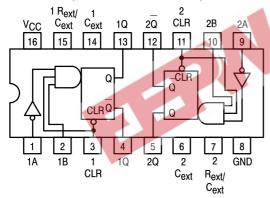


RETRIGGERABLE MONOSTABLE **MULTIVIBRATORS**

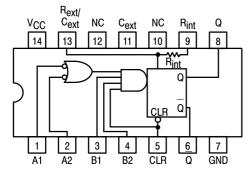
These dc triggered multivibrators feature pulse width control by three methods. The basic pulse width is programmed by selection of external resistance and capacitance values. The LS122 has an internal timing resistor that allows the circuits to be used with only an external capacitor. Once triggered, the basic pulse width may be extended by retriggering the gated low-level-active (A) or high-level-active (B) inputs, or be reduced by use of the overriding clear.

- Overriding Clear Terminates Output Pulse
- Compensated for V_{CC} and Temperature Variations
- DC Triggered from Active-High or Active-Low Gated Logic Inputs
- Retriggerable for Very Long Output Pulses, up to 100% Duty Cycle
- Internal Timing Resistors on LS122

SN54/74LS123 (TOP VIEW) (SEE NOTES 1 THRU 4)



SN54/74LS122 (TOP VIEW) (SEE NOTES 1 THRU 4)



NC - NO INTERNAL CONNECTION.

- 1. An external timing capacitor may be connected between $C_{\mbox{ext}}$ and $R_{\mbox{ext}}/C_{\mbox{ext}}$ (positive).
- 2. To use the internal timing resistor of the LS122, connect $R_{\mbox{\scriptsize int}}$ to $V_{\mbox{\scriptsize CC}}.$
- 3. For improved pulse width accuracy connect an external resistor between Rext/Cext and V_{CC} with R_{int} open-circuited.
- 4. To obtain variable pulse widths, connect an external variable resistance between R_{int}/C_{ext} and V_{CC}.

SN54/74LS122 SN54/74LS123

RETRIGGERABLE MONOSTABLE **MULTIVIBRATORS**

LOW POWER SCHOTTKY



J SUFFIX CERAMIC CASE 620-09



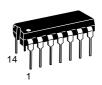
N SUFFIX PLASTIC CASE 648-08



D SUFFIX SOIC CASE 751B-03



J SUFFIX CERAMIC CASE 632-08



N SUFFIX PLASTIC CASE 646-06



D SUFFIX SOIC CASE 751A-02

ORDERING INFORMATION

SN54LSXXXJ Ceramic SN74LSXXXN Plastic SN74LSXXXD SOIC

LS122 FUNCTIONAL TABLE

	INPUTS					
CLEAR	A 1	A2	B1	B2	Q	Q
L	Х	Х	Х	Х	L	Н
X	Н	Н	Χ	X	L	Н
X	Х	Χ	L	Χ	L	Н
X	Х	Χ	Χ	L	L	Н
Н	L	Χ	\uparrow	Н	л	ъ
Н	L	Χ	Н	\uparrow	几	ъ
Н	Х	L	\uparrow	Н	л	T
Н	Х	L	Н	\uparrow	л	ъ
Н	Н	\downarrow	Н	Н	$ \mathcal{L} $	ъ
Н	\downarrow	\downarrow	Н	Н	л	ъ
Н	\downarrow	Н	Н	Н	л	ъ
1	L	Χ	Н	Н	л	ъ
1	Χ	L	Н	Н	л	ъ

LS123 FUNCTIONAL TABLE

INF	OUT	PUTS		
CLEAR	Α	В	Q	Q
L	Х	Х	L	Н
Х	Н	Χ	L	Н
Х	Х	L	L	Н
Н	L	\uparrow		T
Н	\downarrow	Н	л	ъ
1	L	Н	л	ъ

TYPICAL APPLICATION DATA

The output pulse tw is a function of the external components, C_{ext} and R_{ext} or C_{ext} and R_{int} on the LS122. For values of $C_{ext} \ge 1000$ pF, the output pulse at $V_{CC} = 5.0$ V and $V_{RC} = 5.0$ V (see Figures 1, 2, and 3) is given by

 $t_W = K R_{ext} C_{ext}$ where K is nominally 0.45

If C_{ext} is on pF and R_{ext} is in k Ω then twy is in nanoseconds. The C_{ext} terminal of the LS122 and LS123 is an internal connection to ground, however for the best system performance C_{ext} should be hard-wired to ground.

