

CXA1946BQ/BR

Electronic Volume

Description

The CXA1946BQ/BR is a serial control electronic volume IC designed for use in audio systems.

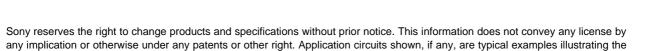
Features

- Loudness
- Volume control (0dB to -87dB in 1dB step, -∞dB)
- Balance
- Tone control (15 steps, 2 bands, -16dB to +16dB)
- Fader (2dB-step to -20dB, -25dB, -35dB, -45dB, -60dB, -∞dB)
- Input selector (4 channels)
- Gain can be set for each input channel (common for channels 3 and 4)
- Serial data control (DATA, CLK, CE)
- Single 8V power supply

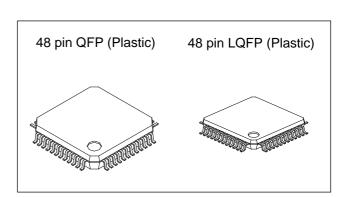
- Absolute Maximum Ratings
 Supply voltage
 Operation
- °C Operating temperature Topr -40 to +85
- Storage temperature -65 to +150 Tstg
- Allowable power dissipation PD 350 mW (Ta = 85° C)

Operating Conditions

Supply voltage Vcc 6 to 12 V

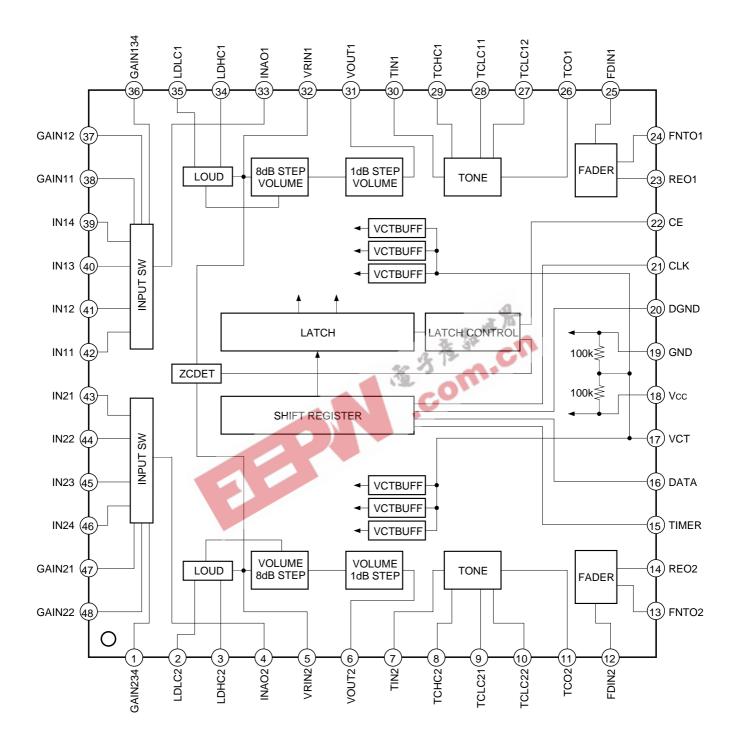


°C



operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits.

Block Diagram and Pin Configuration



Pin Description

Pin No.	Symbol	I/O resistance Pin voltage	Equivalent circuit	Description
1 36	GAIN234 GAIN134	≃∞ VCT	Vcc 1 30 GND	Sets gain for IN3 and IN4.
2 35	LDLC2 LDLC1	6.18kΩ VCT	2 33 CRND	Sets loudness low cut-off frequency.
3 34	LDHC2 LDHC1	8.92kΩ VCT	Vcc	Sets loudness high cut-off frequency.
4 33	INAO2 INAO1	VCT	4 (4) (3) (3) (3) (3) (3) (5) (5) (5) (5) (5) (5) (5) (5	Input selector output

Pin No.	Symbol	I/O resistance Pin voltage	Equivalent circuit	Description
5 32	VRIN2 VRIN1	9.5kΩ VCT	5 32 GND	Volume input
6 31	VOUT2 VOUT1	VCT	6 31 GND	Volume output
7 30	TIN2 TIN1	19kΩ VCT	Vcc 7 30 S GND	Tone input
8 29	TCHC2 TCHC1	5kΩ VCT	€	Sets tone high frequency.

