



Integrated
Circuit
Systems, Inc.

PRELIMINARY

ICS84326

CRYSTAL-TO-3.3V LVPECL

SERIAL ATTACHED SCSI CLOCK SYNTHESIZER/FANOUT BUFFER

GENERAL DESCRIPTION



The ICS84326 is a Crystal-to-3.3V LVPECL Clock Synthesizer/Fanout Buffer designed for Serial Attached SCSI applications and is a member of the HiPerClockS family of High Performance Clock Solutions from ICS. Using a 25MHz crystal, the 6 LVPECL outputs can be set for either 75MHz or 150MHz using the frequency select pins. The low jitter/low phase noise characteristics make it an ideal clock source for use in Serial Attached SCSI applications or for other applications which require a 75MHz or 150MHz reference clock.

FUNCTION TABLE

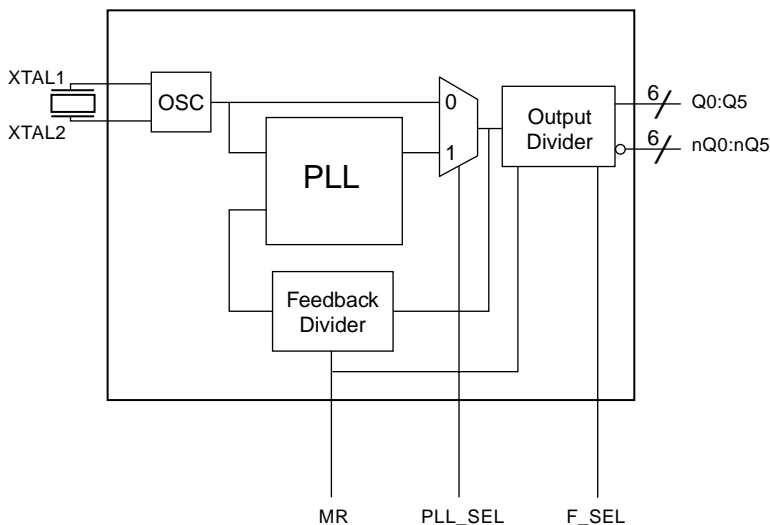
Inputs		Output Frequency
MR	F_SEL	F_OUT
1	X	LOW
0	0	75MHz
0	1	150MHz

FEATURES

- 6 LVPECL outputs
- Crystal oscillator interface
- Output frequency: 75MHz or 150MHz
- Crystal input frequency: 25MHz
- Cycle-to-cycle jitter: 20ps (typical)
- RMS phase jitter at 150MHz, using a 25MHz crystal (899.8KHz to 20MHz): TBD
- Phase noise:

Offset	Noise Power
100Hz	TBD
1KHz	TBD
10KHz	TBD
100KHz	TBD
- Full 3.3V or 3.3V core, 2.5V supply mode
- 0°C to 70°C ambient operating temperature
- Industrial temperature information available upon request

BLOCK DIAGRAM



PIN ASSIGNMENT

Q0	1	24	V _{CC0}
nQ0	2	23	F_SEL
Q1	3	22	nc
nQ1	4	21	MR
Q2	5	20	XTAL1
nQ2	6	19	XTAL2
Q3	7	18	nc
nQ3	8	17	V _{CCA}
Q4	9	16	V _{CC}
nQ4	10	15	PLL_SEL
Q5	11	14	V _{EE}
nQ5	12	13	V _{CC0}

ICS84326

24-Lead, 300-MIL SOIC

7.5mm x 15.33mm x 2.3mm body package

M Package

Top View

The Preliminary Information presented herein represents a product in prototyping or pre-production. The noted characteristics are based on initial product characterization. Integrated Circuit Systems, Incorporated (ICS) reserves the right to change any circuitry or specifications without notice.



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TABLE 1. PIN DESCRIPTIONS

Number	Name	Type		Description
1, 2	Q0, nQ0	Output		Differential output pair. LVPECL interface levels.
3, 4	Q1, nQ1	Output		Differential output pair. LVPECL interface levels.
5, 6	Q2, nQ2	Output		Differential output pair. LVPECL interface levels.
7, 8	Q3, nQ3	Output		Differential output pair. LVPECL interface levels.
9, 10	Q4, nQ4	Output		Differential output pair. LVPECL interface levels.
11, 12	Q5, nQ5	Output		Differential output pair. LVPECL interface levels.
13, 24	V _{CCO}	Power		Output supply pins.
16	V _{CC}	Power		Core supply pin.
14	V _{EE}			Negative supply pin.
15	PLL_SEL	Input	Pullup	Selects between the PLL and crystal inputs as the input to the dividers. When HIGH, selects PLL. When LOW, selects XTAL1, XTAL2. LVCMOS / LVTTTL interface levels.
17	V _{CCA}	Power		Analog supply pin.
18, 22	nc	Unused		No connect.
19, 20	XTAL2, XTAL1	Input		Crystal oscillator interface. XTAL1 is the input. XTAL2 is the output.
21	MR	Input	Pulldown	Active High Master Reset. When logic HIGH, the internal dividers are reset causing the true outputs Qx to go low, and the inverted outputs nQx to go high. When logic LOW, the internal dividers and the outputs are enabled. LVCMOS / LVTTTL interface levels.
23	F_SEL	Input	Pullup	Output frequency select pin. LVCMOS / LVTTTL interface levels.