Care should be taken to keep R_{ext} and C_{ext} as close to the monostable as possible with a minimum amount of inductance between the R_{ext}/C_{ext} junction and the R_{ext}/C_{ext} pin. Good groundplane and adequate bypassing should be designed into the system for optimum performance to insure that no false triggering occurs.

It should be noted that the C_{ext} pin is internally connected to ground on the LS122 and LS123, but not on the LS221. Therefore, if C_{ext} is hard-wired externally to ground, substitution of a LS221 onto a LS123 socket will cause the LS221 to become non-functional.

The switching diode is not needed for electrolytic capacitance application and should not be used on the LS122 and LS123.

To find the value of K for $C_{ext} \ge 1000$ pF, refer to Figure 4. Variations on VCC or VRC can cause the value of K to change, as can the temperature of the LS123, LS122. Figures 5 and 6 show the behavior of the circuit shown in Figures 1 and 2 if



separate power supplies are used for V_{CC} and V_{RC}. If V_{CC} is tied to V_{RC}, Figure 7 shows how K will vary with V_{CC} and temperature. Remember, the changes in R_{ext} and C_{ext} with temperature are not calculated and included in the graph.

As long as $C_{ext} \ge 1000$ pF and $5K \le R_{ext} \le 260K$ (SN74LS122/123) or $5K \le R_{ext} \le 160$ K (SN54LS122/123), the change in K with respect to R_{ext} is negligible.

If $C_{ext} \le 1000$ pF the graph shown on Figure 8 can be used to determine the output pulse width. Figure 9 shows how K will change for $C_{ext} \le 1000$ pF if V_{CC} and V_{RC} are connected to the same power supply. The pulse width t_W in nanoseconds is approximated by

$$t_W = 6 + 0.05 C_{ext} (pF) + 0.45 R_{ext} (k\Omega) C_{ext} + 11.6 R_{ext}$$

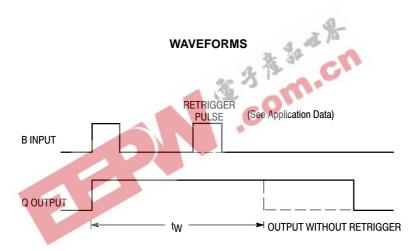
In order to trim the output pulse width, it is necessary to include a variable resistor between VCC and the $R_{\mbox{ext}}/C_{\mbox{ext}}$ pin or between VCC and the $R_{\mbox{ext}}$ pin of the LS122. Figure 10, 11, and 12 show how this can be done. $R_{\mbox{ext}}$ remote should be kept as close to the monostable as possible.

Retriggering of the part, as shown in Figure 3, must not occur before C_{ext} is discharged or the retrigger pulse will not have any effect. The discharge time of C_{ext} in nanoseconds is guaranteed to be less than 0.22 C_{ext} (pF) and is typically 0.05 C_{ext} (pF).

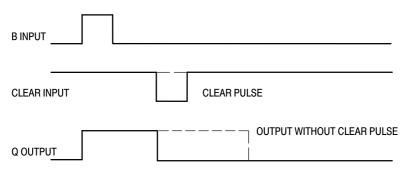
For the smallest possible deviation in output pulse widths from various devices, it is suggested that C_{ext} be kept $\geq 1000 \ \text{pF}.$

GUARANTEED OPERATING RANGES

Symbol	Parameter		Min	Тур	Max	Unit
Vcc	Supply Voltage	54 74	4.5 4.75	5.0 5.0	5.5 5.25	V
T _A	Operating Ambient Temperature Range	54 74	-55 0	25 25	125 70	°C
loн	Output Current — High	54, 74			-0.4	mA
loL	Output Current — Low	54 74			4.0 8.0	mA
R _{ext}	External Timing Resistance	54 74	5.0 5.0		180 260	kΩ
C _{ext}	External Capacitance	54, 74	No Restriction			-
R _{ext} /C _{ext}	Wiring Capacitance at R _{ext} /C _{ext} Terminal	54, 74			50	pF



EXTENDING PULSE WIDTH



OVERRIDING THE OUTPUT PULSE

DC CHARACTERISTICS OVER OPERATING TEMPERATURE RANGE (unless otherwise specified)

