

MC34063A MC33063A

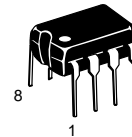
DC-to-DC Converter Control Circuits

The MC34063A Series is a monolithic control circuit containing the primary functions required for DC-to-DC converters. These devices consist of an internal temperature compensated reference, comparator, controlled duty cycle oscillator with an active current limit circuit, driver and high current output switch. This series was specifically designed to be incorporated in Step-Down and Step-Up and Voltage-Inverting applications with a minimum number of external components. Refer to Application Notes AN920A/D and AN954/D for additional design information.

- Operation from 3.0 V to 40 V Input
- Low Standby Current
- Current Limiting
- Output Switch Current to 1.5 A
- Output Voltage Adjustable
- Frequency Operation to 100 kHz
- Precision 2% Reference

DC-to-DC CONVERTER CONTROL CIRCUITS

SEMICONDUCTOR TECHNICAL DATA



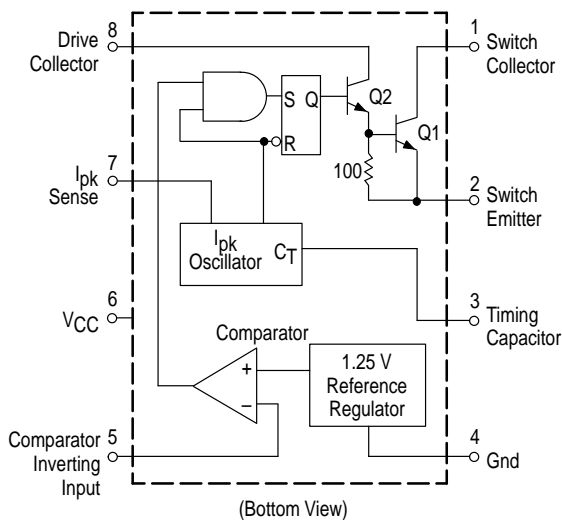
P, P1 SUFFIX
PLASTIC PACKAGE
CASE 626



D SUFFIX
PLASTIC PACKAGE
CASE 751
(SO-8)

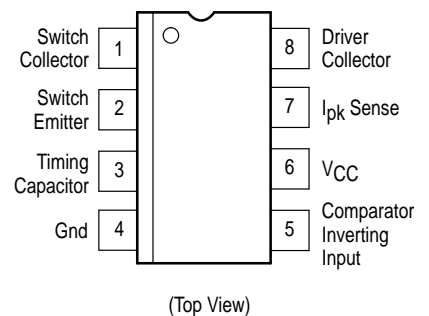
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Representative Schematic Diagram



This device contains 51 active transistors.

PIN CONNECTIONS



ORDERING INFORMATION

Device	Operating Temperature Range	Package
MC33063AD	$T_A = -40^\circ \text{ to } +85^\circ \text{C}$	SO-8
MC33063AP1		Plastic DIP
MC33063AVD	$T_A = -40^\circ \text{ to } +125^\circ \text{C}$	SO-8
MC33063AVP		Plastic DIP
MC34063AD	$T_A = 0^\circ \text{ to } +70^\circ \text{C}$	SO-8
MC34063AP1		Plastic DIP

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MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Power Supply Voltage	V_{CC}	40	Vdc
Comparator Input Voltage Range	V_{IR}	-0.3 to +40	Vdc
Switch Collector Voltage	$V_{C(\text{switch})}$	40	Vdc
Switch Emitter Voltage ($V_{P_{in\ 1}} = 40\text{ V}$)	$V_{E(\text{switch})}$	40	Vdc
Switch Collector to Emitter Voltage	$V_{CE(\text{switch})}$	40	Vdc
Driver Collector Voltage	$V_{C(\text{driver})}$	40	Vdc
Driver Collector Current (Note 1)	$I_{C(\text{driver})}$	100	mA
Switch Current	I_{SW}	1.5	A
Power Dissipation and Thermal Characteristics			
Plastic Package, P, P1 Suffix $T_A = 25^\circ\text{C}$	P_D	1.25	W
Thermal Resistance	$R_{\theta JA}$	100	$^\circ\text{C/W}$
SOIC Package, D Suffix $T_A = 25^\circ\text{C}$	P_D	625	W
Thermal Resistance	$R_{\theta JA}$	160	$^\circ\text{C/W}$
Operating Junction Temperature	T_J	+150	$^\circ\text{C}$
Operating Ambient Temperature Range			
MC34063A	T_A	0 to +70	$^\circ\text{C}$
MC33063AV		-40 to +125	
MC33063A		-40 to +85	
Storage Temperature Range	T_{stg}	-65 to +150	$^\circ\text{C}$

NOTES: 1. Maximum package power dissipation limits must be observed.
2. ESD data available upon request.

ELECTRICAL CHARACTERISTICS ($V_{CC} = 5.0\text{ V}$, $T_A = T_{low}$ to T_{high} [Note 3], unless otherwise specified.)