NOTE: *Pullup* and *Pulldown* refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance				4	pF
R _{PULLUP}	Input Pullup Resistor			51		KΩ
R _{PULLDOWN}	Input Pulldown Resistor			51		KΩ



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ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V_{CC}	4.6V
Inputs, V_I	-0.5V to $V_{CC} + 0.5V$
Outputs, V_O	-0.5V to $V_{CCO} + 0.5V$
Package Thermal Impedance, θ_{JA}	50°C/W (0 lfpm)
Storage Temperature, T_{STG}	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

TABLE 3A. POWER SUPPLY DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		3.135	3.3	3.465	V
V_{CCA}	Analog Supply Voltage		3.135	3.3	3.465	V
V_{CCO}	Output Supply Voltage		3.135	3.3	3.465	V
I_{EE}	Power Supply Current			140		mA
I_{CCA}	Analog Supply Current			20		mA

TABLE 3B. LVCMOS / LVTTTL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{IH}	Input High Voltage	PLL_SEL, MR, F_SEL	2		$V_{CC} + 0.3$	V
V_{IL}	Input Low Voltage	PLL_SEL, MR, F_SEL	-0.3		0.8	V
I_{IH}	Input High Current	MR	$V_{CC} = V_{IN} = 3.465V$		150	μA
		PLL_SEL, F_SEL	$V_{CC} = V_{IN} = 3.465V$		5	μA
I_{IL}	Input Low Current	MR	$V_{CC} = 3.465V, V_{IN} = 0V$	-5		μA
		PLL_SEL, F_SEL	$V_{CC} = 3.465V, V_{IN} = 0V$	-150		μA

TABLE 3C. LVPECL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage; NOTE 1		$V_{CCO} - 1.4$		$V_{CCO} - 1.0$	V
V_{OL}	Output Low Voltage; NOTE 1		$V_{CCO} - 2.0$		$V_{CCO} - 1.7$	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with 50Ω to $V_{CCO} - 2V$.



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TABLE 3D. POWER SUPPLY DC CHARACTERISTICS, $V_{CC} = V_{CCA} = 3.3V \pm 5\%$, $V_{CCO} = 2.5V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		3.135	3.3	3.465	V
V_{CCA}	Analog Supply Voltage		3.135	3.3	3.465	V
V_{CCO}	Output Supply Voltage		2.375	2.5	2.625	V
I_{EE}	Power Supply Current			140		mA
I_{CCA}	Analog Supply Current			20		mA

TABLE 3E. LVCMOS / LVTTTL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = 3.3V \pm 5\%$, $V_{CCO} = 2.5V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{IH}	Input High Voltage	PLL_SEL, MR, F_SEL	2		$V_{CC} + 0.3$	V
V_{IL}	Input Low Voltage	PLL_SEL, MR, F_SEL	-0.3		0.8	V
I_{IH}	Input High Current	MR	$V_{CC} = V_{IN} = 3.465V$		150	μA
		PLL_SEL, F_SEL	$V_{CC} = V_{IN} = 3.465V$		5	μA
I_{IL}	Input Low Current	MR	$V_{CC} = 3.465V, V_{IN} = 0V$	-5		μA
		PLL_SEL, F_SEL	$V_{CC} = 3.465V, V_{IN} = 0V$	-150		μA

TABLE 3F. LVPECL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = 3.3V \pm 5\%$, $V_{CCO} = 2.5V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage; NOTE 1		$V_{CCO} - 1.4$		$V_{CCO} - 1.0$	V
V_{OL}	Output Low Voltage; NOTE 1		$V_{CCO} - 2.0$		$V_{CCO} - 1.7$	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with 50Ω to $V_{CCO} - 2V$.

TABLE 4. CRYSTAL CHARACTERISTICS

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency				25	MHz
Equivalent Series Resistance (ESR)				70	Ω
Shunt Capacitance				7	pF

NOTE: Characterized using an 18pf parallel resonant crystal.



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TABLE 5A. AC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
F_{OUT}	Output Frequency		75		150	MHz
$f_{jit(cc)}$	Cycle-to-Cycle Jitter; NOTE 2			20		ps
$f_{jit(per)}$	Period Jitter, RMS			TBD		ps
$t_{sk(o)}$	Output Skew; NOTE 1, 2			50		ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle			50		%
t_{LOCK}	PLL Lock Time				1	ms

See Parameter Measurement Information section.

NOTE 1: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential crossing points.

NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

TABLE 5B. AC CHARACTERISTICS, $V_{CC} = V_{CCA} = 3.3V \pm 5\%$, $V_{CCO} = 2.5V \pm 5\%$, $T_A = 0^\circ C$ TO $70^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
F_{OUT}	Output Frequency		75		150	MHz
$f_{jit(cc)}$	Cycle-to-Cycle Jitter; NOTE 2			20		ps
$f_{jit(per)}$	Period Jitter, RMS			TBD		ps
$t_{sk(o)}$	Output Skew; NOTE 1, 2			35		ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle			50		%
t_{LOCK}	PLL Lock Time				1	ms

See Parameter Measurement Information section.