Pin No.	Symbol	I/O resistance Pin voltage	Equivalent circuit	Description
9 28	TCLC21 TCLC11	8kΩ VCT	9 (9) (28) (28) (3) (4) (5) (5) (5) (5) (5) (5) (5) (5) (5) (5	Sets tone low frequency.
10 27	TCLC22 TCLC12	8kΩ VCT	10 Vcc Vcc (10 Vcc (10) (27) (10) (27) (10) (27) (10) (27) (10) (27) (10) (27) (10) (27) (10) (27)	Sets tone low frequency.
11 26	TCO2 TCO1	VCT	Vcc 11 20 GND	Tone control output
12 25	FDIN2 FDIN1	24kΩ VCT	12 (25) GND	Fader input

Pin No.	Symbol	I/O resistance Pin voltage	Equivalent circuit	Description
13 24	FNTO2 FNTO1	VCT	Vcc (13) (24) (24) (3) (3) (3) (4) (5) (5) (5) (5) (5) (5) (5) (5	Front output
14 23	REO2 REO1	VCT	Vcc (14) (23) GND	Rear output
15	TIMER		Vcc 15 GND	Sets timer.
16	DATA	~ ~	TIG TIG TIG TIG TIG TIG TIG TIG	Serial data input

SONY

Pin No.	Symbol	I/O resistance Pin voltage	Equivalent circuit	Description
17	VCT	VCT		Center electric potential
18	Vcc	Vcc		+ power supply
19	GND	GND		GND
20	DGND	_		Digital GND
21	CLK	≃∞ _	21 W GND	Serial clock input
22	CE	≃∞ _	22 With GND	Latch enable input
37 48	GAIN12 GAIN22	≃∞ VCT	37 48 GND	Sets gain for IN2.

Pin No.	Symbol	I/O resistance Pin voltage	Equivalent circuit	Description
38 47	GAIN11 GAIN21	≈ ∞	38 47 GND	Sets gain for IN1.
39 40 41 42 43 44 45 46	IN14 IN13 IN12 IN11 IN21 IN22 IN23 IN24	50kΩ VCT	39 (43) (40) (44) (41) (45) (42) (46) GND	Signal input
		F		

Data Format

(a) Data allocation

FAST BIT	D1 D2	NOP	MSB
	D3 D4	ISW	
	D5	LOUD	
	D6 D7 D8 D9	VRC1	
	D10 D11 D12	VRF1	
	D13 D14 D15 D16	VRC2	
	D17 D18 D19	VRF2	· Com. cn
	D20 D21 D22 D23	TONE BASS	.com
	D24 D25 D26 D27	TONE TREBLE	
	D28 D29 D30 D31	FADER	
LAST BIT	D32	FADER SELECT	LSB

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(b) Setting table

• NOP

Setting value	D1	D2
_	0	0

• ISW

Setting value	D3	D4
IN14/IN24	1	1
IN13/IN23	1	0
IN12/IN22	0	1
IN11/IN21	0	0

• LOUD

Setting value	D5
ON	1
OFF	0

• VRC1/VRC2

• VRC1/VRC2					- 8-
Setting value	D6/D13	D7/D14	D8/D15	D9/D16	34 34 1
0	1	1	1	1 4	B G
-8	1	1	1	0	
-16	1	1	0	1	.0`
-24	1	1	0	0	
-32	1	0	1	1	
-40	1	0	(1)	0	
-48	1	0	0	1	
-56	1	0	0	0	
-64	0	1	1	1	
-72	0	1	1	0	
-80	0	1	0	1	
	0	1	0	0	
	0	0	0	0	

