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# ZN429E8/ZN429D

## LOW COST 8-BIT D-A CONVERTER

The ZN429 is a monolithic 8-bit D-A converter containing an R-2R ladder network of diffused resistors with precision bipolar switches.

### FEATURES

- Linearity Error  $\pm 1/2$  LSB
- Single +5V Supply
- Low Power Consumption 25mW Typical
- Settling Time 1 Microsecond Typical
- TTL and 5V CMOS Compatible
- Designed for Low Cost Applications

### ABSOLUTE MAXIMUM RATINGS

Supply voltage, $V_{CC}$	+7.0V
Max. voltage, logic and $V_{REF}$ inputs	+5.5V
Storage temperature range	-55°C to +125°C

### ORDERING INFORMATION

Ambient operating temperature	-40°C to +85°C
Package, ZN429D	MP14
Package, ZN429E8	DP14

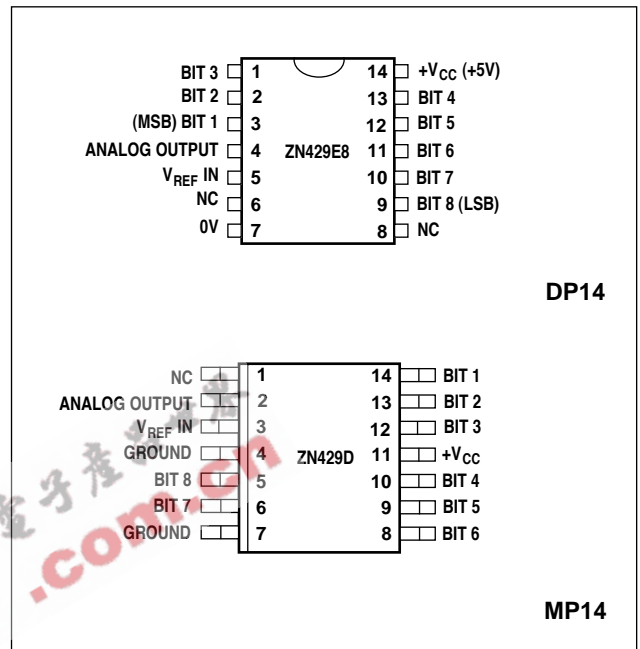
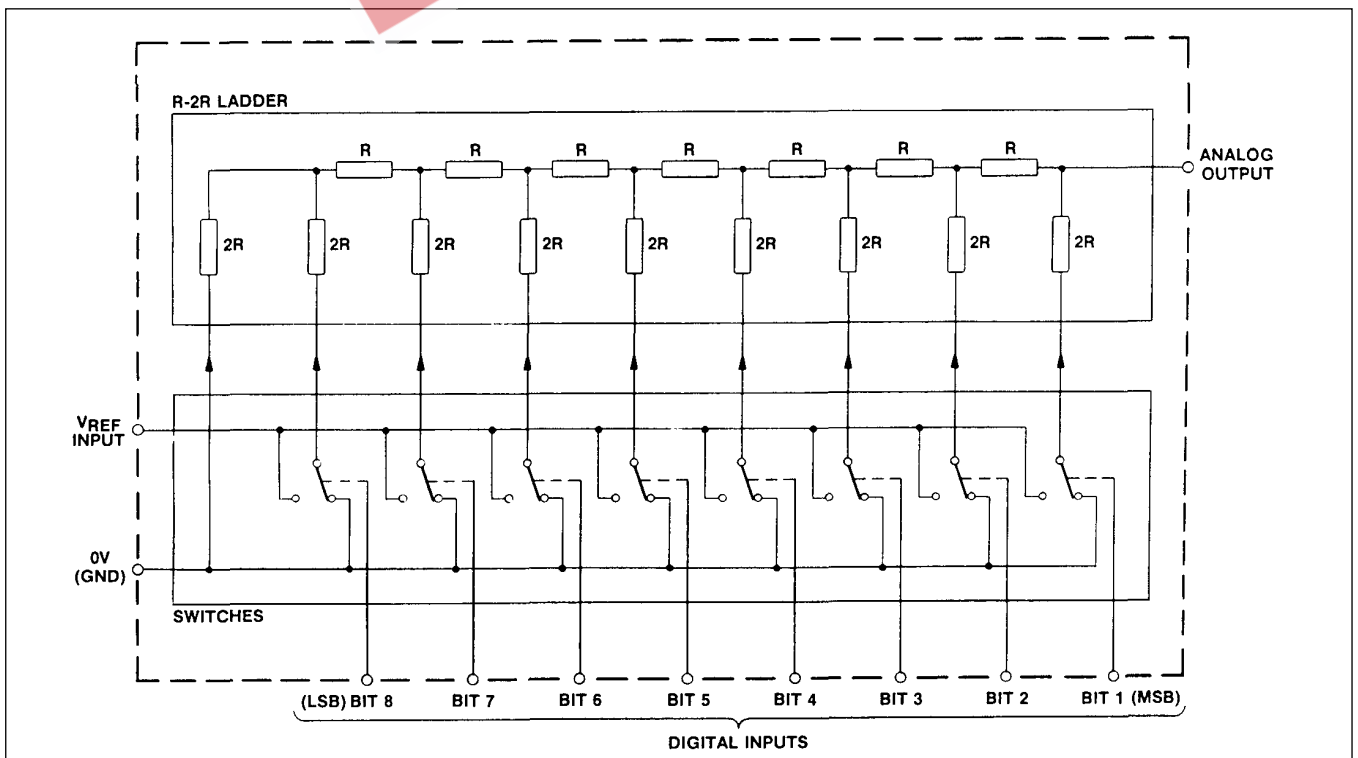