NOTE 1: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential crossing points.

NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.



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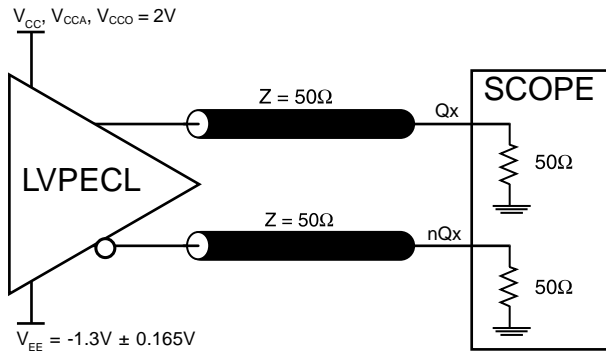
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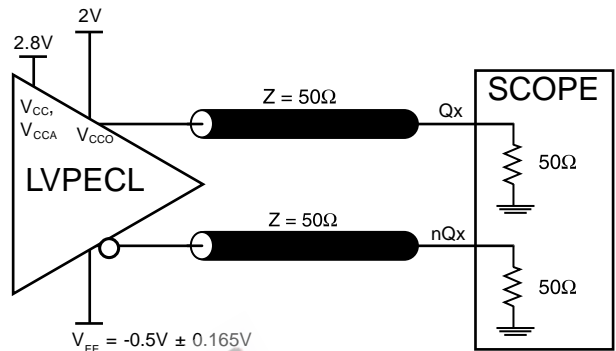
CRYSTAL-TO-3.3V LVPECL

SERIAL ATTACHED SCSI CLOCK SYNTHESIZER/FANOUT BUFFER

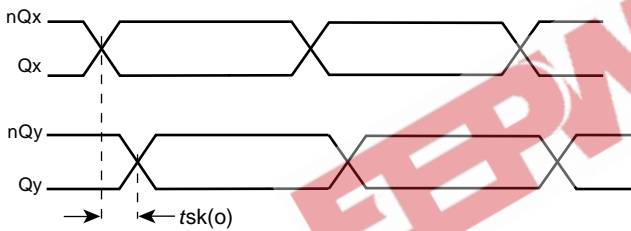
PARAMETER MEASUREMENT INFORMATION



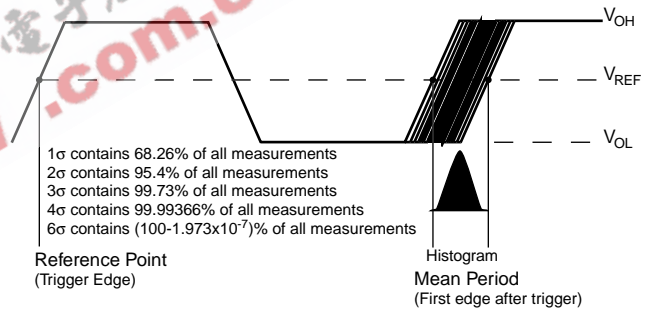
3.3V OUTPUT LOAD AC TEST CIRCUIT



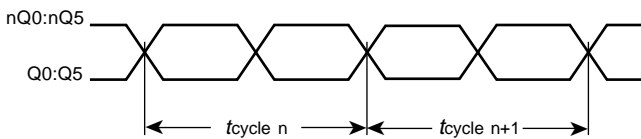
3.3V/2.5V OUTPUT LOAD AC TEST CIRCUIT



OUTPUT SKEW



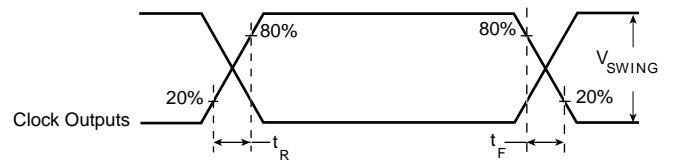
PERIOD JITTER



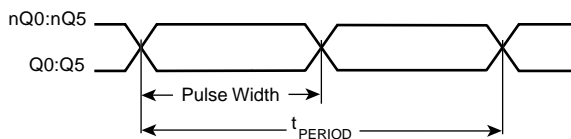
$$t_{jit(cc)} = t_{cycle\ n} - t_{cycle\ n+1}$$

1000 Cycles

CYCLE-TO-CYCLE JITTER



OUTPUT RISE/FALL TIME



$$odc = \frac{t_{PW}}{t_{PERIOD}}$$

odc & t_{PERIOD}



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APPLICATION INFORMATION

POWER SUPPLY FILTERING TECHNIQUES

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The ICS84326 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. V_{CC} , V_{CCA} and V_{CCO} should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. *Figure 1* illustrates how a 20Ω resistor along with a $10\mu\text{F}$ and a $.01\mu\text{F}$ bypass capacitor should be connected to each V_{CCA} pin.