• VRF1/VRF2

Setting value	D10/D17	D11/D18	D12/D19
0	1	1	1
-1	1	1	0
-2	1	0	1
-2 -3 -4 -5	1	0	0
-4	0	1	1
-5	0	1	0
-6	0	0	1
-7	0	0	0

• TONE BASS/TREBLE

Setting value	D20/D24	D21/D25	D22/D26
14	1	1	1
12	1	1	0
10	1	0	1
8	1	0	0
6	0	1	1
4	0	1	0
2	0	0	1
0	0	0	0

• BOOST/CUT

Setting value	D23/D27
BOOST	1
CUT	0

• FADER

Setting value	D28	D29	D30	D31	
	1	1	1	1	
-60	1	1	1	0	0
-45	1	1	0	1	
-35	1	1	0	0	A CN
-25	1	0	1	1	A ar
-20	1	0	1	0	5
–18	1	0	0	1 1	an' a
–16	1	0	0	0	
-14	0	1	1	1	
–12	0	1	1	0	
-10	0	1	0	1	
8 6	0	1	0	0	
-6	0	0	1	1	
-4	0	0	1	0	
-2	0	0	0	1	
0	0	0	0	0	

• FADER SELECT

Setting value	D32
Attenuation of front signal	1
Attenuation of rear signal	0

• RESET

Reset is performed automatically when power is first supplied to the IC; there is no reset pin.

The following table shows the respective statuses of various settings after a reset has been performed. However, from the time when power is first supplied until the first data transfer, keep CE high by pulling it up to Vcc, etc.

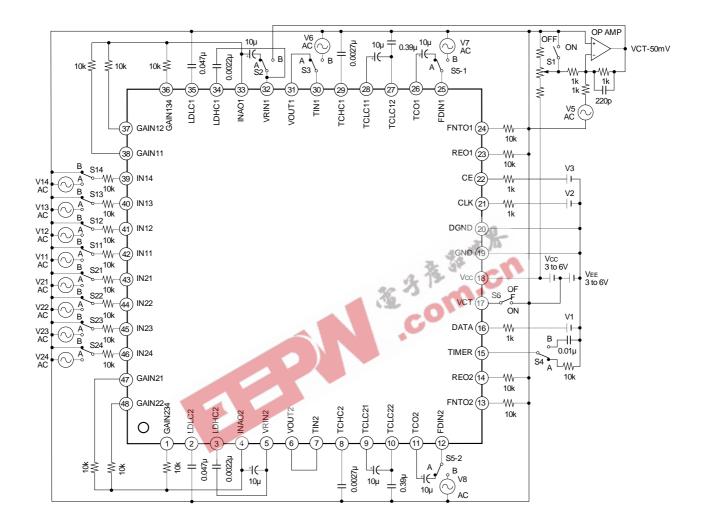
MODE	Setting value
INPUT	1
VRC1	
VRF1	–7dB
VRC2	- ∞
VRF2	–7dB
LOUD	OFF
TONE BASS	0dB
TONE TREBLE	0dB
FADER	0dB, REAR

					,		
Item		Symbol	Measurement conditions		Тур.	Max.	Unit
Current consumption Icc		lcc	No signal		20	25	mA
Total harmonic distor	tion	THD	1kHz, 5dBm output	_	0.005	0.01	%
Output noise voltage		Vn	input shorted, A weight	_	7	10	µVrms
Maximum output volta	age	Vom	1kHz		—		dBm
Separation		CS	1kHz		90		dB
Volume maximum att	enuation	ATTm	1kHz	85	90	_	dB
Loudness	Low	Glb	100Hz, VRC = -16dB	7	8	9	dB
Loudiless	High	Glh	10kHz, VRC = -16dB	5	6	7	dB
Bass max. boost gair	n	Gbb		14	16	18	dB
Bass max. cut gain		Gbc		14	16	18	dB
Treble max. boost ga	ainn	Gtb		14	16	18	dB
Treble max. cut gain	Treble max. cut gain Gto			14	16	18	dB
Low		Vsl		0	—	1.5	V
Input voltage	High	Vsh	h DATA, CLK, CE		—	6	V
Input voltage range Vin		IN11 to 14, IN21 to 24, VRIN1, VRIN2, TIN1, TIN2, FDIN1, FDIN2	1	_	Vcc–1	V	