Characteristics	Symbol	Min	Typ	Max	Unit
OSCILLATOR					
Frequency ($V_{P_{in\ 5}} = 0\text{ V}$, $C_T = 1.0\text{ nF}$, $T_A = 25^\circ\text{C}$)	f_{osc}	24	33	42	kHz
Charge Current ($V_{CC} = 5.0\text{ V to }40\text{ V}$, $T_A = 25^\circ\text{C}$)	I_{chg}	24	35	42	μA
Discharge Current ($V_{CC} = 5.0\text{ V to }40\text{ V}$, $T_A = 25^\circ\text{C}$)	I_{dischg}	140	220	260	μA
Discharge to Charge Current Ratio (Pin 7 to V_{CC} , $T_A = 25^\circ\text{C}$)	I_{dischg}/I_{chg}	5.2	6.5	7.5	-
Current Limit Sense Voltage ($I_{chg} = I_{dischg}$, $T_A = 25^\circ\text{C}$)	$V_{ipk(\text{sense})}$	250	300	350	mV

OUTPUT SWITCH (Note 4)

Saturation Voltage, Darlington Connection (Note 5) ($I_{SW} = 1.0\text{ A}$, Pins 1, 8 connected)	$V_{CE(\text{sat})}$	-	1.0	1.3	V
Saturation Voltage, Darlington Connection ($I_{SW} = 1.0\text{ A}$, $R_{P_{in\ 8}} = 82\ \Omega$ to V_{CC} , Forced $\beta \approx 20$)	$V_{CE(\text{sat})}$	-	0.45	0.7	V
DC Current Gain ($I_{SW} = 1.0\text{ A}$, $V_{CE} = 5.0\text{ V}$, $T_A = 25^\circ\text{C}$)	h_{FE}	50	75	-	-
Collector Off-State Current ($V_{CE} = 40\text{ V}$)	$I_{C(\text{off})}$	-	0.01	100	μA

NOTES: 3. $T_{low} = 0^\circ\text{C}$ for MC34063A, -40°C for MC33063A, AV $T_{high} = +70^\circ\text{C}$ for MC34063A, $+85^\circ\text{C}$ for MC33063A, $+125^\circ\text{C}$ for MC33063AV
4. Low duty cycle pulse techniques are used during test to maintain junction temperature as close to ambient temperature as possible.
5. If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents ($\leq 300\text{ mA}$) and high driver currents ($\geq 30\text{ mA}$), it may take up to $2.0\ \mu\text{s}$ for it to come out of saturation. This condition will shorten the off time at frequencies $\geq 30\text{ kHz}$, and is magnified at high temperatures. This condition does not occur with a Darlington configuration, since the output switch cannot saturate. If a non-Darlington configuration is used, the following output drive condition is recommended:

$$\text{Forced } \beta \text{ of output switch : } \frac{I_{C \text{ output}}}{I_{C \text{ driver}} - 7.0\text{ mA}^*} \geq 10$$

*The $100\ \Omega$ resistor in the emitter of the driver device requires about 7.0 mA before the output switch conducts.

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ELECTRICAL CHARACTERISTICS (continued) ($V_{CC} = 5.0\text{ V}$, $T_A = T_{low}$ to T_{high} [Note 3], unless otherwise specified.)

Characteristics	Symbol	Min	Typ	Max	Unit
COMPARATOR					
Threshold Voltage $T_A = 25^\circ\text{C}$ $T_A = T_{low}$ to T_{high}	V_{th}	1.225 1.21	1.25 –	1.275 1.29	V
Threshold Voltage Line Regulation ($V_{CC} = 3.0\text{ V}$ to 40 V) MC33063A, MC34063A MC33363AV	Regline	– –	1.4 1.4	5.0 6.0	mV
Input Bias Current ($V_{in} = 0\text{ V}$)	I_{IB}	–	–20	–400	nA
TOTAL DEVICE					
Supply Current ($V_{CC} = 5.0\text{ V}$ to 40 V , $C_T = 1.0\text{ nF}$, Pin 7 = V_{CC} , $V_{Pin 5} > V_{th}$, Pin 2 = Gnd, remaining pins open)	I_{CC}	–	–	4.0	mA

- NOTES:** 3. $T_{low} = 0^\circ\text{C}$ for MC34063A, -40°C for MC33063A, AV $T_{high} = +70^\circ\text{C}$ for MC34063A, $+85^\circ\text{C}$ for MC33063A, $+125^\circ\text{C}$ for MC33063AV
 4. Low duty cycle pulse techniques are used during test to maintain junction temperature as close to ambient temperature as possible.
 5. If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents ($\leq 300\text{ mA}$) and high driver currents ($\geq 30\text{ mA}$), it may take up to $2.0\ \mu\text{s}$ for it to come out of saturation. This condition will shorten the off time at frequencies $\geq 30\text{ kHz}$, and is magnified at high temperatures. This condition does not occur with a Darlington configuration, since the output switch cannot saturate. If a non-Darlington configuration is used, the following output drive condition is recommended:

$$\text{Forced } \beta \text{ of output switch : } \frac{I_{C \text{ output}}}{I_{C \text{ driver}} - 7.0\text{ mA}^*} \geq 10$$

*The $100\ \Omega$ resistor in the emitter of the driver device requires about 7.0 mA before the output switch conducts.