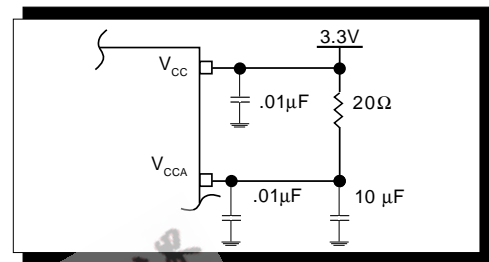


FIGURE 1. POWER SUPPLY FILTERING

TERMINATION FOR 3.3V LVPECL OUTPUTS

The clock layout topology shown below is a typical termination for 3.3V LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

F_{OUT} and nF_{OUT} are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are

designed to drive 50Ω transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 2A and 2B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

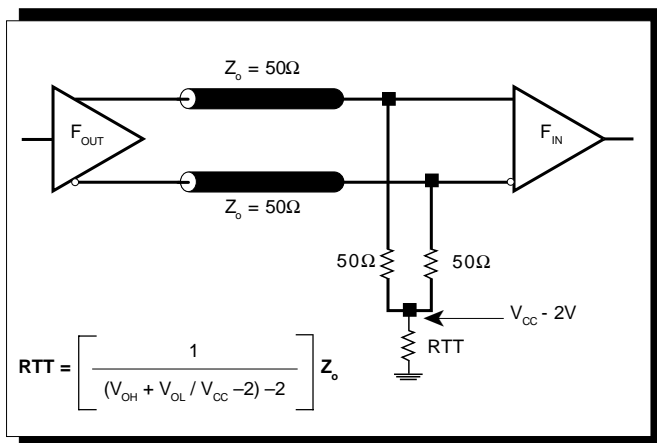


FIGURE 2A. LVPECL OUTPUT TERMINATION

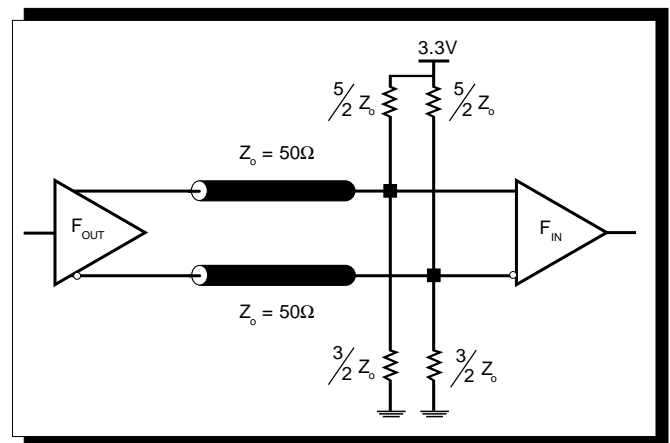


FIGURE 2B. LVPECL OUTPUT TERMINATION



TERMINATION FOR 2.5V LVPECL OUTPUT

Figure 3A and Figure 3B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to $V_{CC} - 2V$. For $V_{CC} = 2.5V$, the $V_{CC} - 2V$ is very close to

ground level. The R3 in Figure 3B can be eliminated and the termination is shown in Figure 3C.