Fig.1 Pin connections (not to scale) - top view



## ZN429

### ELECTRICAL CHARACTERISTICS

(at  $T_{amb} = 25^{\circ}\text{C}$  and  $V_{CC} = +5\text{V}$  unless otherwise specified)

Parameter	Symbol	Min.	Typ.	Max.	Units	Conditions
<b>Converter Resolution</b>		8	-	-	bits	
Accuracy		8	-	-	bits	
Non-linearity		-	-	$\pm 0.5$	LSB	
Differential non-linearity		-	$\pm 0.5$	-	LSB	Note 1
Settling time to 0.5LSB		-	1.0	-	$\mu\text{s}$	1 LSB step
Settling time to 0.5LSB		-	2.0	-	$\mu\text{s}$	All bits ON to OFF or OFF to ON
Offset voltage ZN429E8, ZN429D	$V_{OS}$	-	3.0	5.0	mV	All bits OFF
$V_{OS}$ temperature coefficient		-	5	-	$\mu\text{V}/^{\circ}\text{C}$	
Full-scale output		2.545	2.550	2.555	V	All bits ON Ext. $V_{REF} = 2.56\text{V}$
Full-scale temp. coefficient		-	3	-	ppm/ $^{\circ}\text{C}$	Ext. $V_{REF} = 2.560\text{V}$
Non-linearity temp. coefficient		-	7.5	-	ppm/ $^{\circ}\text{C}$	Relative to F.S.R.
Analog output resistance	$R_O$	-	10	-	$\text{k}\Omega$	
External reference voltage		0	-	3.0	V	
Supply voltage	$V_{CC}$	4.5	-	5.5	V	
Supply current	$I_S$	-	5	9	mA	
High level input voltage	$V_{IH}$	2.0	-	-	V	
Low level input voltage	$V_{IL}$	-	-	0.7	V	
High level input current	$I_{IH}$	-	-	10	$\mu\text{A}$	$V_{CC} = \text{max.}$ $V_I = 2.4\text{V}$
				100	$\mu\text{A}$	$V_{CC} = \text{max.}$ $V_I = 5.5\text{V}$
Low level input current	$I_{IL}$	-	-	-0.18	mA	$V_{CC} = \text{max.}$ $V_I = 0.3\text{V}$

NOTE 1: Monotonic over full temperature range.

### INTRODUCTION

The ZN429 is an 8-bit D-A converter. It contains an advanced design of R-2R ladder network and an array of precision bipolar switches on a single monolithic chip.

The special design of the ladder network results in full 8-bit accuracy using normal diffused resistors.

The converter is of the voltage switching type and uses an R-2R resistor ladder network as shown in Fig.3.

Each 2R element is connected either to 0V or  $V_{REF}$  by transistor switches specially designed for low offset voltage (typically 1mV).