Figure 1. Output Switch On–Off Time versus Oscillator Timing Capacitor

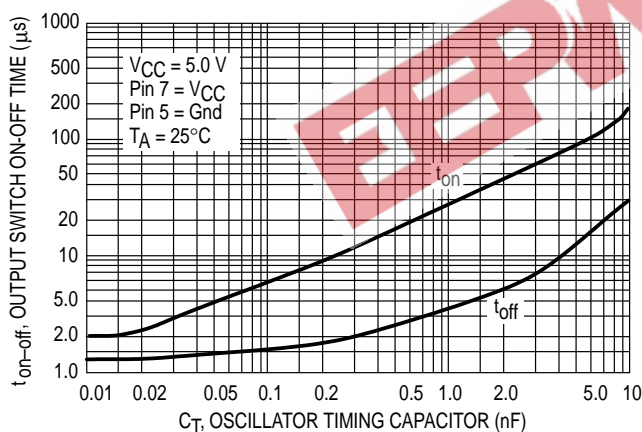
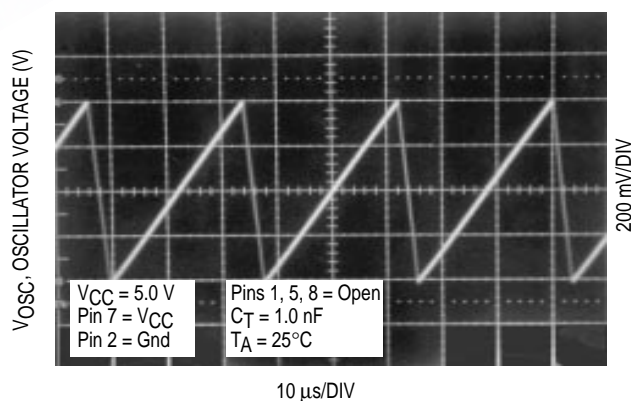


Figure 2. Timing Capacitor Waveform



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Figure 3. Emitter Follower Configuration Output Saturation Voltage versus Emitter Current

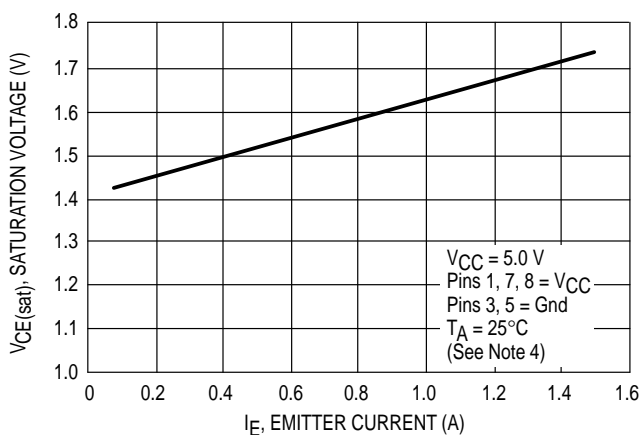


Figure 4. Common Emitter Configuration Output Switch Saturation Voltage versus Collector Current

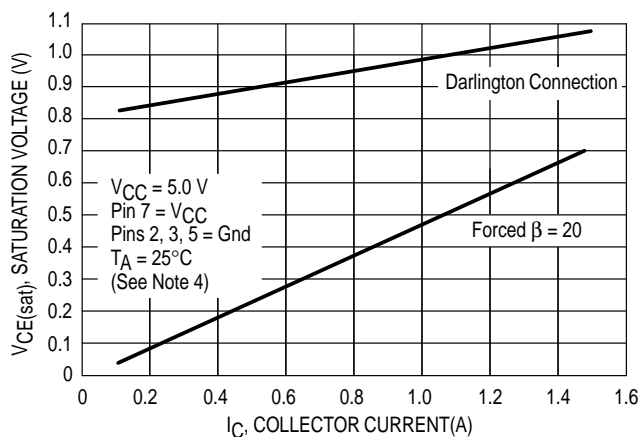


Figure 5. Current Limit Sense Voltage versus Temperature

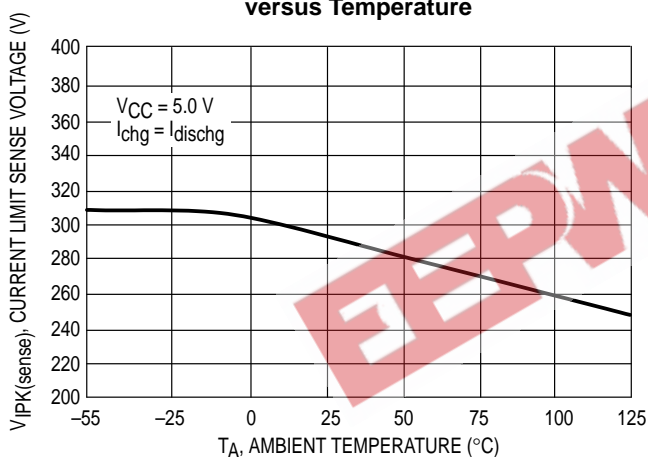
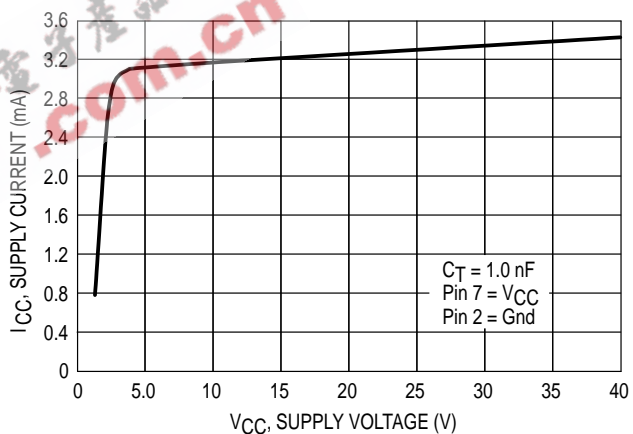