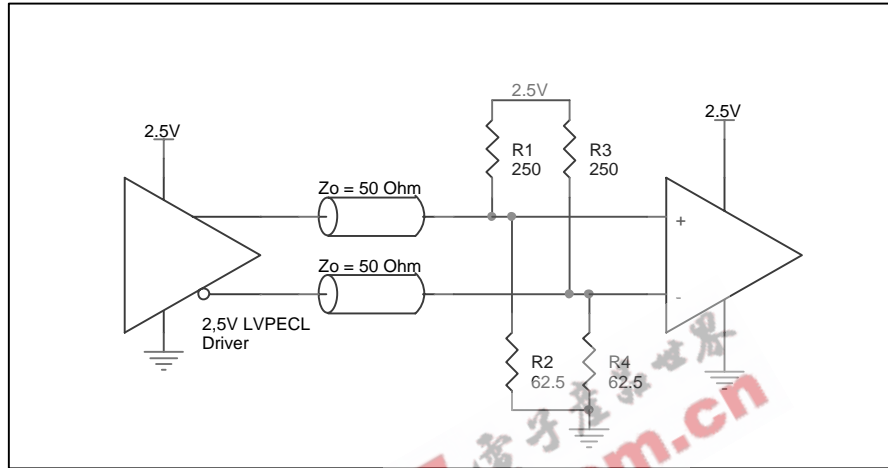


FIGURE 3A. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

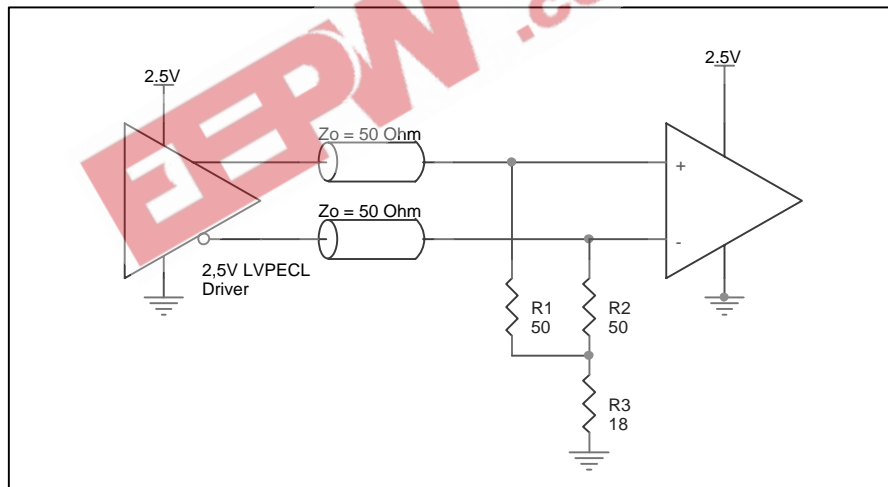


FIGURE 3B. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

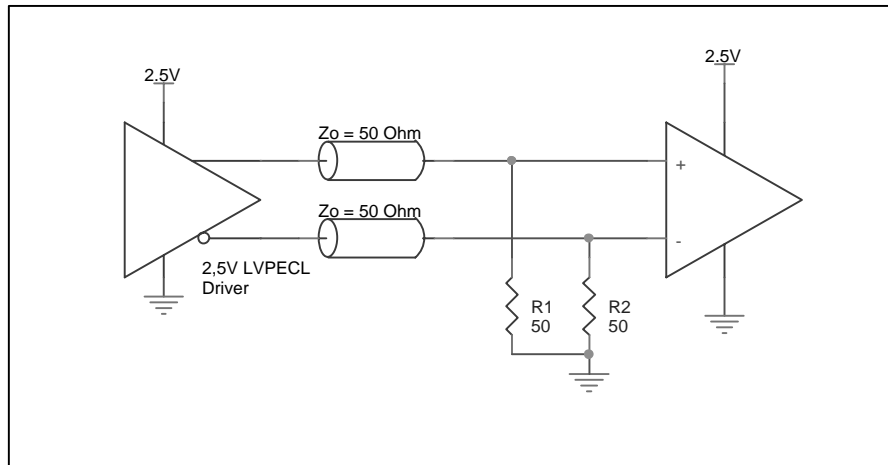


FIGURE 3C. 2.5V LVPECL TERMINATION EXAMPLE



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CRYSTAL INPUT INTERFACE

The ICS84326 has been characterized with 18pF parallel resonant crystals. The capacitor values, C1 and C2, shown in Figure 4 below were determined using a 25MHz, 18pF

parallel resonant crystal and were chosen to minimize the ppm error. The optimum C1 and C2 values can be slightly adjusted for different board layouts.

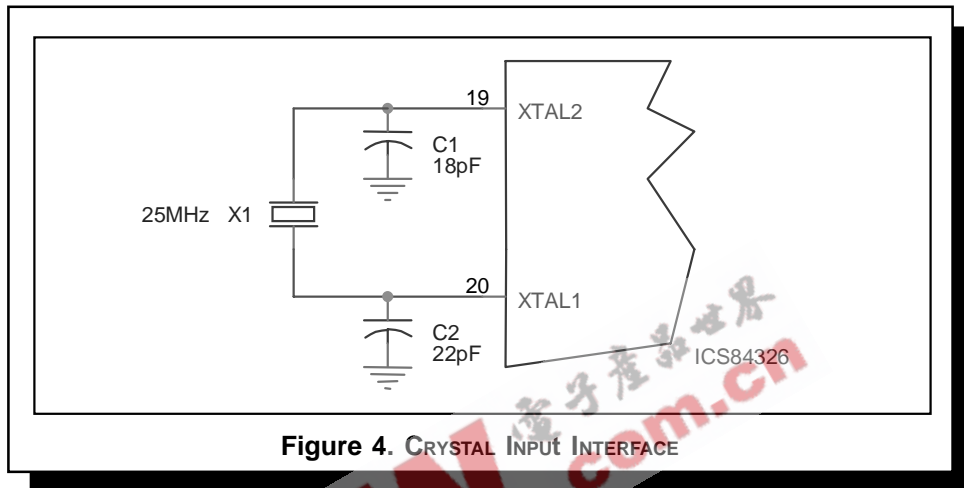


Figure 4. CRYSTAL INPUT INTERFACE

SCHEMATIC EXAMPLE

Figure 5A shows a schematic example of using an ICS84326. In this example, the input is a 25MHz parallel resonant crystal with load capacitor $CL=18pF$. The frequency fine tuning capacitors C1 and C2 is 22pF and 18pF respectively. This example also shows logic control input handling. The configuration is set at $F_SEL=0$, therefore, the output frequency is 150MHz. It is recommended to

have one decouple capacitor per power pin. Each decoupling capacitor should be located as close as possible to the power pin. The low pass filter R7, C11 and C16 for clean analog supply should also be located as close to the V_{CCA} pin as possible.

