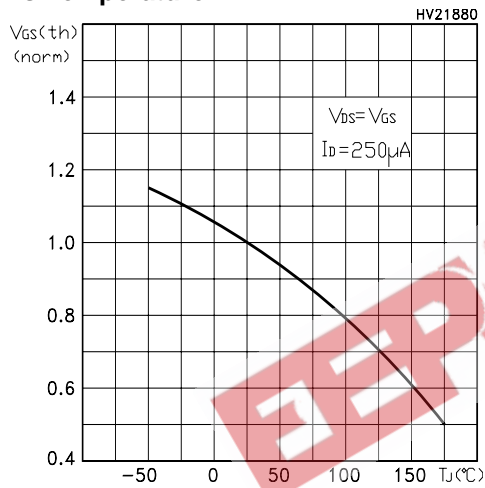


Figure 11: Dource-Drain Diode Forward Characteristics

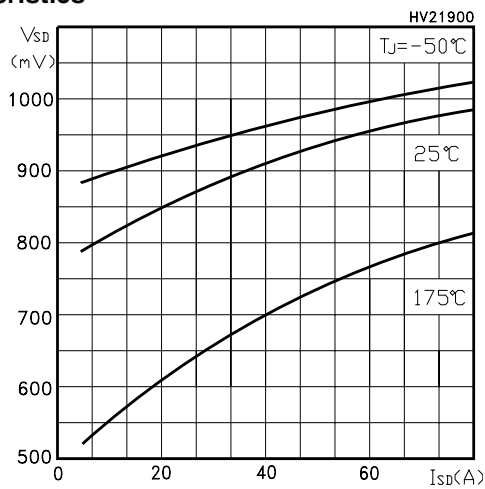


Figure 12: Capacitance Variations

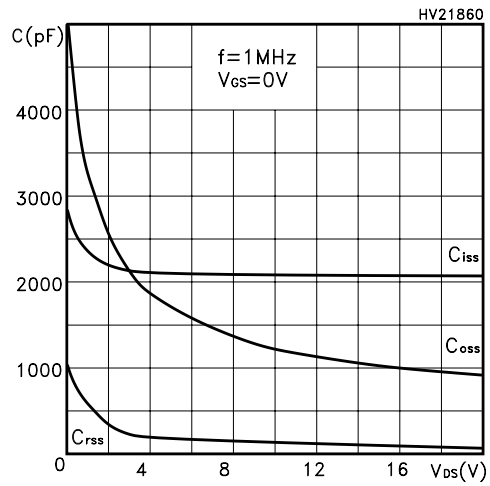


Figure 13: Normalized On Resistance vs Temperature

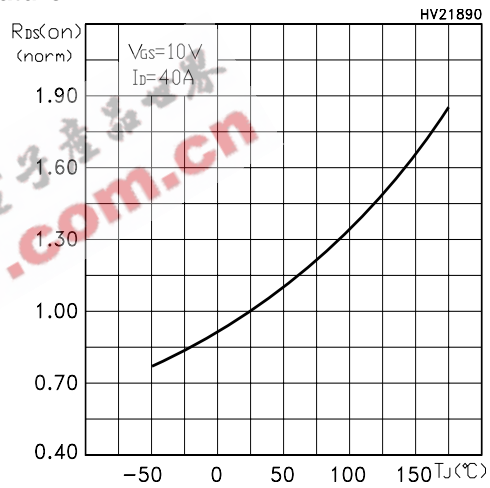


Figure 14: Normalized Breakdown Voltage vs Temperature

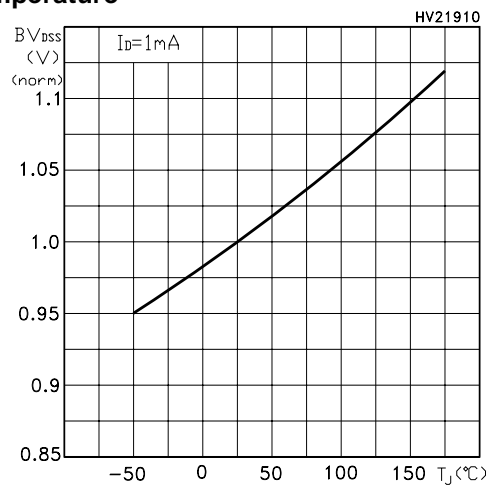


Figure 15: Unclamped Inductive Load Test Circuit

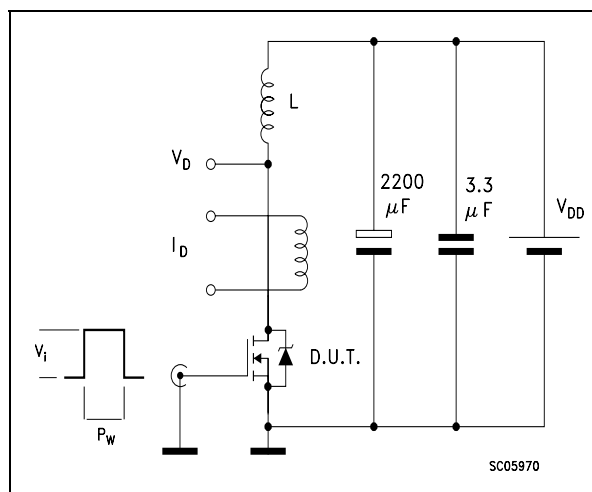