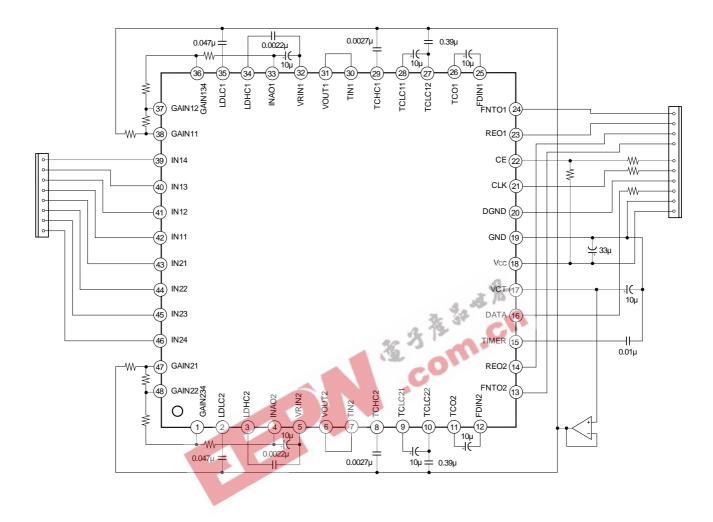
Electrical Characteristics

(Unless otherwise specified, Vcc = 8V, $Ta = 25^{\circ}C$)