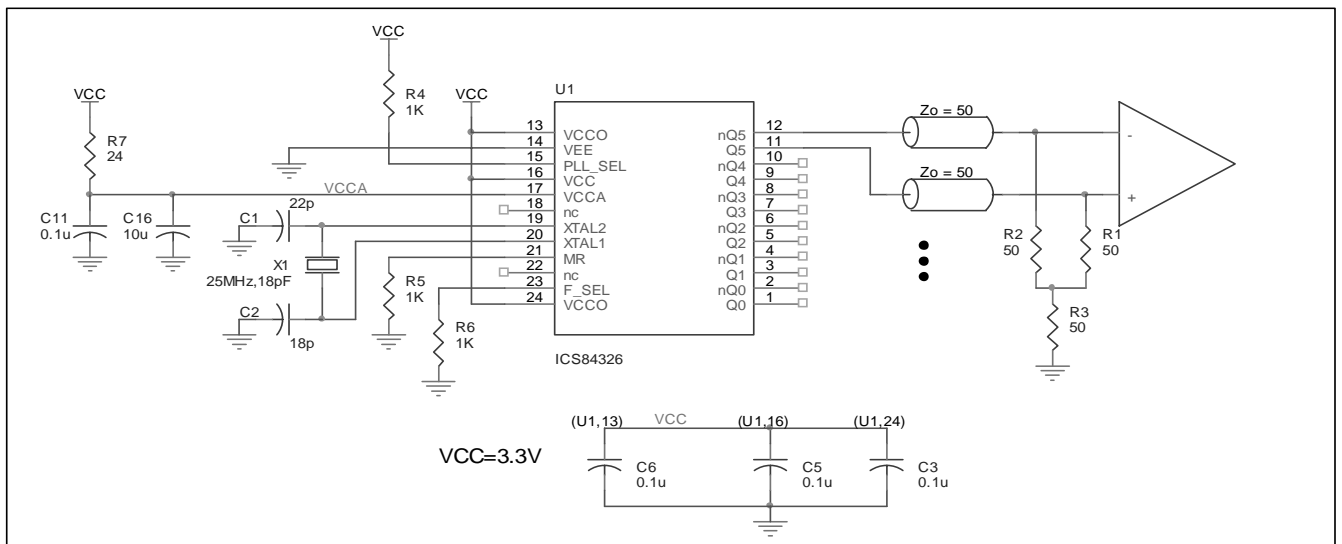


FIGURE 5A. ICS84326 SCHEMATIC EXAMPLE



The following component footprints are used in this layout example:

All the resistors and capacitors are size 0603.

POWER AND GROUNDING

Place the decoupling capacitors C5, C6 and C3, as close as possible to the power pins. If space allows, placement of the decoupling capacitor on the component side is preferred. This can reduce unwanted inductance between the decoupling capacitor and the power pin caused by the via.

Maximize the power and ground pad sizes and number of vias capacitors. This can reduce the inductance between the power and ground planes and the component power and ground pins.

The RC filter consisting of R7, C11, and C16 should be placed as close to the V_{CCA} pin as possible.

CLOCK TRACES AND TERMINATION

Poor signal integrity can degrade the system performance or cause system failure. In synchronous high-speed digital systems, the clock signal is less tolerant to poor signal integrity than other signals. Any ringing on the rising or falling edge or excessive ring back can cause system failure. The shape of the trace and the trace delay might be restricted by the available space on the board and the component location. While routing the traces, the clock signal traces should be routed first and should be locked prior to routing other signal traces.

- The differential 50Ω output traces should have the same length.
- Avoid sharp angles on the clock trace. Sharp angle turns cause the characteristic impedance to change on the transmission lines.
- Keep the clock traces on the same layer. Whenever possible, avoid placing vias on the clock traces. Placement of vias on the traces can affect the trace characteristic impedance and hence degrade signal integrity.
- To prevent cross talk, avoid routing other signal traces in parallel with the clock traces. If running parallel traces is unavoidable, allow a separation of at least three trace widths between the differential clock trace and the other signal trace.
- Make sure no other signal traces are routed between the clock trace pair.
- The matching termination resistors should be located as close to the receiver input pins as possible.

CRYSTAL

The crystal X1 should be located as close as possible to the pins 20 (XTAL1) and 19 (XTAL2). The trace length between the X1 and U1 should be kept to a minimum to avoid unwanted parasitic inductance and capacitance. Other signal traces should not be routed near the crystal traces.