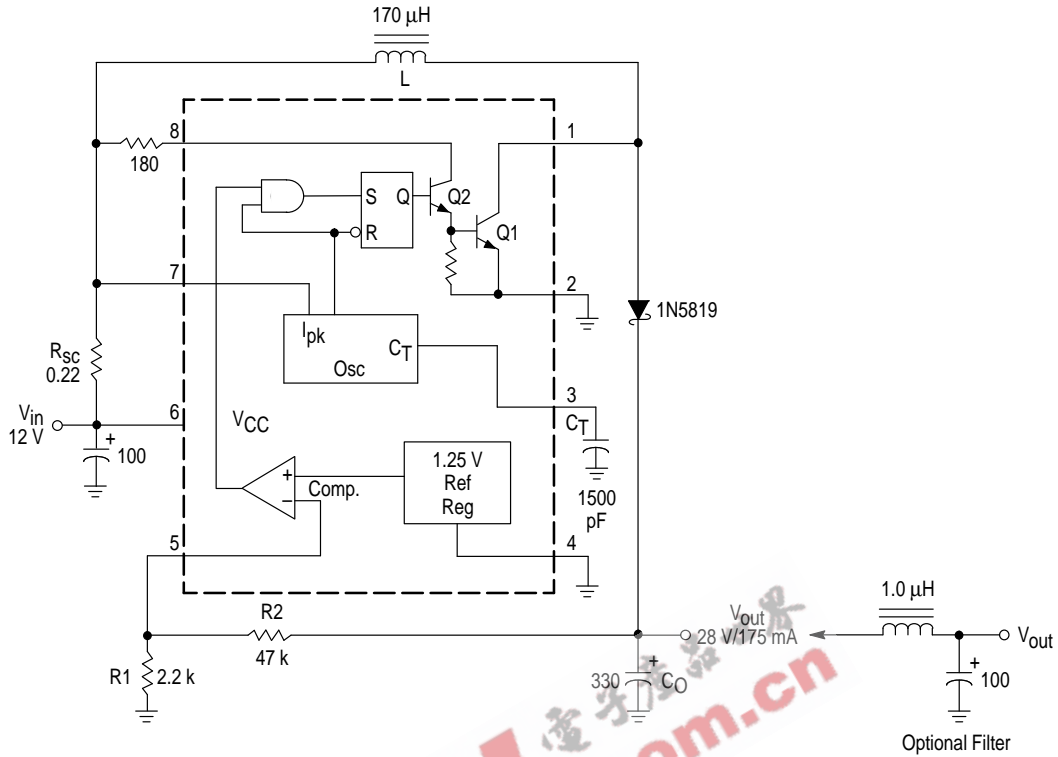
Figure 6. Standby Supply Current versus Supply Voltage



NOTE: 4. Low duty cycle pulse techniques are used during test to maintain junction temperature as close to ambient temperature as possible.

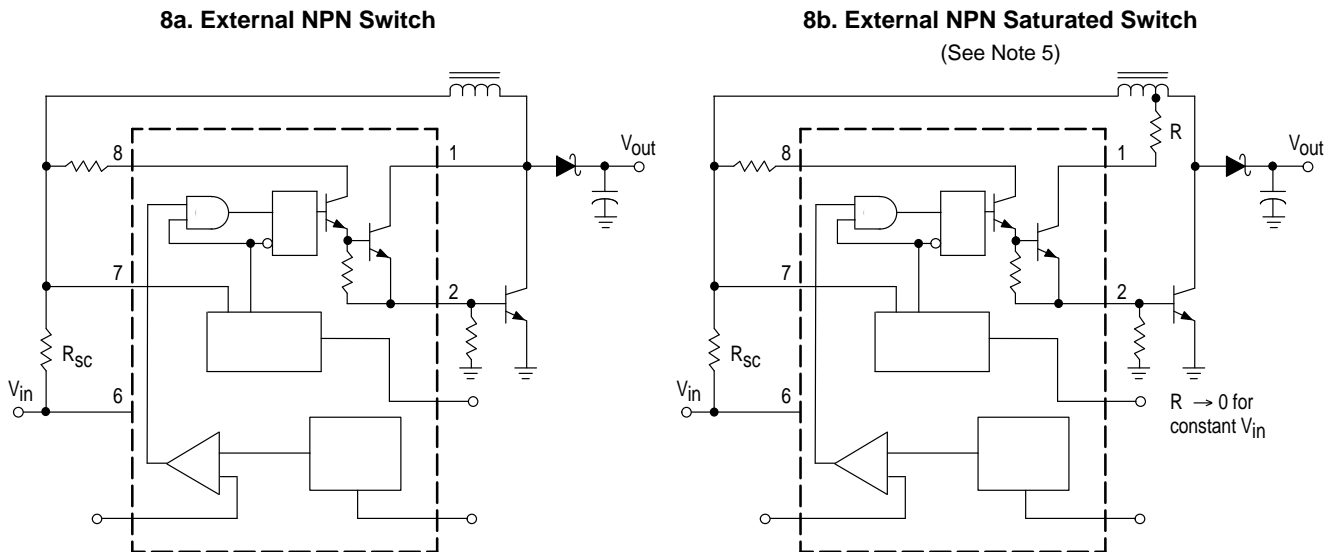
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Figure 7. Step-Up Converter



Test	Conditions	Results
Line Regulation	$V_{in} = 8.0 \text{ V to } 16 \text{ V}, I_O = 175 \text{ mA}$	30 mV = $\pm 0.05\%$
Load Regulation	$V_{in} = 12 \text{ V}, I_O = 75 \text{ mA to } 175 \text{ mA}$	10 mV = $\pm 0.017\%$
Output Ripple	$V_{in} = 12 \text{ V}, I_O = 175 \text{ mA}$	400 mVpp
Efficiency	$V_{in} = 12 \text{ V}, I_O = 175 \text{ mA}$	87.7%
Output Ripple With Optional Filter	$V_{in} = 12 \text{ V}, I_O = 175 \text{ mA}$	40 mVpp