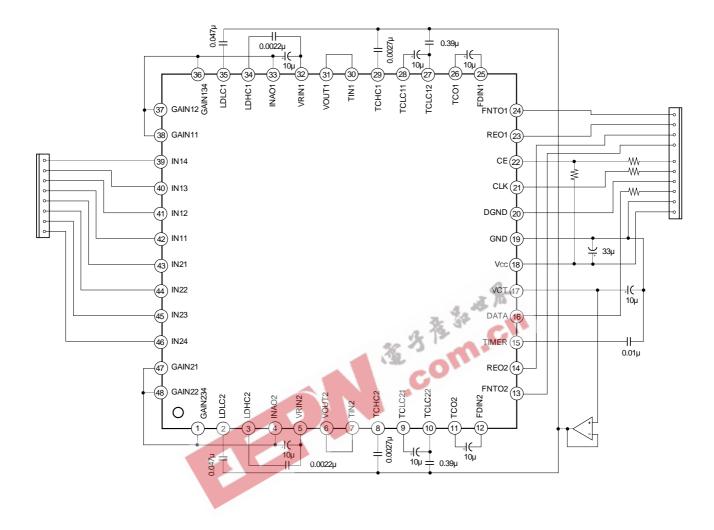


Application Circuit 1

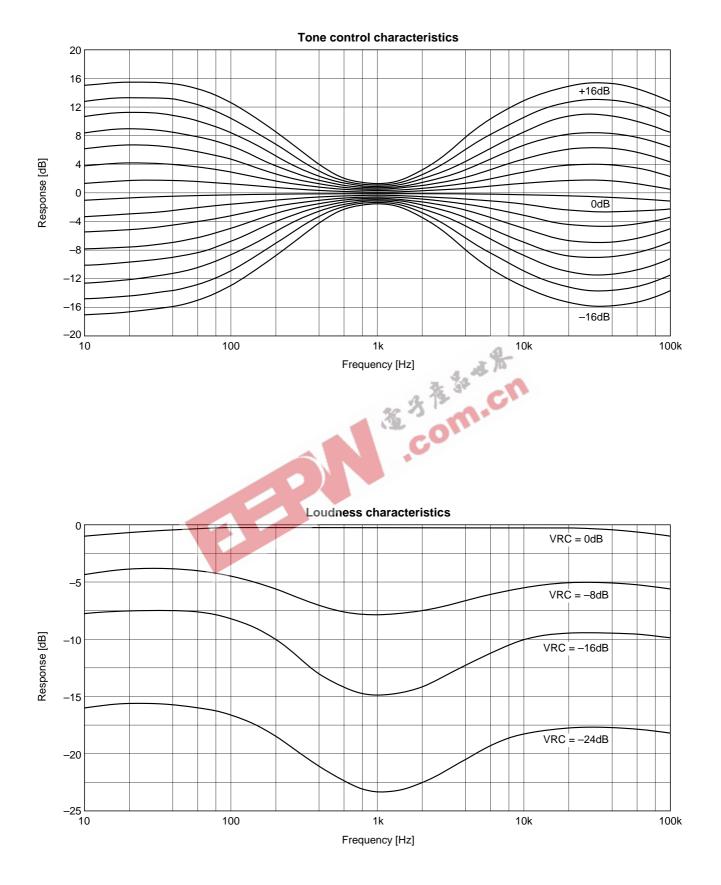


Application circuits shown are typical examples illustrating the operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits or for any infringement of third party patent and other right due to same.

Application Circuit 2



Application circuits shown are typical examples illustrating the operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits or for any infringement of third party patent and other right due to same.



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Description of Operation

The CXA1946BQ/BR is a serial control electronic volume IC designed for use in audio systems. The internal circuit of the IC consists of the following blocks:

- 1. Input selector
- 2. Volume
- 3. Loudness
- 4. Tone control
- 5. Fader
- 6. VCT buffer
- 7. Serial data I/O
- 8. Zero-cross detector (with timer)
- 9. Power-off mute

The operation of each block and notes on their use are described below.

Note that when the circuits for channels 1 and 2 are identical, the suffix "X" is added to pin names and device names in order to distinguish between the two channels.

1. Input selector

There are two channels (stereo), each with four systems of input pins; the input selector selects one of those input systems.

The gain between the input pins and the output pin of the input selector can be set independently for each input system, except the gain for inputs 3 and 4 is common.

Determine the gain for each system through the settings of the feedback circuit constants as shown in Figs. 1 and 2. When each input gain is set to \times 1, short INAOX and GAIN \times 1,GAIN \times 2,GAIN \times 34.

The input impedance is $50k\Omega$ (typ.) for each input.

The output impedance for INAO1 and INAO2 is low impedance (roughly 0Ω). The gain is not affected by the load impedance.

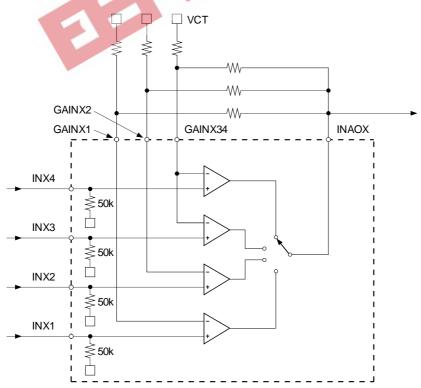
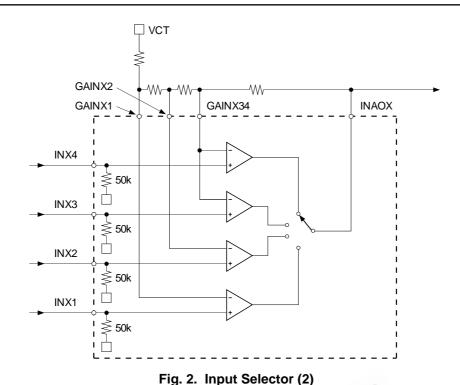


Fig. 1. Input Selector (1)



2. Volume

The volume circuit consists of two sections, an 8dB/step section and a 1dB/step section, as shown in Fig. 3. This circuit also serves as a balance control because the volume for channel 1 and channel 2 can be set independently.

To mute the output signal, send $-\infty dB$ data.

The input impedance is $9.5k\Omega$ (typ.) for VRIN1 and VRIN2.

The output impedance for VOUT1 and VOUT2 is low impedance (roughly 0Ω). The volume step width and gain are not affected by the load impedance.