Binary weighted voltages are produced at the output of the R-2R ladder, the value depending on the digital number applied to the bit inputs.

An external fixed or varying reference is required which should have a slope resistance less than  $2\Omega$ .

Suggested external reference sources are the ZN404 or one of the ZN458 range. Each ZN404 is capable of supplying up to five ZN429 circuits and this is increased to ten for the ZN458 range.

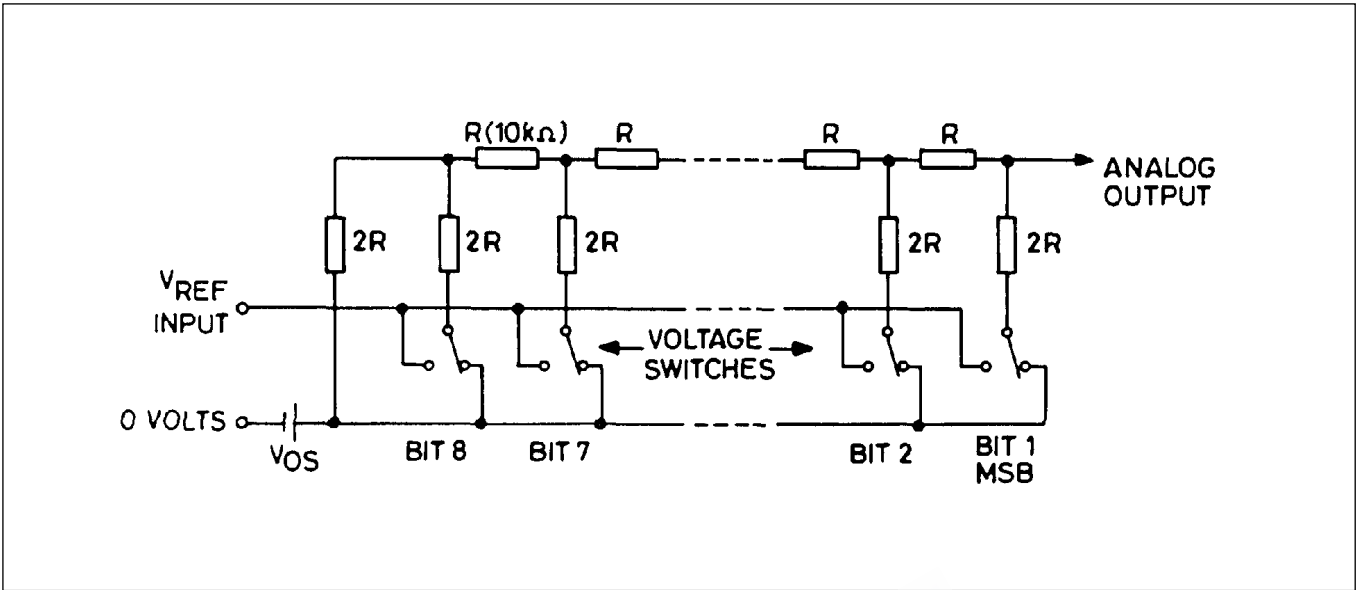


Fig.3 The R-2R ladder network

**APPLICATIONS**

**(1) Unipolar D-A Converter**

The nominal output range of the ZN429 is 0 to  $V_{REF IN}$  through a  $10\Omega$  resistance. Other output ranges can readily be obtained by using an external amplifier.

The resulting full-scale range is given by

$$V_{OUT FS} = \left(1 + \frac{R_1}{R_2}\right) V_{REF IN} = G \cdot V_{REF IN}$$

The impedance at the inverting input is  $R_1/R_2$  and for low drift with temperature this parallel combination should be equal to the ladder resistance ( $10k\Omega$ ). The required nominal values of  $R_1$  and  $R_2$  are given by

$$R_1 = 10Gk\Omega \text{ and } R_2 = 10G/(G-1)k\Omega.$$

Using these relationships a table of nominal resistance values for  $R_1$  and  $R_2$  can be constructed for  $V_{REF IN} = 2.5V$ .

