



**OPA337 OPA2337 OPA338 OPA2338** 

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# MicroSIZE, Single-Supply **CMOS OPERATIONAL AMPLIFIERS** MicroAmplifier™ Series

#### **FEATURES**

- MicroSIZE PACKAGES: SOT23-5 SOT23-8
- SINGLE-SUPPLY OPERATION
- RAIL-TO-RAIL OUTPUT SWING
- FET-INPUT:  $I_B = 10pA max$
- HIGH SPEED: OPA337: 3MHz, 1.2V/µs (G = 1) OPA338: 12.5MHz, 4.6V/us (G = 5)
- OPERATION FROM 2.5V to 5.5V
- HIGH OPEN-LOOP GAIN: 120dB
- LOW QUIESCENT CURRENT: 525µA/amp
- SINGLE AND DUAL VERSIONS

### APPLICATIONS

- BATTERY-POWERED INSTRUMENTS
- PHOTODIODE PRE-AMPS
- MEDICAL INSTRUMENTS
- TEST EQUIPMENT
- AUDIO SYSTEMS
- DRIVING ADCs
- CONSUMER PRODUCTS

OPA337, OPA338

SPICE Model available at www.burr-brown.com

#### DESCRIPTION

The OPA337 and OPA338 series rail-to-rail output CMOS operational amplifiers are designed for low cost and miniature applications. Packaged in the new SOT23-8, the OPA2337EA and OPA2338EA are Burr-Brown's smallest dual op amps. At 1/4 the size of a conventional SO-8 surface mount, they are ideal for space-sensitive applications.

Utilizing advanced CMOS technology, OPA337 and OPA338 op amps provide low bias current, high-speed operation, high open-loop gain, and rail-to-rail output swing. They operate on a single supply with operation as low as 2.5V while drawing only 525µA quiescent current. In addition, the input common-mode voltage range includes ground-ideal for single-supply operation.

The OPA337 series is unity-gain stable. The OPA338 series is optimized for gains greater than or equal to five. They are easy to use and free from phase inversion and overload problems found in some other op amps. Excellent performance is maintained as the amplifiers swing to their specified limits. The dual versions feature completely independent circuitry for lowest crosstalk and freedom from interaction, even when overdriven or overloaded.

	G = 1 S	TABLE	$G \ge 5 STABLE$		
PACKAGE	SINGLE OPA337	DUAL OPA2337	SINGLE OPA338	DUAL OPA2338	
SOT23-5	~		~		
SOT23-8		~		~	
MSOP-8	~				
SO-8	~	~	~	~	
DIP-8	~	~			

8 V+

7

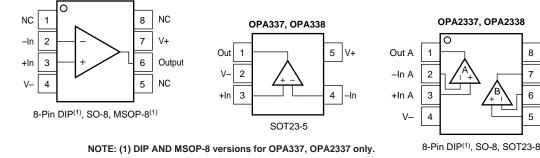
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5

Out B

–In B

+In B



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## SPECIFICATIONS: $V_S = 2.7V$ to 5.5V

At T<sub>A</sub> = +25°C, and R<sub>L</sub> = 25k $\Omega$  connected to V<sub>S</sub>/2, unless otherwise noted. **Boldface** limits apply over the specified temperature range, -40°C to +85°C, V<sub>S</sub> = 5V.