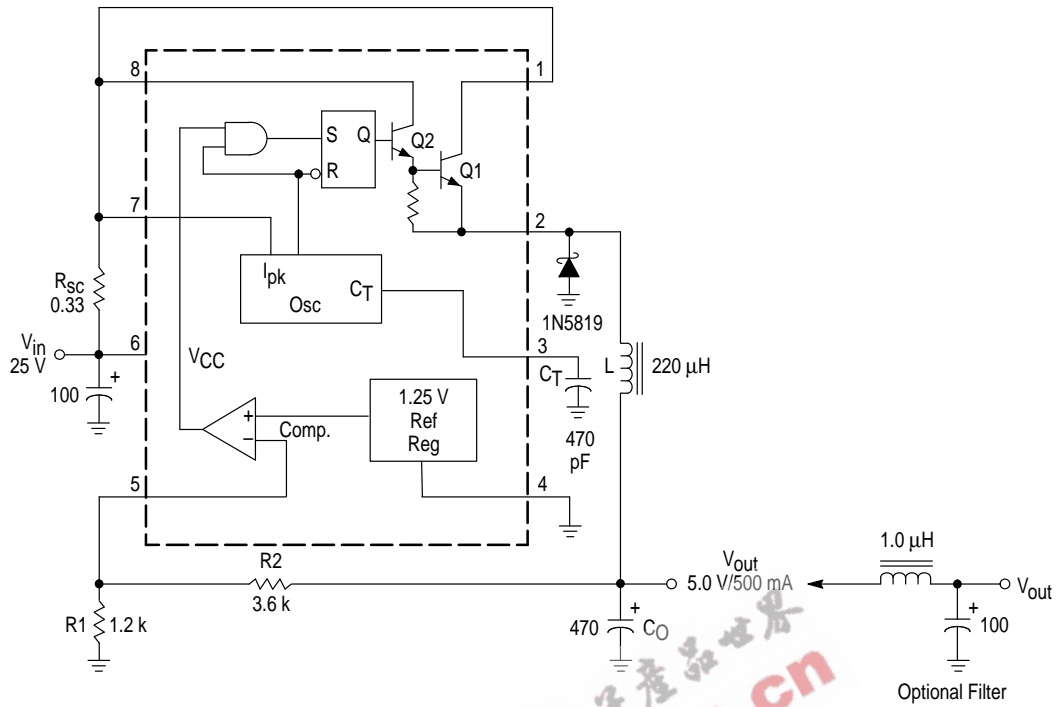
Figure 8. External Current Boost Connections for I_C Peak Greater than 1.5 A



NOTE: 5. If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents ($\leq 300 \text{ mA}$) and high driver currents ($\geq 30 \text{ mA}$), it may take up to $2.0 \mu\text{s}$ to come out of saturation. This condition will shorten the off time at frequencies $\geq 30 \text{ kHz}$, and is magnified at high temperatures. This condition does not occur with a Darlington configuration, since the output switch cannot saturate. If a non-Darlington configuration is used, the following output drive condition is recommended.

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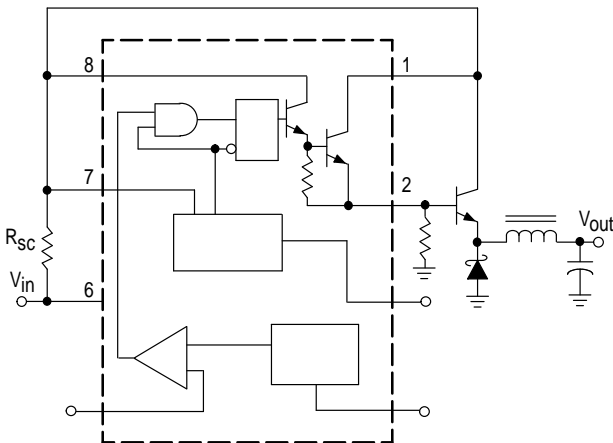
Figure 9. Step-Down Converter



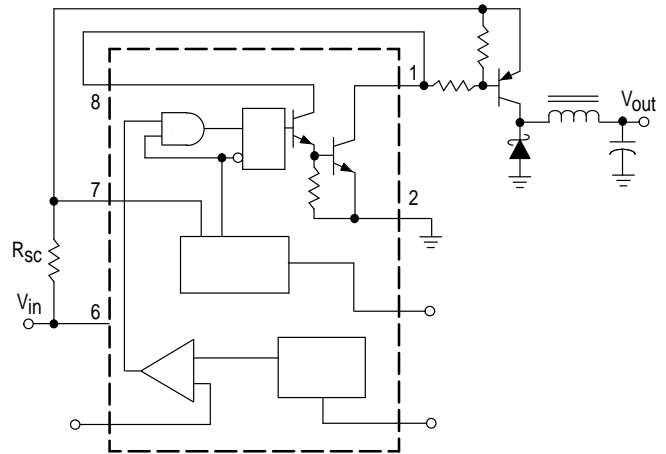
Test	Conditions	Results
Line Regulation	$V_{in} = 15\text{ V to }25\text{ V}, I_O = 500\text{ mA}$	$12\text{ mV} = \pm 0.12\%$
Load Regulation	$V_{in} = 25\text{ V}, I_O = 50\text{ mA to }500\text{ mA}$	$3.0\text{ mV} = \pm 0.03\%$
Output Ripple	$V_{in} = 25\text{ V}, I_O = 500\text{ mA}$	120 mVpp
Short Circuit Current	$V_{in} = 25\text{ V}, R_L = 0.1\ \Omega$	1.1 A
Efficiency	$V_{in} = 25\text{ V}, I_O = 500\text{ mA}$	83.7%
Output Ripple With Optional Filter	$V_{in} = 25\text{ V}, I_O = 500\text{ mA}$	40 mVpp