Output Range	G	$R_1$	$R_2$
+5V	2	20k $\Omega$	20k $\Omega$
+10V	4	40k $\Omega$	13.33k $\Omega$

For gain setting  $R_1$  is adjusted about its nominal value. Practical circuit realisations (including amplifier stabilising components) for +5 and +10V output ranges are given in Fig.5. Settling time for a major transition is 2.5 $\mu s$  typical.

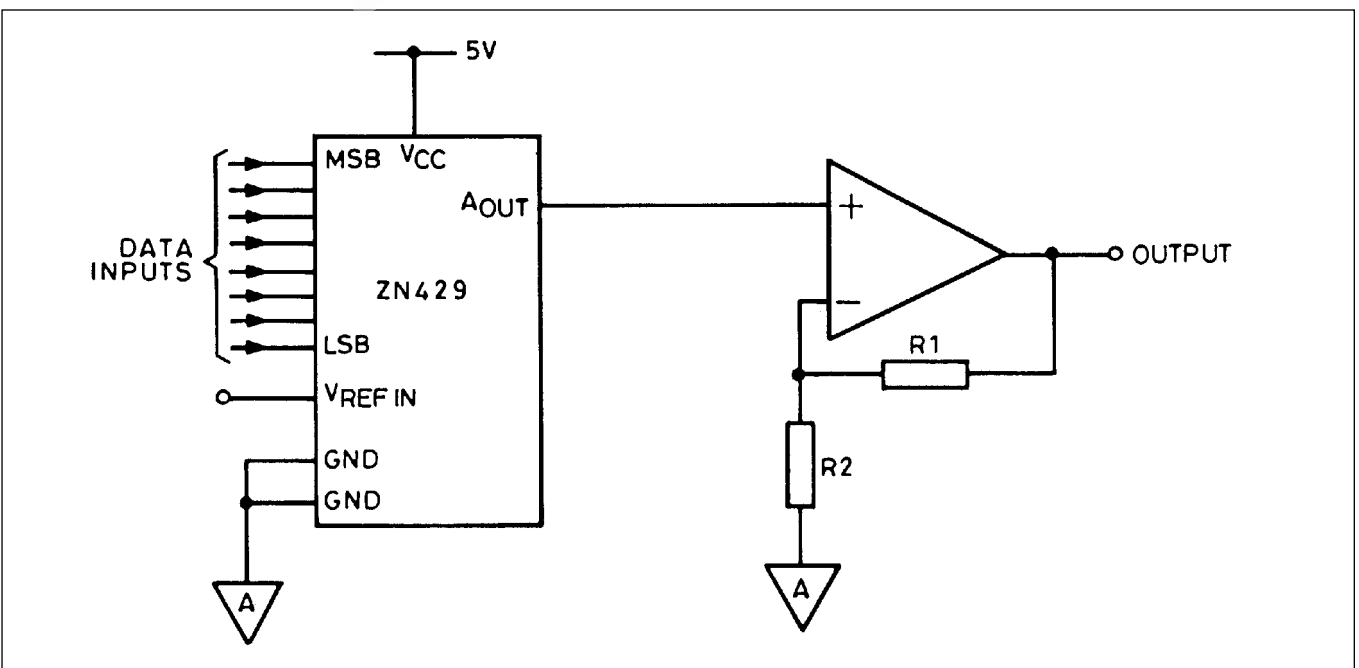


Fig.4 Unipolar operation - basic circuit

# ZN429

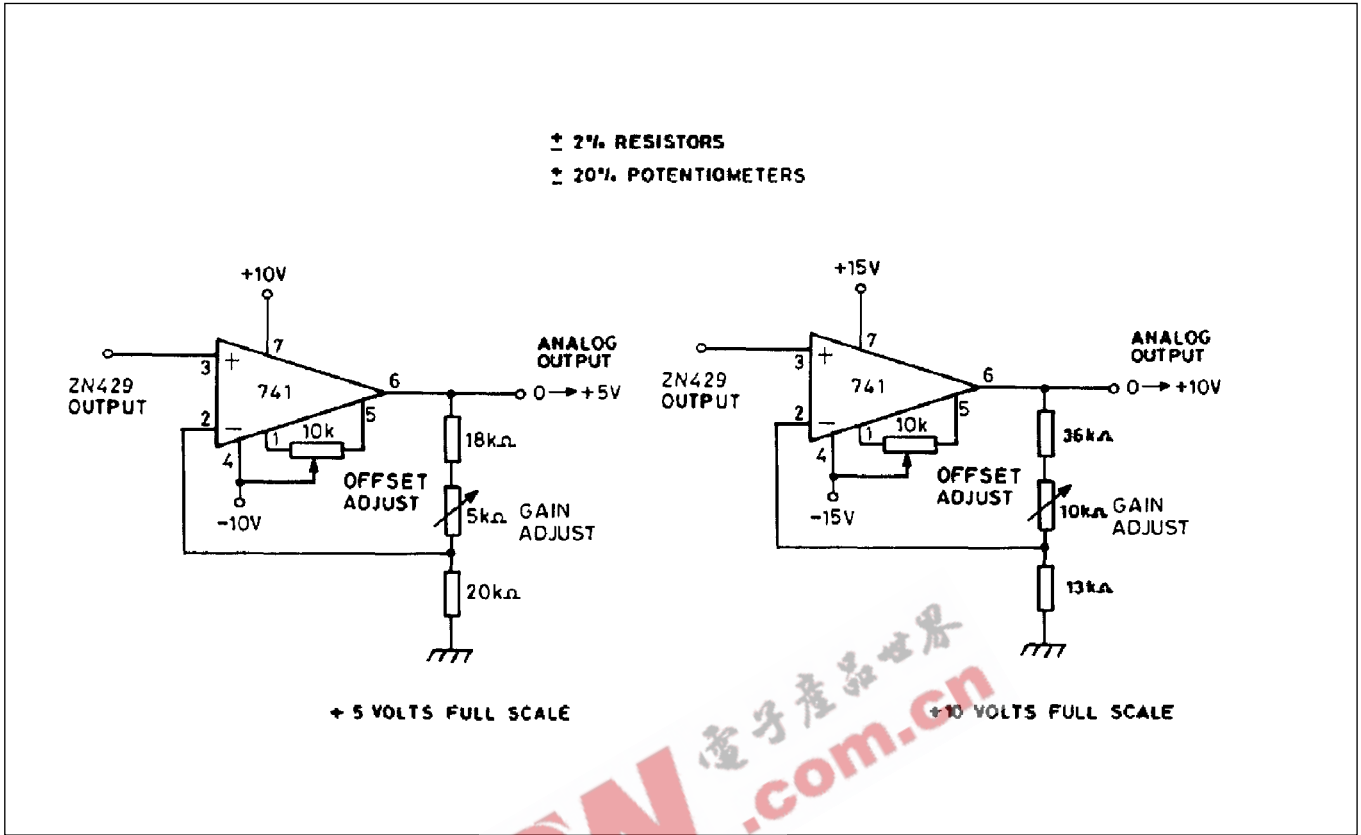


Fig.5 Unipolar operation - component values

## UNIPOLAR ADJUSTMENT PROCEDURE

- (i) Set all bits to OFF (LOW) and adjust zero until  $V_{OUT} = 0.0000V$ .
- (ii) Set all bits ON (HIGH) and adjust gain until  $V_{OUT} = FS - 1LSB$ .

## UNIPOLAR SETTING UP POINTS

Output Range, +FS	LSB	FS - 1LSB
+5V	19.5 mV	4.9805V
+10V	39.1mV	9.9609V

$$1LSB = \frac{FS}{256}$$

## UNIPOLAR LOGIC CODING

Input Code (Binary)	Analog Output (Nominal Value)
11111111	FS - 1LSB
11111110	FS - 2 LSB
11000000	$\frac{3}{4}$ FS
10000001	$\frac{1}{2}$ FS + 1LSB
10000000	$\frac{1}{2}$ FS
01111111	$\frac{1}{2}$ FS - 1LSB
01000000	$\frac{1}{4}$ FS
00000001	1LSB
00000000	0

## (2) Bipolar D-A Converter

For bipolar operation the output from the ZN429 is offset by half full-scale by connecting a resistor  $R_3$  between  $V_{REF IN}$  and the inverting input of the buffer amplifier (Fig.6).

When the digital input of the ZN429 is zero the analog output is zero and the amplifier output should be -full-scale. An input of all ones to the D-A will give a ZN429 output of  $\pm V_{REF IN}$  and the amplifier output required is +full-scale. Also, to match the ladder resistance the parallel combination of  $R_1$ ,  $R_2$  and  $R_3$  should be  $10k\Omega$ .