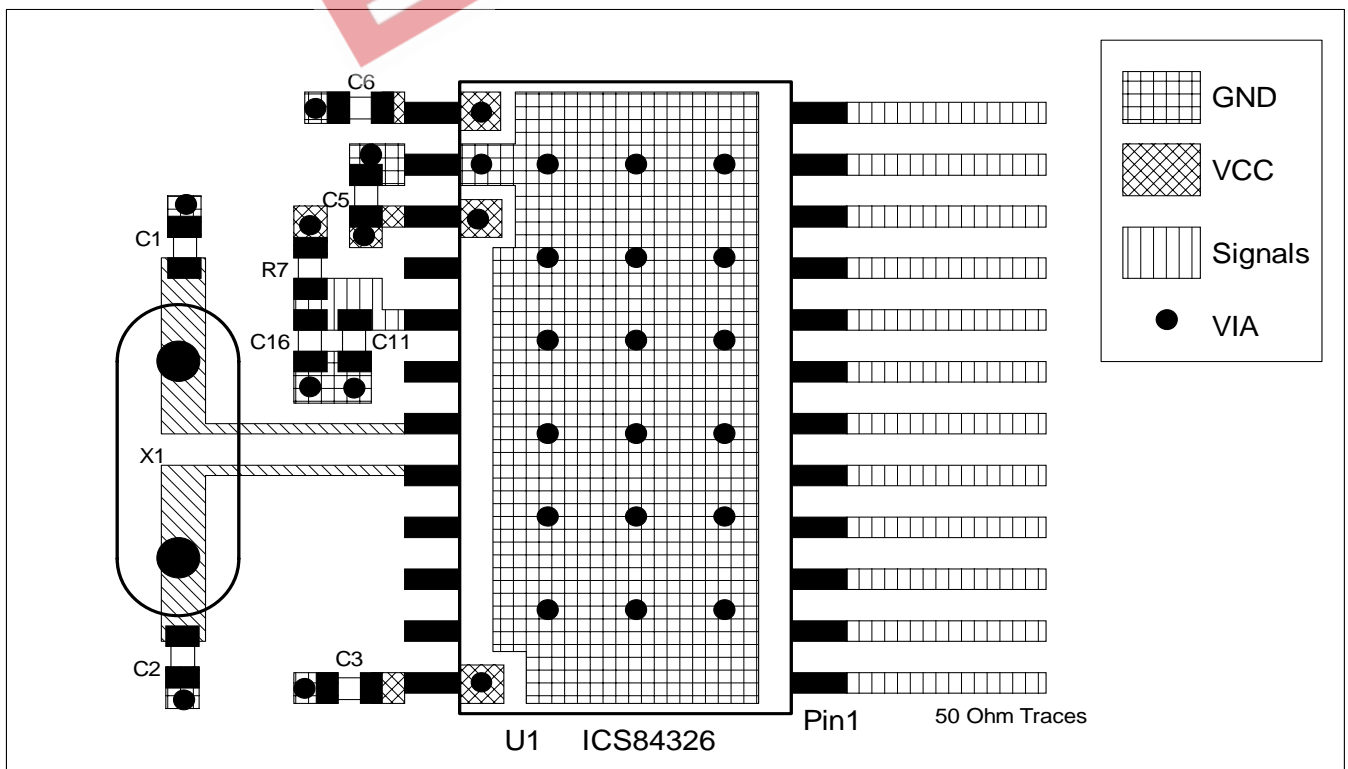


FIGURE 5B. ICS84326 P.C. BOARD LAYOUT EXAMPLE



POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS84326. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS84326 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = $V_{CC_MAX} * I_{EE_MAX} = 3.465V * 140mA = 485mW$
- Power (outputs)_{MAX} = **30.2mW/Loaded Output pair**
If all outputs are loaded, the total power is $6 * 30.2mW = 181mW$

Total Power_{MAX} (3.465V, with all outputs switching) = $485mW + 181mW = 666mW$

2. Junction Temperature.

Junction temperature, T_j , is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS™ devices is 125°C.

The equation for T_j is as follows: $T_j = \theta_{JA} * Pd_total + T_A$

T_j = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 43°C/W per Table 6 below.

Therefore, T_j for an ambient temperature of 70°C with all outputs switching is:

$70°C + 0.666W * 43°C/W = 98.6°C$. This is well below the limit of 125°C.

This calculation is only an example. T_j will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

TABLE 6. THERMAL RESISTANCE θ_{JA} FOR 24-PIN SOIC, FORCED CONVECTION

θ_{JA} by Velocity (Linear Feet per Minute)			
	0	200	500
Multi-Layer PCB, JEDEC Standard Test Boards	50°C/W	43°C/W	38°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.



3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in *Figure 6*.

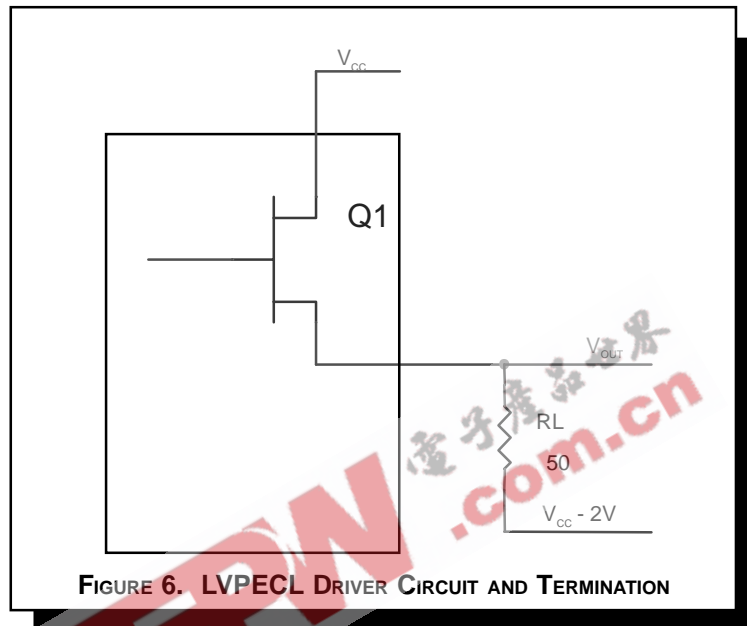


FIGURE 6. LVPECL DRIVER CIRCUIT AND TERMINATION

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of $V_{CCO} - 2V$.