				Limits				
Symbol	Parameter		Min	Тур	Max	Unit	Test C	onditions
VIH	Input HIGH Voltage		2.0			V	Guaranteed Inpu All Inputs	ut HIGH Voltage for
V	Input I OW Voltage	54			0.7	V	Guaranteed Input LOW Voltage for All Inputs	
V _{IL}	Input LOW Voltage	74			8.0	V		
VIK	Input Clamp Diode Voltage			-0.65	-1.5	V	V _{CC} = MIN, I _{IN}	= -18 mA
\/	Outrot HIGH Voltage	54	2.5	3.5		V	V_{CC} = MIN, I_{OH} = MAX, V_{IN} = V_{IH} or V_{IL} per Truth Table	
VOH	Output HIGH Voltage	74	2.7	3.5		V		
	Output LOW Valtage	54, 74		0.25	0.4	V	I _{OL} = 4.0 mA	$V_{CC} = V_{CC} MIN,$ $V_{IN} = V_{IL} \text{ or } V_{IH}$
VOL	Output LOW Voltage	74		0.35	0.5	V	I _{OL} = 8.0 mA	per Truth Table
	lament I II Cl I Commant				20	μΑ	V _{CC} = MAX, V _{II}	v = 2.7 V
l ¹IH	Input HIGH Current				0.1	mA	VCC = MAX, VII	v = 7.0 V
Ι _Ι L	Input LOW Current				-0.4	mA	V _{CC} = MAX, V _{IN} = 0.4 V	
los	Short Circuit Current (Note 1)		-20		-100	mA	V _{CC} = MAX	
laa	Dower Cumply Current	LS122		80	11	-0-0		
lcc	Power Supply Current	LS123			20	mA	VCC = MAX	

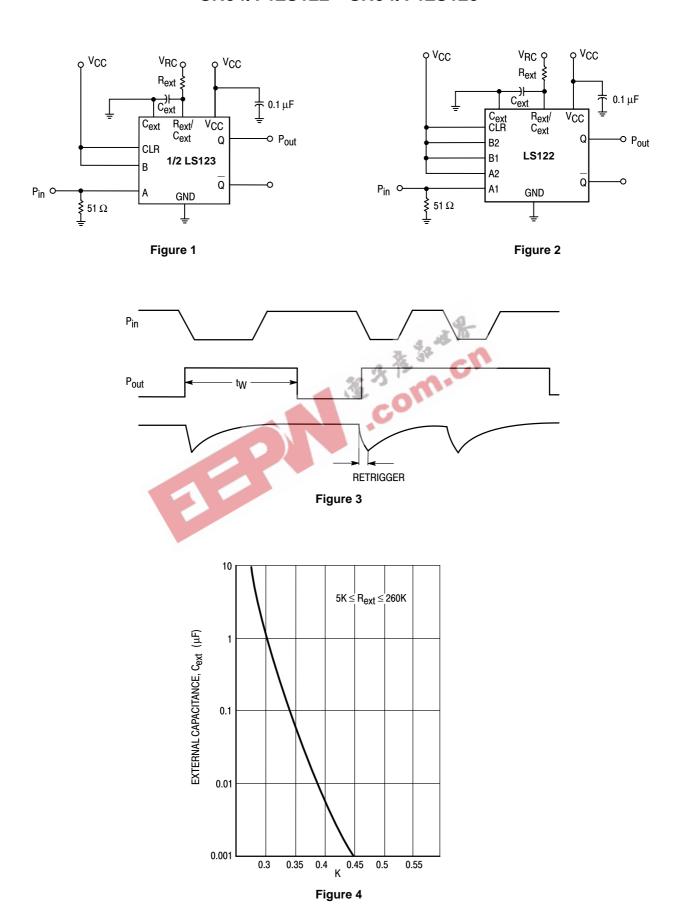
Note 1: Not more than one output should be shorted at a time, nor for more than 1 second.

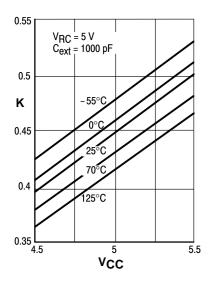
AC CHARACTERISTICS (T_A = 25°C, V_{CC} = 5.0 V)

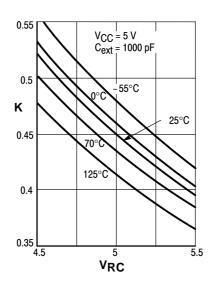
			Limits			
Symbol	Parameter	Min	Тур	Max	Unit	Test Conditions
tPLH	Propagation Delay, A to Q		23	33	20	
^t PHL	Propagation Delay, A to Q		32	45	ns	C _{ext} = 0
tPLH	Propagation Delay, B to Q		23	44	20	C _L = 15 pF
^t PHL	Propagation Delay, B to Q		34	56	ns	$R_{\text{ext}} = 5.0 \text{ k}\Omega$
tPLH	Propagation Delay, Clear to Q		28	45	20	$R_L = 2.0 \text{ k}\Omega$
^t PHL	Propagation Delay, Clear to Q		20	27	ns	
t _W min	A or B to Q		116	200	ns	$C_{ext} = 1000 \text{ pF}, R_{ext} = 10 \text{ k}\Omega,$
t _W Q	A to B to Q	4.0	4.5	5.0	μs	$C_L = 15 \text{ pF}, R_L = 2.0 \text{ k}\Omega$