The nominal values of  $R_1$ ,  $R_2$  and  $R_3$  which meet these conditions are given by

$$R_1 = 20Gk\Omega, R_2 = 20G/(G-1)k\Omega \text{ and } R_3 = 20k\Omega.$$

where the resultant output range is  $\pm G \cdot V_{REF IN}$ .

Assuming that  $V_{REF IN} = 2.5V$  the nominal values of resistors for  $\pm 5$  and  $\pm 10V$  output ranges are given in the following table:

Output Range	G	$R_1$	$R_2$	$R_3$
$\pm 5V$	2	$40k\Omega$	$40k\Omega$	$20k\Omega$
$\pm 10V$	4	$80k\Omega$	$26.67k\Omega$	$20k\Omega$

Minus full scale (OFFSET) is set by adjusting  $R_1$  about its nominal value relative to  $R_3$ . Plus full-scale (GAIN) is set by adjusting  $R_2$  relative to  $R_1$ .

Settling time for a major transition is  $2.5\mu s$  typical.

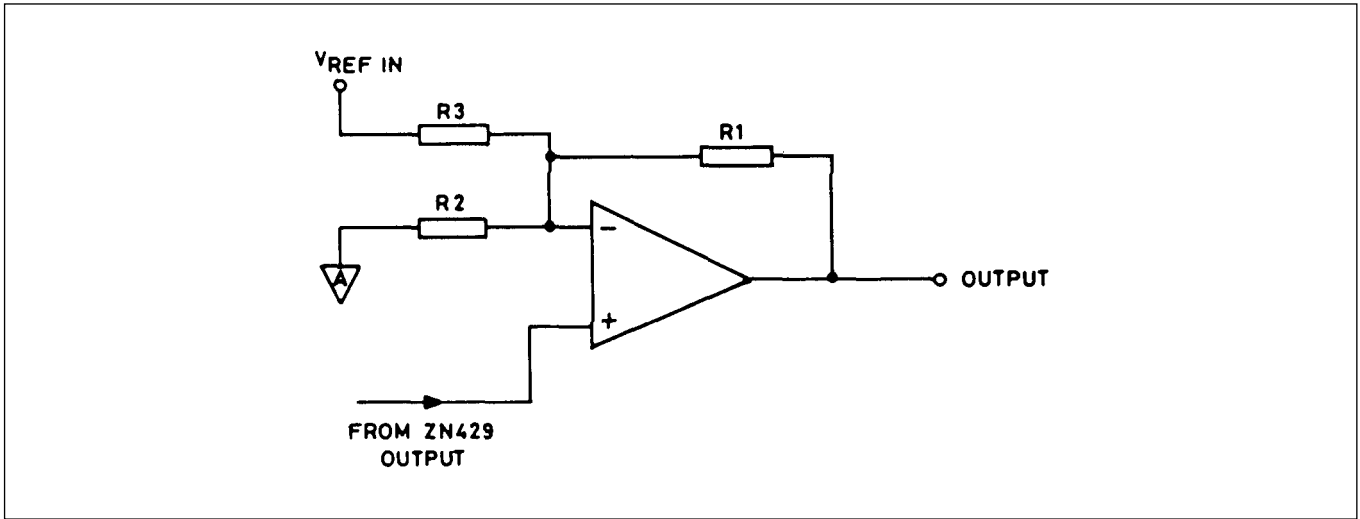


Fig.6 Bipolar operation - basic circuit

**BIPOLAR ADJUSTMENT PROCEDURE**

- (i) Set all bits to OFF (LOW) and adjust OFFSET until the amplifier output reads -FULL-SCALE.
- (ii) Set all bits ON (HIGH) and adjust gain until the amplifier reads +(FULL-SCALE - 1LSB).