3. Loudness

The configuration of the loudness circuit is shown in Fig. 3. CLDHCX and CLDLCX are connected externally, and the loudness frequency characteristics are determined by these constants. The relationships between CLDHCX/CLDLCX and the frequency characteristics are as follows:

 $1/fL = 2\pi CLDLCXR_1$

 $1/f_{H} = 2\pi C_{LDHCX}R_{2}$

The loudness characteristics are not affected by the load impedance of VOUT1 and VOUT2. Loudness is turned on and off by serial data bit D5.

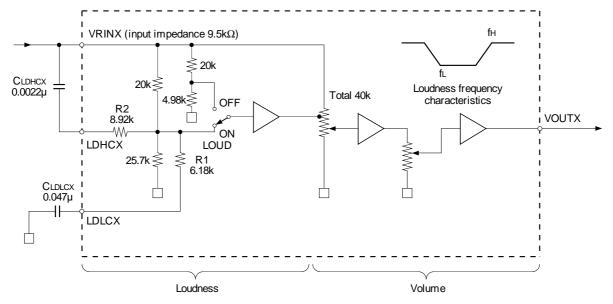


Fig. 3. Volume and Loudness

4. Tone control

The configuration of the tone control circuit is shown in Fig. 4. CTCLCX2 and CTCHCX are connected externally, and the tone control frequency characteristics can be changed by changing these constants. The relationships between CTCLCx2/CTCHCX and the frequency characteristics are as follows:

 $1/f_L = 2\pi C_{TCLCX2} (R_3//R_4)$

 $1/f_{H} = 2\pi C_{TCHC} R_{5}$

The maximum bass boost and cut can be made smaller than in the Application Circuit by connecting an external resistance to the TCLCX1 pin in series, or else connecting an external resistance to CTCLCX2 in parallel. (See Fig. 5.) Furthermore, the maximum treble boost and cut can be made smaller than in the Application Circuit by connecting an external resistance to CTCHCX in series. (See Fig. 6.) However, when these methods are used, variations in the absolute value of the CXA1946B internal resistance (±20% max.) and in the external resistance will cause variations in the tone control characteristics. Set these constants after studying all considerations carefully. Note that when the method illustrated in the Application Circuit is used, variations in the internal resistance of the CXA1946B have no effect on the tone control characteristics.

The input impedance is $19k\Omega$ (typ.) for TIN1 and TIN2.

The output impedance for TCO1 and TCO2 is low impedance (roughly 0Ω). The tone step width and gain are not affected by the load impedance.

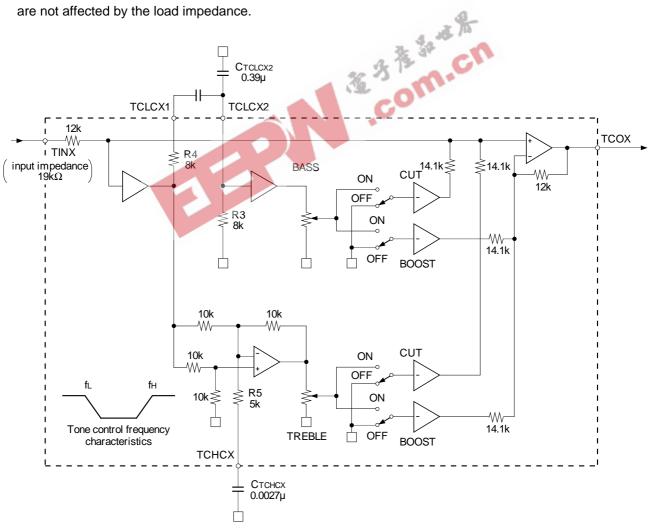
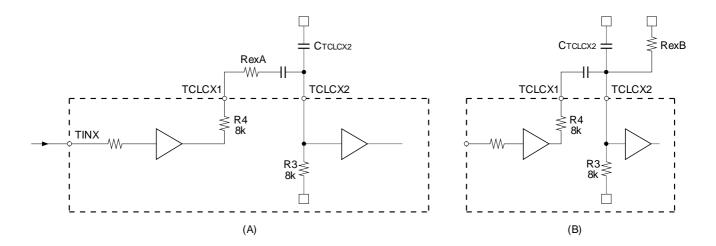
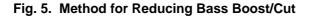
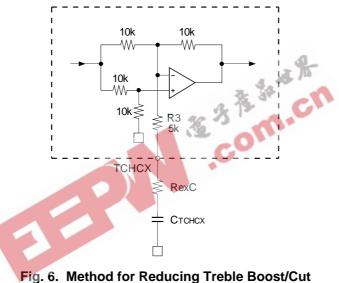