AC SETUP REQUIREMENTS (T_A = 25°C, V_{CC} = 5.0 V)

		Limits				
Symbol	Parameter	Min	Тур	Max	Unit	Test Conditions
tw	Pulse Width	40			ns	







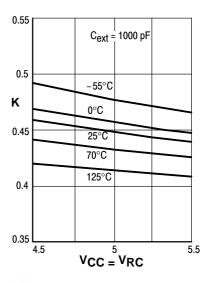


Figure 5. K versus VCC

Figure 6. K versus V_{RC}

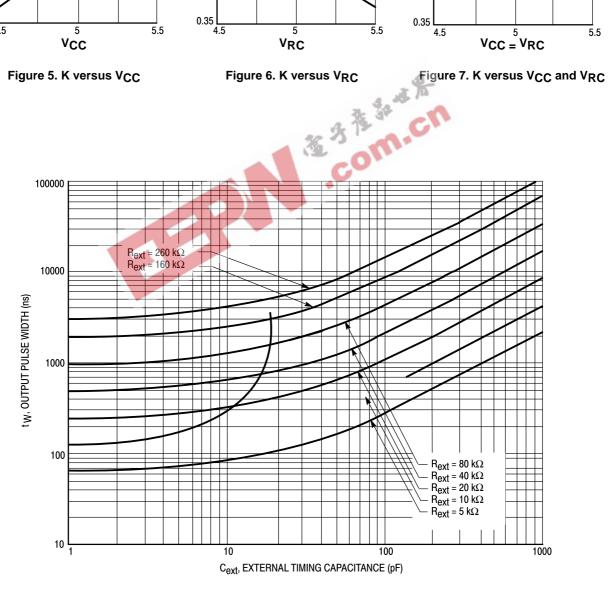
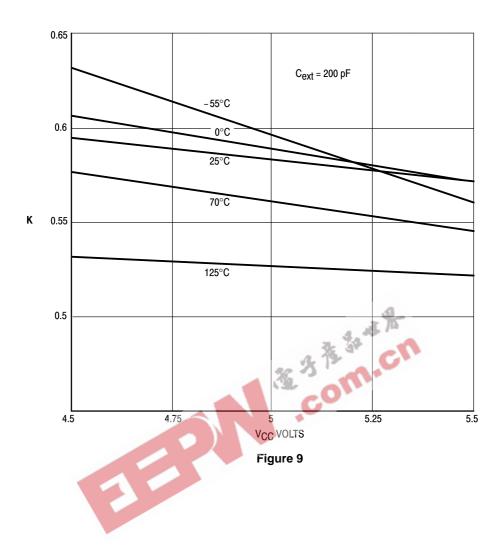


Figure 8



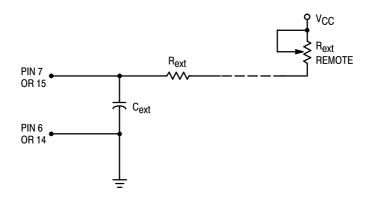


Figure 10. LS123 Remote Trimming Circuit

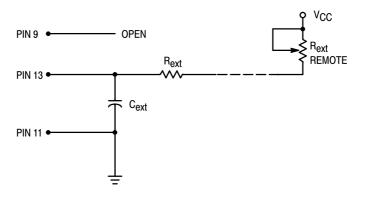


Figure 11. LS122 Remote Trimming Circuit Without Rext

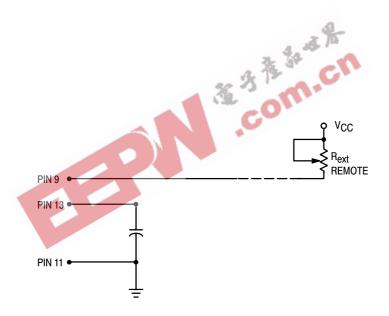
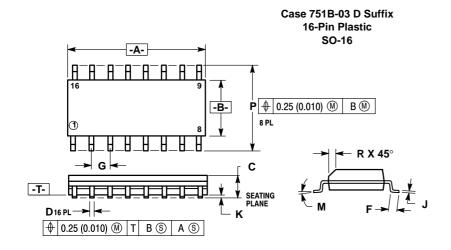