**BIPOLAR LOGIC CODING**

Input Code (Offset Binary)	Analog Output (Nominal Value)
11111111	+(FS - 1LSB)
11111110	+(FS - 2 LSB)
11000000	+ <sup>1</sup> / <sub>2</sub> FS
10000001	+ 1LSB
10000000	0
01111111	-1 LSB
01000000	- <sup>1</sup> / <sub>2</sub> FS
00000001	-(FS - 1LSB)
00000000	-FS

**BIPOLAR SETTING UP POINTS**

Input Range, ± FS	LSB	-FS	+(FS - 1LSB)
±5V	39.1 mV	-5.0000V	+4.9609V
±10V	78.1mV	-10.0000V	9.9219V

1LSB =  $\frac{2FS}{256}$

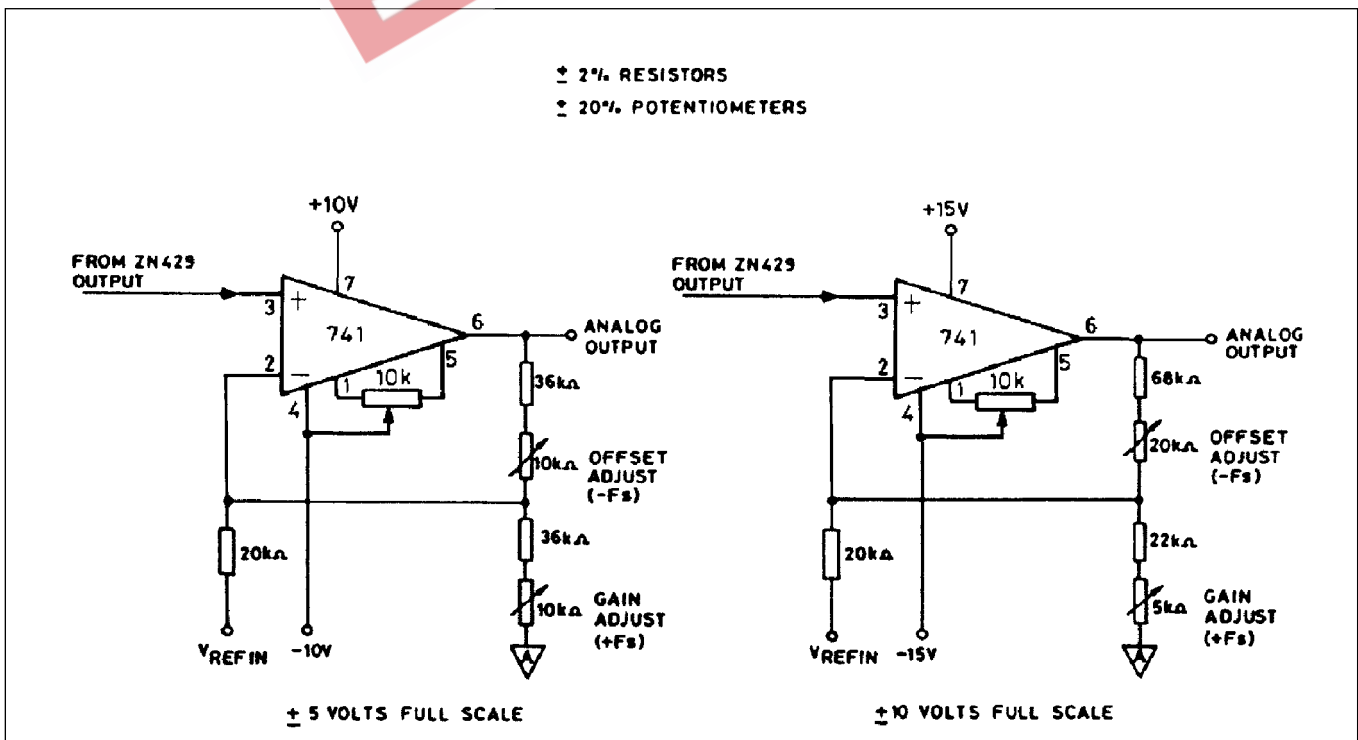


Fig.7 Bipolar operation - component values

ZN429

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HEADQUARTERS OPERATIONS

**GEC PLESSEY SEMICONDUCTORS**

Cheney Manor, Swindon,  
Wiltshire, United Kingdom. SN2 2QW  
Tel: (01793) 518000  
Fax: (01793) 518411

**GEC PLESSEY SEMICONDUCTORS**

P.O. Box 660017,  
1500 Green Hills Road,  
Scotts Valley, California 95067-0017,  
United States of America.  
Tel (408) 438 2900  
Fax: (408) 438 5576

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