Figure 10. External Current Boost Connections for I_C Peak Greater than 1.5 A

10a. External NPN Switch

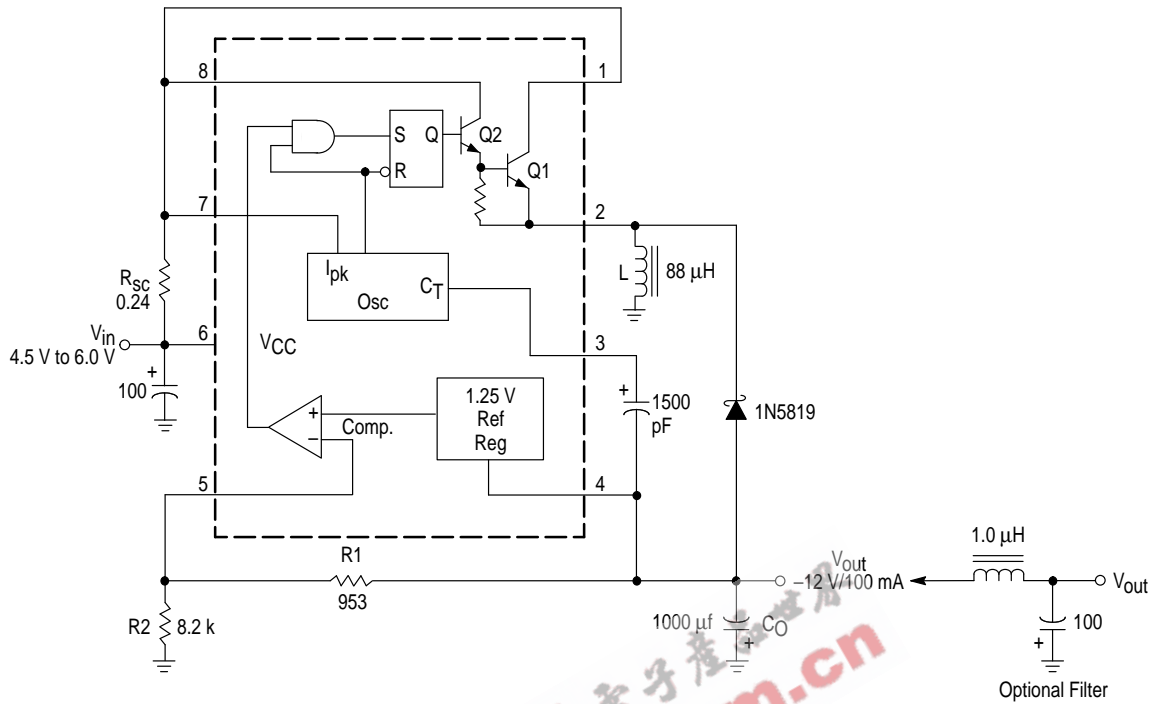


10b. External PNP Saturated Switch



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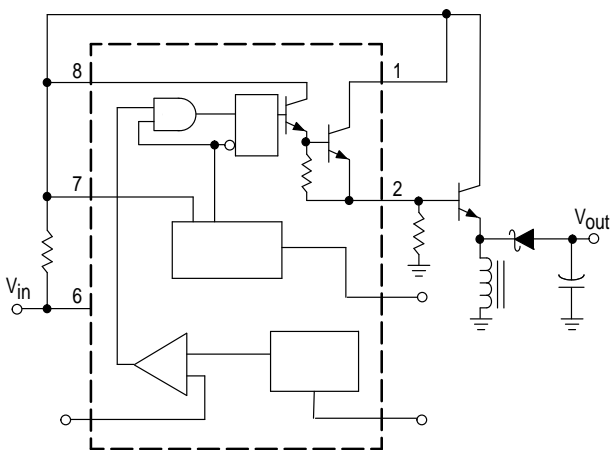
Figure 11. Voltage Inverting Converter



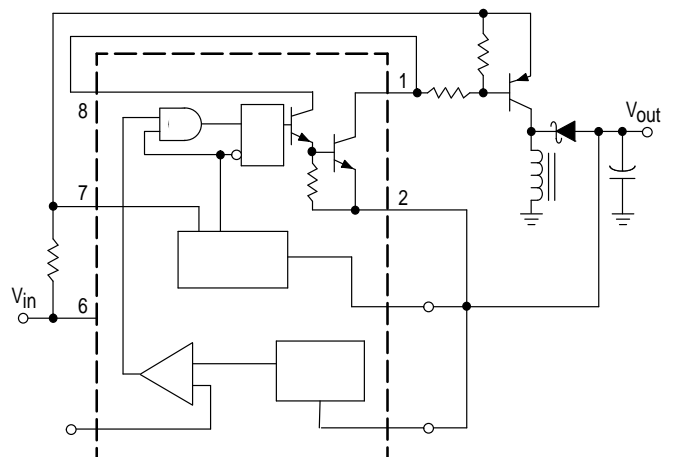
Test	Conditions	Results
Line Regulation	$V_{in} = 4.5 \text{ V to } 6.0 \text{ V}$, $I_O = 100 \text{ mA}$	$3.0 \text{ mV} = \pm 0.012\%$
Load Regulation	$V_{in} = 5.0 \text{ V}$, $I_O = 10 \text{ mA to } 100 \text{ mA}$	$0.022 \text{ V} = \pm 0.09\%$
Output Ripple	$V_{in} = 5.0 \text{ V}$, $I_O = 100 \text{ mA}$	500 mVpp
Short Circuit Current	$V_{in} = 5.0 \text{ V}$, $R_L = 0.1 \Omega$	910 mA
Efficiency	$V_{in} = 5.0 \text{ V}$, $I_O = 100 \text{ mA}$	62.2%
Output Ripple With Optional Filter	$V_{in} = 5.0 \text{ V}$, $I_O = 100 \text{ mA}$	70 mVpp