Figure 16: Switching Times Test Circuit For Resistive Load

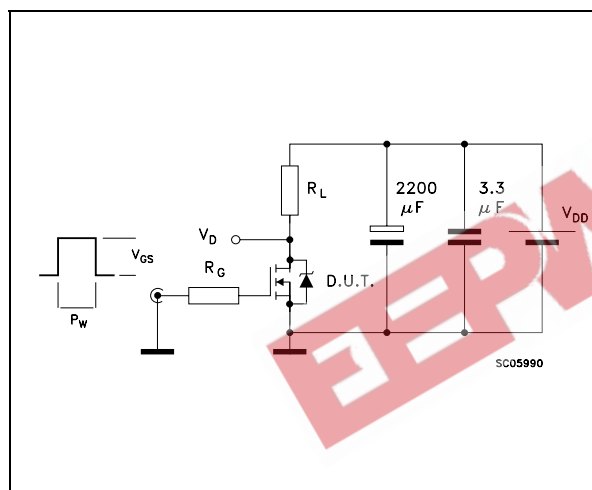


Figure 17: Test Circuit For Inductive Load Switching and Diode Recovery Times

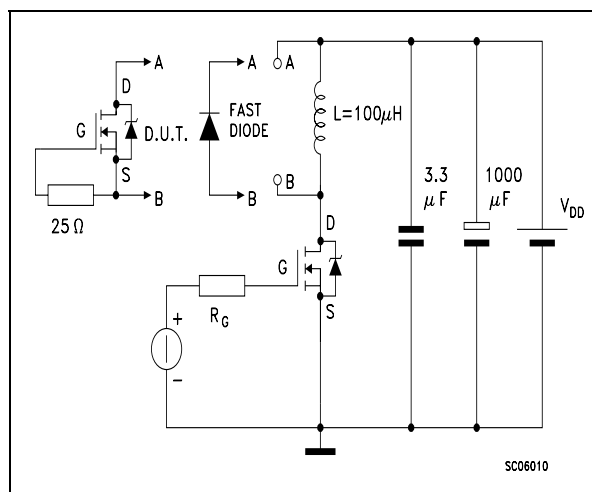


Figure 18: Unclamped Inductive Waferform

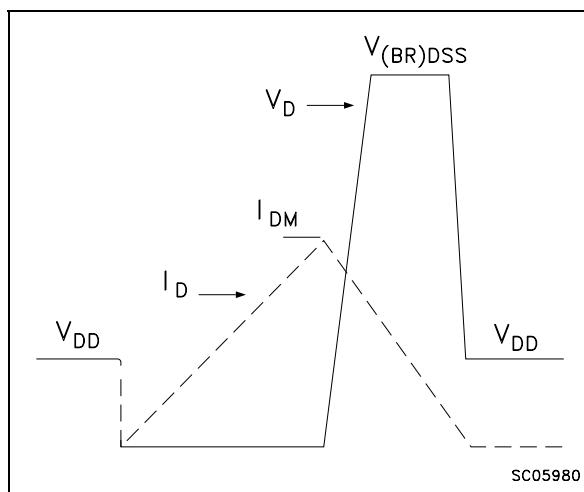
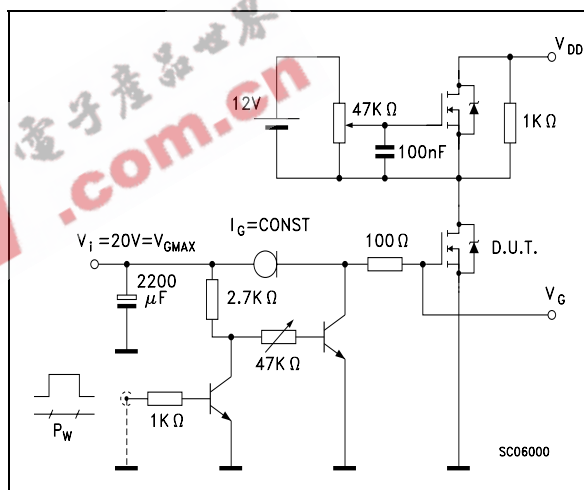
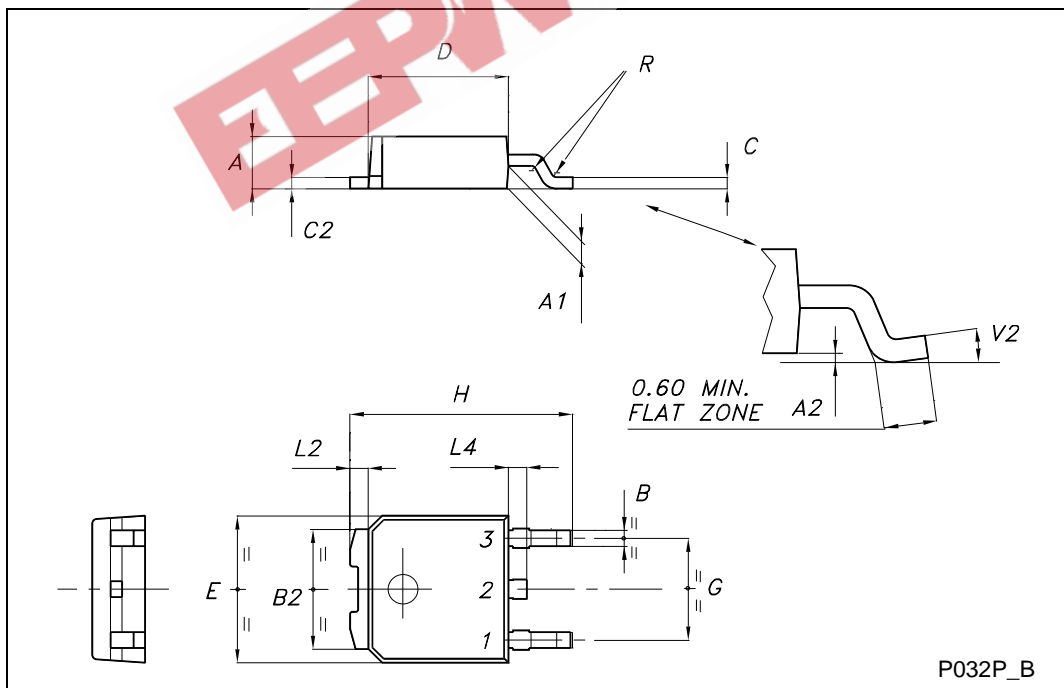


Figure 19: Gate Charge Test Circuit

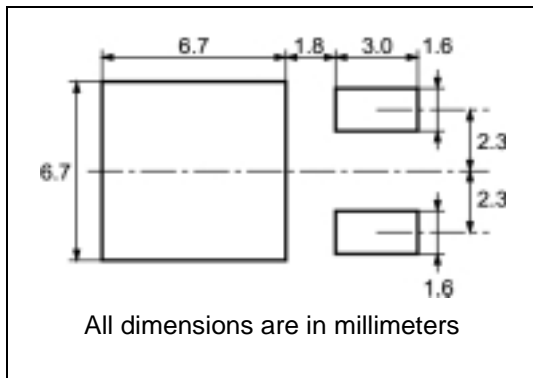


TO-252 (DPAK) MECHANICAL DATA

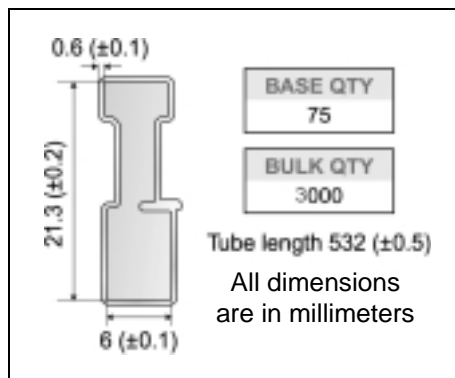
DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	2.20		2.40	0.087		0.094
A1	0.90		1.10	0.035		0.043
A2	0.03		0.23	0.001		0.009
B	0.64		0.90	0.025		0.035
B2	5.20		5.40	0.204		0.213
C	0.45		0.60	0.018		0.024
C2	0.48		0.60	0.019		0.024
D	6.00		6.20	0.236		0.244
E	6.40		6.60	0.252		0.260
G	4.40		4.60	0.173		0.181
H	9.35		10.10	0.368		0.398
L2		0.8			0.031	
L4	0.60		1.00	0.024		0.039
V2	0°		8°	0°		0°