- For logic high, $V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} - 1.0V$

$$(V_{CCO_MAX} - V_{OH_MAX}) = 1.0V$$

- For logic low, $V_{OUT} = V_{OL_MAX} = V_{CCO_MAX} - 1.7V$

$$(V_{CCO_MAX} - V_{OL_MAX}) = 1.7V$$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$$Pd_H = [(V_{OH_MAX} - (V_{CCO_MAX} - 2V))/R_L] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - (V_{CCO_MAX} - V_{OH_MAX}))/R_L] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - 1V)/50\Omega] * 1V = \mathbf{20.0mW}$$

$$Pd_L = [(V_{OL_MAX} - (V_{CCO_MAX} - 2V))/R_L] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - (V_{CCO_MAX} - V_{OL_MAX}))/R_L] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = \mathbf{10.2mW}$$

Total Power Dissipation per output pair = $Pd_H + Pd_L = \mathbf{30.2mW}$



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SERIAL ATTACHED SCSI CLOCK SYNTHESIZER/FANOUT BUFFER

RELIABILITY INFORMATION

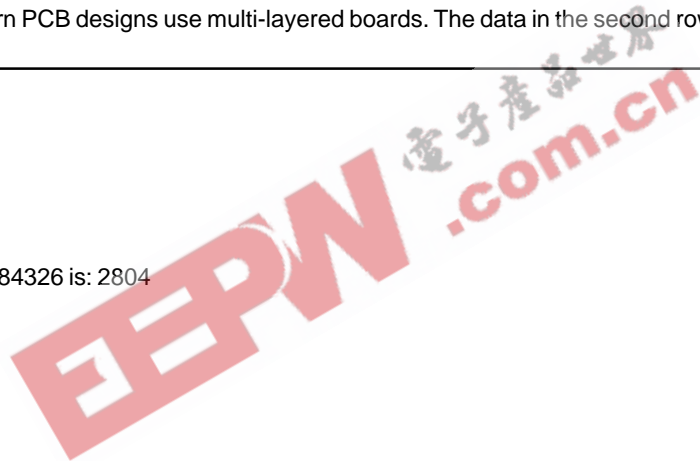
TABLE 7. θ_{JA} VS. AIR FLOW TABLE

θ_{JA} by Velocity (Linear Feet per Minute)			
	0	200	500
Multi-Layer PCB, JEDEC Standard Test Boards	50°C/W	43°C/W	38°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

TRANSISTOR COUNT

The transistor count for ICS84326 is: 2804





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PACKAGE OUTLINE - M SUFFIX

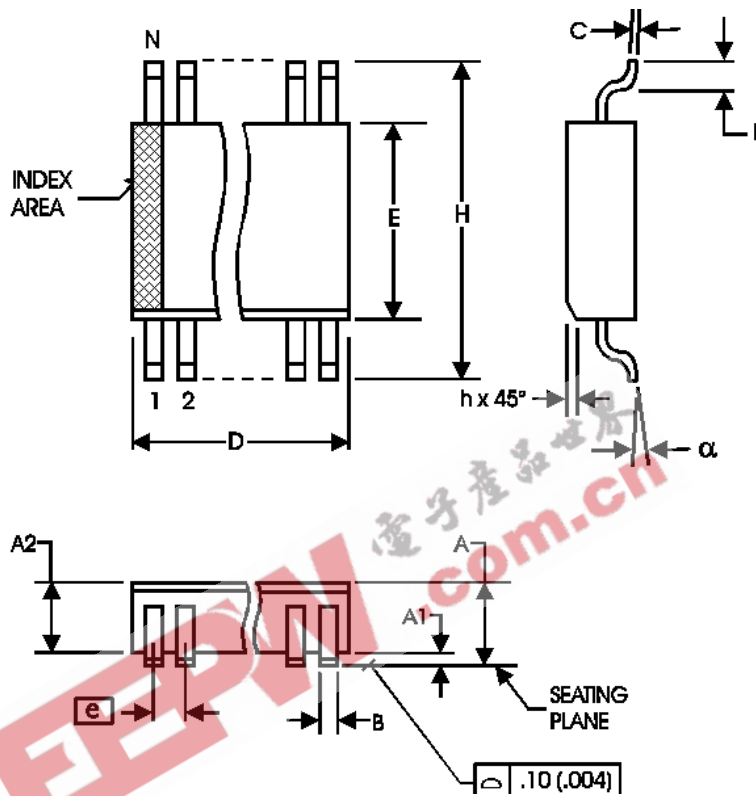


TABLE 8. PACKAGE DIMENSIONS

SYMBOL	Millimeters	
	Minimum	Maximum
N	24	
A	--	2.65
A1	0.10	--
A2	2.05	2.55
B	0.33	0.51
C	0.18	0.32
D	15.20	15.85
E	7.40	7.60
e	1.27 BASIC	
H	10.00	10.65
h	0.25	0.75
L	0.40	1.27
α	0°	8°

Reference Document: JEDEC Publication 95, MS-013, MO-119



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TABLE 9. ORDERING INFORMATION

Part/Order Number	Marking	Package	Count	Temperature
ICS84326AM	ICS84326AM	24 Lead SOIC	30 per tube	0°C to 70°C
ICS84326AMT	ICS84326AM	24 Lead SOIC on Tape and Reel	1000	0°C to 70°C

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