Figure 12. External Current Boost Connections for I_C Peak Greater than 1.5 A

12a. External NPN Switch

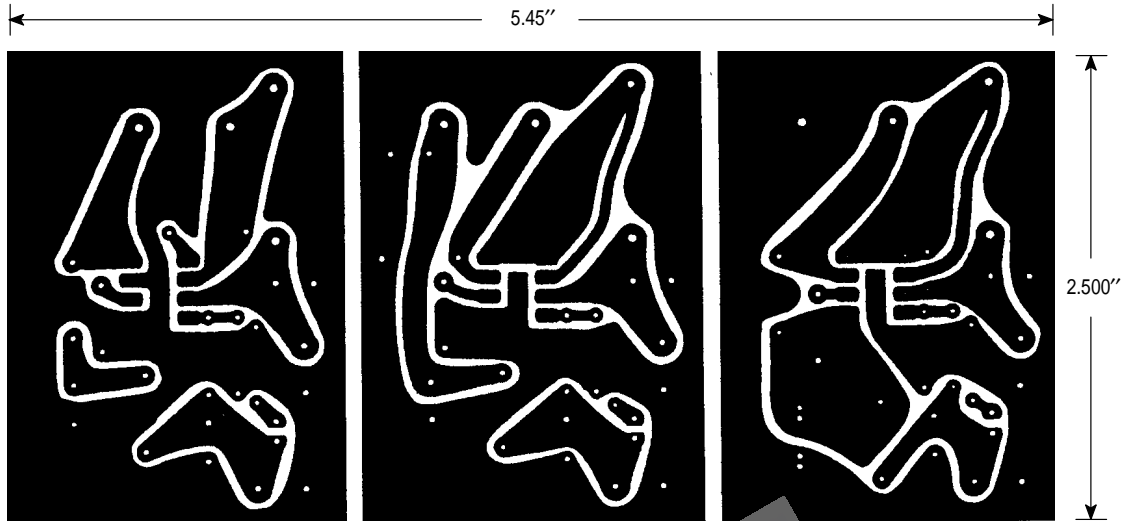


12b. External PNP Saturated Switch

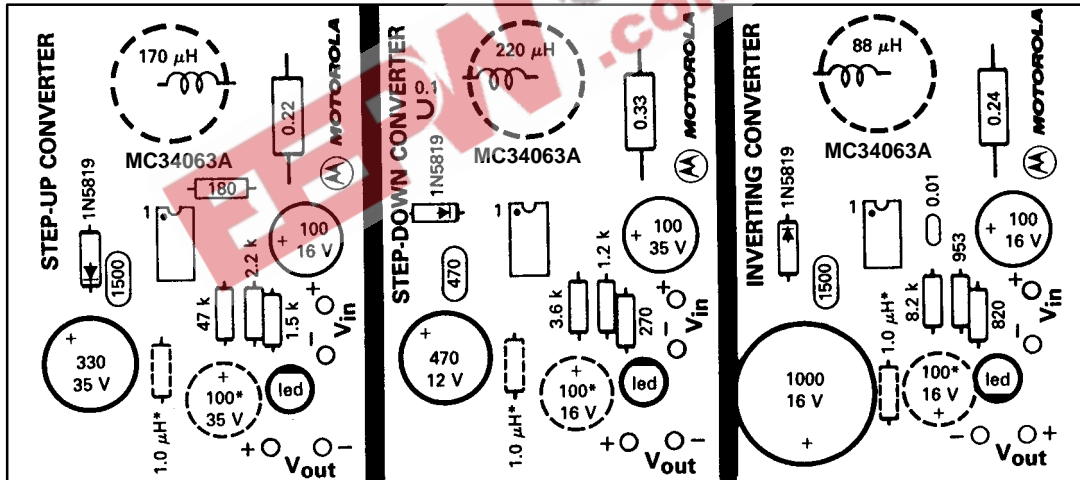


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Figure 13. Printed Circuit Board and Component Layout
(Circuits of Figures 7, 9, 11)



(Top view, copper foil as seen through the board from the component side)



(Top View, Component Side)

*Optional Filter.

INDUCTOR DATA

Converter	Inductance (μH)	Turns/Wire
Step-Up	170	38 Turns of #22 AWG
Step-Down	220	48 Turns of #22 AWG
Voltage-Inverting	88	28 Turns of #22 AWG

All inductors are wound on Magnetics Inc. 55117 toroidal core.