DPAK FOOTPRINT



TUBE SHIPMENT (no suffix)*



TAPE AND REEL SHIPMENT (suffix "T4")*

40 mm min. Access hole at slot location

Full radius

Tape slot in core for tape start 2.5mm min. width

G measured at hub

REEL MECHANICAL DATA

DIM.	mm		inch	
	MIN.	MAX.	MIN.	MAX.
A		330		12.992
B	1.5		0.059	
C	12.8	13.2	0.504	0.520
D	20.2		0.795	
G	16.4	18.4	0.645	0.724
N	50		1.968	
T		22.4		0.881

BASE QTY	BULK QTY
2500	2500

TAPE MECHANICAL DATA

DIM.	mm		inch	
	MIN.	MAX.	MIN.	MAX.
A0	6.8	7	0.267	0.275
B0	10.4	10.6	0.409	0.417
B1		12.1		0.476
D	1.5	1.6	0.059	0.063
D1	1.5		0.059	
E	1.65	1.85	0.065	0.073
F	7.4	7.6	0.291	0.299
K0	2.55	2.75	0.100	0.108
P0	3.9	4.1	0.153	0.161
P1	7.9	8.1	0.311	0.319
P2	1.9	2.1	0.075	0.082
R	40		1.574	
W	15.7	16.3	0.618	0.641

TOP COVER TAPE

10 pitches cumulative tolerance on tape +/- 0.2 mm

Center line of cavity

User Direction of Feed

FEED DIRECTION

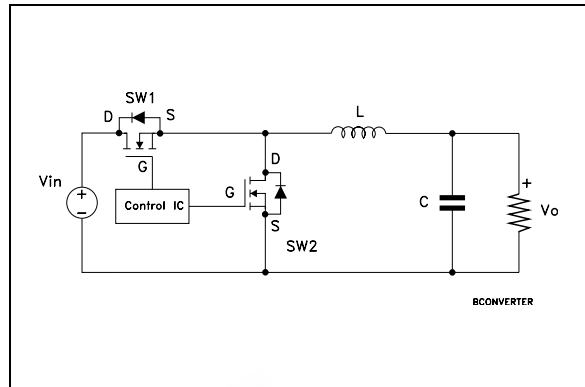
Bending radius R min.

* on sales type

Appendix A: Buck Converter Power Losses Estimation

DESCRIPTION

The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is removed to allow for a safer working junction temperature.



The low side (SW2) device requires:

- Very low $R_{DS(on)}$ to reduce conduction losses
- Small Q_{gls} to reduce the gate charge losses
- Small C_{oss} to reduce losses due to output capacitance
- Small Q_{rr} to reduce losses on SW1 during its turn-on
- The C_{gd}/C_{gs} ratio lower than V_{th}/V_{GG} ratio especially with low drain to source voltage to avoid the cross conduction phenomenon

The high side (SW1) device requires:

- Small R_g and L_s to allow higher gate current peak and to limit the voltage feedback on the gate
- Small Q_g to have a faster commutation and to reduce gate charge losses
- Low $R_{DS(on)}$ to reduce the conduction losses

		High Side Switch (SW1)	Low Side Switch (SW2)
$P_{conduction}$		$R_{DS(on)SW1} * I_L^2 * \delta$	$R_{DS(on)SW2} * I_L^2 * (1-\delta)$
$P_{switching}$		$V_{in} * (Q_{gsth(SW1)} + Q_{gd(SW1)}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P_{diode}	Recovery	Not Applicable	$V_{in} * Q_{rr(SW2)} * f$
	Conduction	Not Applicable	$V_{r(SW2)} * I_L * t_{deadtime} * f$
$P_{gate(Q_g)}$		$Q_{g(SW1)} * V_{gg} * f$	$Q_{g(SW2)} * V_{gg} * f$
P_{Qoss}		$\frac{V_{in} * Q_{oss(SW1)} * f}{2}$	$\frac{V_{in} * Q_{oss(SW2)} * f}{2}$

Parameter	Meaning
δ	Duty-Cycle
Q_{gsth}	Post Threshold Gate Charge
Q_{gls}	Third Quadrant Gate Charge
$P_{conduction}$	On State Losses
$P_{switching}$	On-off Transition Losses
P_{diode}	Conduction and Reverse Recovery Diode Losses
P_{diode}	Gate Drive Losses
P_{Qoss}	Output Capacitance Losses

STD95NH02L

Table 8: Revision History

Date	Revision	Description of Changes
27-Aug-2004	1	First Release.
10-Sep-2004	2	Values changed in table 7

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