Fig. 4. Tone Control







5. Fader

The configuration of the fader circuit is shown in Fig. 7. The fader operates by specifying the amount of attenuation for either the front or rear output signal and by specifying which output signal (front or rear) is to be attenuated.

The input impedance is $24k\Omega$ (typ.) for FDIN1 and FDIN2.

The output impedance for FNTO1, FNTO2, REO1, and REO2 is low impedance (roughly 0Ω). The gain and fader step width are not affected by the load impedance.

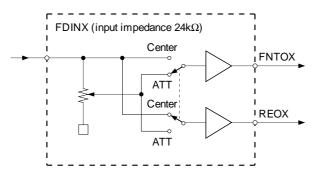
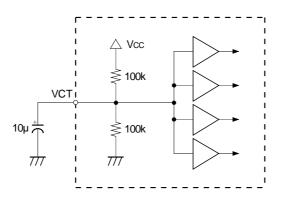


Fig. 7. Fader

6. VCT buffer

The internal circuit for the VCT pin is shown in Fig. 8.

This circuit generates the electric potential for the center between Vcc and GND (Vcc/2). The IC internal operation reference potential is equal to the output potential of VCT buffer. The impedance for the VCT pin (Pin 17) is high since it is connected to a bypass capacitor. Add an external buffer when using the electric potential of the VCT pin as the external reference potential for the CXA1946.





7. Serial data I/O

The serial data has a 32-bit structure as indicated in the specifications. Data input is conducted using three inputs: DATA, CLK, and CE. DATA is shifted in the CXA1946B internal shift register at the rising edge of CLK. The data in the shift register is latched at the falling edge of CE. Refer to this specification for details on the timing.

The CXA1946B does not have a reset (initialize) pin. The internal shift register and latch are reset automatically when power is first supplied to the IC. To execute a reset at other times, send the data (statuses after reset) shown in the item "RESET" of this specification to the CXA1946B.

8. Zero-cross detector (with timer)

Using the zero-cross detector, the internal latch data is overwritten the first time the input signal becomes roughly 0 after serial data is sent (after CE goes low). This operation reduces noise when overwriting data. Although there are usually no problems when a normal audio signal is input, in rare cases there may be nothing except a large-amplitude input signal of the high band, causing the slew rate to become abnormally high; the zero-cross detection signal is not output in such a case because the zero-cross detector response speed is too slow. Another rare situation would be that the zero-cross detection signal is output very infrequently because the input signal frequency is extremely low. In these types of instances, it is conceivable that the internal latch data will not be overwritten after data is sent, or that it will take much time until the data is overwritten. Therefore, to an external observer it will appear that the data is not being overwritten regardless of the fact that data is being sent.

As a countermeasure, the IC is designed to permit the internal latch data to be forcibly overwritten if the zero-cross detection signal is not output within a certain waiting period after the data is sent (after CE goes low). This function is called the "timer." If the zero-cross detection signal is output within a certain waiting period, the internal latch data is overwritten in synchronization with the zero-cross of the input signal.

The waiting period mentioned above can be changed according to the value of the external capacitor connected to the TIMER pin. When the value of the capacitor is 0.01μ F, the waiting period is approximately 500 μ s.