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Figure 14. Design Formula Table

Calculation	Step-Up	Step-Down	Voltage-Inverting
t_{on}/t_{off}	$\frac{V_{out} + V_F - V_{in(min)}}{V_{in(min)} - V_{sat}}$	$\frac{V_{out} + V_F}{V_{in(min)} - V_{sat} - V_{out}}$	$\frac{ V_{out} + V_F}{V_{in} - V_{sat}}$
$(t_{on} + t_{off})$	$\frac{1}{f}$	$\frac{1}{f}$	$\frac{1}{f}$
t_{off}	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$
t_{on}	$(t_{on} + t_{off}) - t_{off}$	$(t_{on} + t_{off}) - t_{off}$	$(t_{on} + t_{off}) - t_{off}$
C_T	$4.0 \times 10^{-5} t_{on}$	$4.0 \times 10^{-5} t_{on}$	$4.0 \times 10^{-5} t_{on}$
$I_{pk(switch)}$	$2I_{out(max)} \left(\frac{t_{on}}{t_{off}} + 1 \right)$	$2I_{out(max)}$	$2I_{out(max)} \left(\frac{t_{on}}{t_{off}} + 1 \right)$
R_{sc}	$0.3/I_{pk(switch)}$	$0.3/I_{pk(switch)}$	$0.3/I_{pk(switch)}$
$L_{(min)}$	$\left(\frac{V_{in(min)} - V_{sat}}{I_{pk(switch)}} \right) t_{on(max)}$	$\left(\frac{V_{in(min)} - V_{sat} - V_{out}}{I_{pk(switch)}} \right) t_{on(max)}$	$\left(\frac{V_{in(min)} - V_{sat}}{I_{pk(switch)}} \right) t_{on(max)}$
C_O	$9 \frac{I_{out} t_{on}}{V_{ripple(pp)}}$	$\frac{I_{pk(switch)}(t_{on} + t_{off})}{8V_{ripple(pp)}}$	$9 \frac{I_{out} t_{on}}{V_{ripple(pp)}}$

V_{sat} = Saturation voltage of the output switch.
 V_F = Forward voltage drop of the output rectifier.

The following power supply characteristics must be chosen:

V_{in} – Nominal input voltage.

V_{out} – Desired output voltage, $|V_{out}| = 1.25 \left(1 + \frac{R2}{R1} \right)$

I_{out} – Desired output current.

f_{min} – Minimum desired output switching frequency at the selected values of V_{in} and I_O .

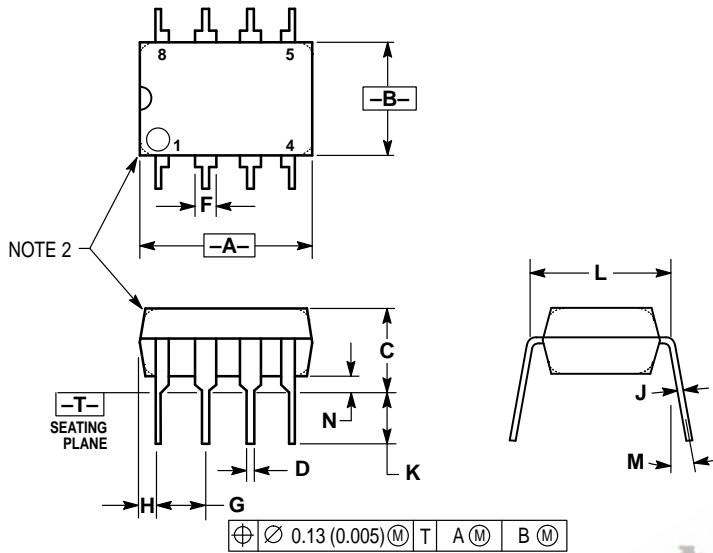
$V_{ripple(pp)}$ – Desired peak-to-peak output ripple voltage. In practice, the calculated capacitor value will need to be increased due to its equivalent series resistance and board layout. The ripple voltage should be kept to a low value since it will directly affect the line and load regulation.

NOTE: For further information refer to Application Note AN920A/D and AN954/D.

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OUTLINE DIMENSIONS

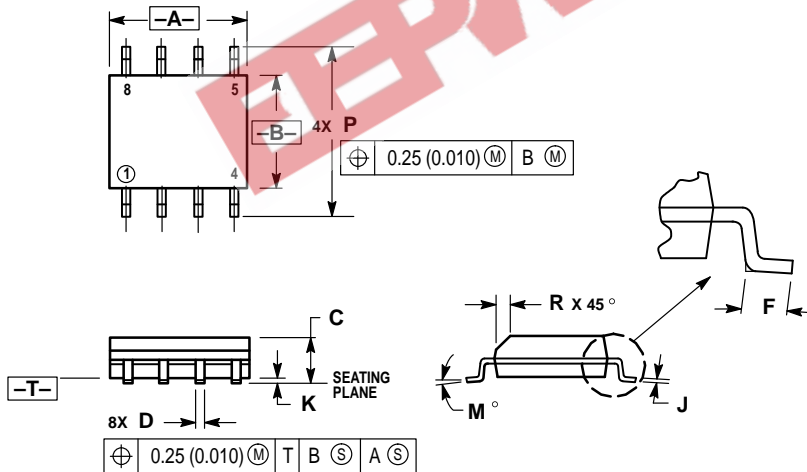
P, P1 SUFFIX
 PLASTIC PACKAGE
 CASE 626-05
 ISSUE K



NOTES:

1. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
2. PACKAGE CONTOUR OPTIONAL (ROUND OR SQUARE CORNERS).
3. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.

D SUFFIX
 PLASTIC PACKAGE
 CASE 751-05
 (SO-8)
 ISSUE P



NOTES:


1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.

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NOTES

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INTERNET: <http://Design-NET.com>

JAPAN: Nippon Motorola Ltd.; Tatsumi-SPD-JLDC, 6F Seibu-Butsuryu-Center,
3-14-2 Tatsumi Koto-Ku, Tokyo 135, Japan. 03-81-3521-8315

ASIA/PACIFIC: Motorola Semiconductors H.K. Ltd.; 8B Tai Ping Industrial Park,
51 Ting Kok Road, Tai Po, N.T., Hong Kong. 852-26629298



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MC34063A/D