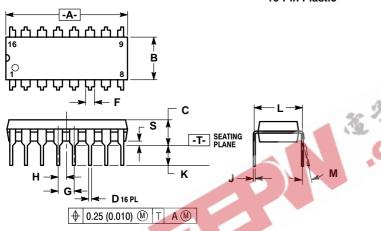
Figure 12. LS122 Remote Trimming Circuit with Rint



- NOTES:
 1. DIMENSIONING AND TOLERANCING PER ANSI
- Y14.5M, 1982.
 CONTROLLING DIMENSION: MILLIMETER.
 DIMENSION A AND B DO NOT INCLUDE MOLD 3. PROTRUSION.
 MAXIMUM MOLD PROTRUSION 0.15 (0.006)
- PER SIDE
- 751B-01 IS OBSOLETE, NEW STANDARD 751B-03.

	MILLIM	ETERS	INC	HES	
DIM	MIN	MAX	MIN	MAX	
Α	9.80	10.00	0.386	0.393	
В	3.80	4.00	0.150	0.157	
С	1.35	1.75	0.054	0.068	
D	0.35	0.49	0.014	0.019	
F	0.40	1.25	0.016	0.049	
G	1.27	BSC	0.050 BSC		
J	0.19	0.25	0.008	0.009	
K	0.10	0.25	0.004	0.009	
M	0°	7°	0°	7°	
P	5.80	6.20	0.229	0.244	
R	0.25	0.50	0.010	0.019	

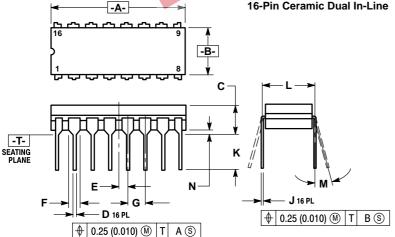
Case 648-08 N Suffix 16-Pin Plastic



NOTES:

ouffix c	1. DIMENS Y14.5M, 2. CONTR 3. DIMENS FORME 4. DIMENS FLASH. 5. ROUND 6. 648-01 648-08.	, 1982. Olling I Sion "L" T D Paral Sion "B" I Ed Cori	DIMENSIO TO CENTI LEL. DOES NO NERS OP	ON: INCH ER OF LE T INCLUI	ADS WH	EN
78. a	W 1	MILLIM	ETERS	INC	HES]
13	DIM	MIN	MAX	MIN	MAX	
90 43	A	18.80	19.55	0.740	0.770	
130	В	6.35	6.85	0.250	0.270	
	С	3.69	4.44	0.145	0.175	
	D	0.39	0.53	0.015	0.021	
	F	1.02	1.77	0.040	0.070	
	G	2.54	BSC	0.100	BSC	
M	Н	1.27	BSC	0.050	BSC	
	J	0.21	0.38	0.008	0.015	
	K	2.80	3.30	0.110	0.130	
	L	7.50	7.74	0.295	0.305	
	M	0°	10°	0°	10°	
	S	0.51	1.01	0.020	0.040	

Case 620-09 J Suffix 16-Pin Ceramic Dual In-Line



NOTES:

- OTES:
 1. DIMENSIONING AND TOLERANCING PER ANSI
 Y14.5M, 1982.
 2. CONTROLLING DIMENSION: INCH.
 3. DIMENSION L TO CENTER OF LEAD WHEN
 FORMED PARALLEL.

- DIM F MAY NARROW TO 0.76 (0.030) WHERE THE LEAD ENTERS THE CERAMIC BODY.
 5. 620-01 THRU -08 OBSOLETE, NEW STANDARD

	MILLIM	ETERS	INC	HES	
DIM	MIN	MAX	MIN	MAX	
Α	19.05	19.55	0.750	0.770	
В	6.10	7.36	0.240	0.290	
С	-	4.19	_	0.165	
D	0.39	0.53	0.015	0.021	
E	1.27	BSC	0.050	BSC	
F	1.40	1.77	0.055	0.070	
G	2.54	2.54 BSC		BSC	
J	0.23	0.27	0.009	0.011	
K	_	5.08	_	0.200	
L	7.62	BSC	0.300	BSC	
М	0°	15°	0°	15°	
N	0.39	0.88	0.015	0.035	



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