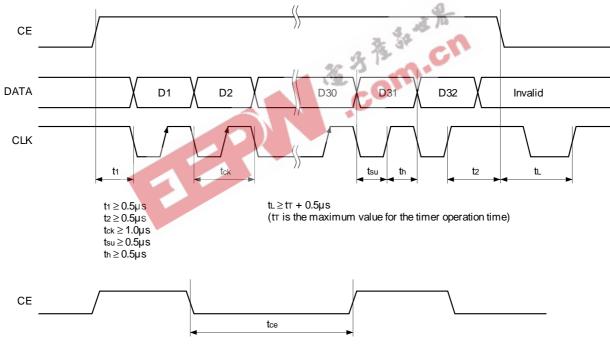
9. Power-off mute

When Vcc goes below 5V, the output stage bias of the fader output pins FNTO1, FNTO2, REO1, and REO2 is turned off and the pins go to high impedance. This operation prevents popping noises caused by the output pin potential deviating from Vcc/2 when the power is turned off.

Connections and Characteristics of Each Block

In the Application Circuit, the signal path goes from the input selector to the volume (+loudness) to the tone control to the fader. The sequence of the blocks in the signal path can be changed because the I/O pins for each block are independent of each other. For example, it is possible to switch the sequence of the volume circuit and the tone control so that the signal path goes from the input selector to the tone control to the volume (+loudness) to the fader. When this connection method is used, the noise voltage in the fader output can be reduced in actual use because the noise and signal up to the tone control are attenuated by the volume. However, because the maximum output amplitude of the tone control circuit is limited by the supply voltage, care should be given to the setting of the input signal level.

Although blocks in the Application Circuit are linked either by coupling capacitors or by direct connection, it is also possible to insert external circuits between blocks. In this case, the gain will change according to the input impedance of the following block and the impedance of the external circuit. In addition, the input impedance of each block can vary by $\pm 20\%$ due to the characteristics of the IC. Consequently, the overall gain also varies. Give careful consideration to the effects of this variation when setting the constants. The step widths (control characteristics) of the volume, tone control, and fader are not affected.



Timing Chart

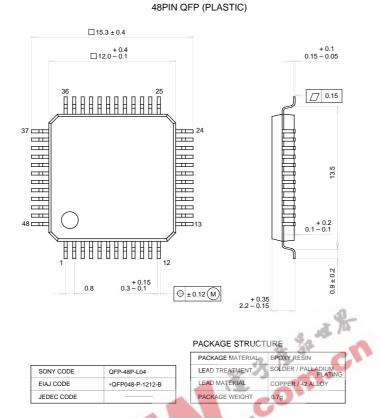
Timer Waiting Period Setting	g Chart (Vcc = 6 to 12V)	, operating temperature =	–35 to +85°C)
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TIMER pin	Waiting peroid				
capacitance C	Min.	Тур.	Max.		
C = 100pF	3µs	5µs	9µs		
C = 0.001µF	30µs	50µs	90µs		
C = 0.01µF	300µs	500µs	900µs		
C = 0.1µF	3ms	5ms	9ms		
C = 1µF	30ms	50ms	90ms		
C = 10µF	300ms	500ms	900ms		

Package Outline

Unit: mm

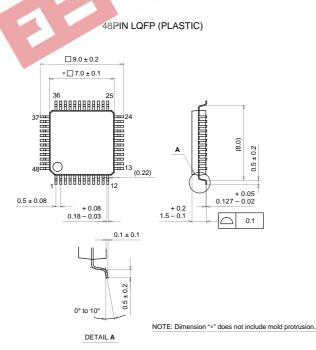
CXA1946BQ



NOTE : PALLADIUM PLATING

This product uses S-PdPPF (Sony Spec.-Palladium Pre-Plated Lead Frame).

CXA1946BR



PACKAGE	STRUCTURE

			PACKAGE MATERIAL	EPOXY RESIN
SONY CODE	LQFP-48P-L01		LEAD TREATMENT	SOLDER/PALLADIUM PLATING
EIAJ CODE	LQFP048-P-0707		LEAD MATERIAL	42/COPPER ALLOY
JEDEC CODE			PACKAGE MASS	0.2g