			OPA337NA, EA, UA, PA OPA2337EA, UA, PA OPA338NA, UA OPA2338EA, UA			
PARAMETER		CONDITION	MIN	TYP <sup>(1)</sup>	МАХ	UNITS
OFFSET VOLTAGEInput Offset Voltage $T_A = -40^{\circ}C$ to $+85^{\circ}C$ vs Temperaturevs Power Supply Rejection Ratio $T_A = -40^{\circ}C$ to $+85^{\circ}C$ Channel Separation (dual versions)	V <sub>os</sub> dV <sub>os</sub> /dT PSRR	$V_S = 2.7V$ to 5.5V $V_S = 2.7V$ to 5.5V dc		±0.5 ±2 25 0.3	±3 ± <b>3.5</b> 125 <b>125</b>	mV mV μV/°C μV/ν μV/ν μV/ν
INPUT BIAS CURRENT Input Bias Current $T_A = -40^{\circ}C$ to +85°C Input Offset Current	I <sub>B</sub> I <sub>OS</sub>		S	±0.2 See Typical Cur   ±0.2	±10 ve   ±10	pA pA
NOISE Input Voltage Noise, f = 0.1Hz to 10H Input Voltage Noise Density, f = 1kHz Current Noise Density, f = 1kHz				6 26 0.6		μVp-p nV/√Hz fA/√Hz
$\label{eq:interm} \begin{array}{l} \text{INPUT VOLTAGE RANGE} \\ \text{Common-Mode Voltage Range} \\ \text{Common-Mode Rejection Ratio} \\ \text{T}_{\text{A}} = -40^{\circ}\text{C} \text{ to } +85^{\circ}\text{C} \end{array}$	V <sub>CM</sub> CMRR	$\begin{array}{l} \textbf{T_A} = -40^{\circ}\text{C to } +85^{\circ}\text{C} \\ -0.2\text{V} < \text{V}_{\text{CM}} < (\text{V+}) - 1.2\text{V} \\ -0.2\text{V} < \text{V}_{\text{CM}} < (\text{V+}) - 1.2\text{V} \end{array}$	- <b>0.2</b> 74 74	90	(V+) – 1.2	V dB dB
INPUT IMPEDANCE Differential Common-Mode			· R	10 <sup>13</sup>    2 10 <sup>13</sup>    4		Ω    pF Ω    pF
OPEN-LOOP GAIN Open-Loop Voltage Gain $T_A = -40^{\circ}C$ to +85°C $T_A = -40^{\circ}C$ to +85°C	A <sub>OL</sub>	$\begin{array}{l} R_L = 25 k \Omega, \ 125 mV < V_O < (V+) - 125 mV \\ R_L = 25 k \Omega, \ 125 mV < V_O < (V+) - 125 mV \\ R_L = 5 k \Omega, \ 500 mV < V_O < (V+) - 500 mV \\ R_L = 5 k \Omega, \ 500 mV < V_O < (V+) - 500 mV \end{array}$	100 100 100 100	120 114		dB dB dB dB
OPA337 FREQUENCY RESPONSE Gain-Bandwidth Product Slew Rate Settling Time: 0.1% 0.01% Overload Recovery Time Total Harmonic Distortion + Noise	GBW SR THD+N	$\begin{split} V_{S} &= 5V, \ G = 1 \\ V_{S} &= 5V, \ G = 1 \\ V_{S} &= 5V, \ 2V \ Step, \ C_{L} &= 100pF, \ G = 1 \\ V_{S} &= 5V, \ 2V \ Step, \ C_{L} &= 100pF, \ G = 1 \\ V_{IN} \bullet G &= V_{S} \\ V_{S} &= 5V, \ V_{O} &= 3Vp\text{-p}, \ G &= 1, \ f &= 1\text{kHz} \end{split}$		3 1.2 2 2.5 2 0.001		MHz V/μs μs μs μs %
OPA338 FREQUENCY RESPONSE Gain-Bandwidth Product Slew Rate Settling Time: 0.1% 0.01% Overload Recovery Time Total Harmonic Distortion + Noise	GBW SR THD+N	$\begin{split} V_{S} &= 5V, \ G = 5 \\ V_{S} &= 5V, \ G = 5 \\ V_{S} &= 5V, \ 2V \ Step, \ C_{L} &= 100 \text{pF}, \ G = 5 \\ V_{S} &= 5V, \ 2V \ Step, \ C_{L} &= 100 \text{pF}, \ G = 5 \\ V_{IN} \bullet G &= V_{S} \\ V_{S} &= 5V, \ V_{O} &= 3V \text{p-p}, \ G &= 5, \ f &= 1 \text{kHz} \end{split}$		12.5 4.6 1.4 1.9 0.5 0.0035		MHz V/μs μs μs %
OUTPUT Voltage Output Swing from Rail <sup>(2)</sup> $T_A = -40^{\circ}C$ to +85°C $T_A = -40^{\circ}C$ to +85°C		$\begin{split} R_L &= 25 k \Omega, \ A_{OL} \geq 100 dB \\ R_L &= 25 k \Omega, \ A_{OL} \geq 100 dB \\ R_L &= 5 k \Omega, \ A_{OL} \geq 100 dB \\ R_L &= 5 k \Omega, \ A_{OL} \geq 100 dB \\ R_L &= 5 k \Omega, \ A_{OL} \geq 100 dB \end{split}$		40 150	125 <b>125</b> 500 <b>500</b>	mV mV mV mV
Short-Circuit Current Capacitive Load Drive	I <sub>SC</sub> C <sub>LOAD</sub>		5	±9 See Typical Cur	ve	mA
POWER SUPPLYSpecified Voltage RangeMinimum Operating VoltageQuiescent Current (per amplifier) $T_A = -40^{\circ}C$ to $+85^{\circ}C$	V <sub>S</sub> I <sub>Q</sub>	$T_{A} = -40^{\circ}C \text{ to } +85^{\circ}C$ $I_{O} = 0$ $I_{O} = 0$	2.7	2.5 0.525	5.5 1 1.2	V V mA mA

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## SPECIFICATIONS: V<sub>S</sub> = 2.7V to 5.5V (Cont.)

At T<sub>A</sub> = +25°C, and R<sub>L</sub> = 25k $\Omega$  connected to V<sub>S</sub>/2, unless otherwise noted. Boldface limits apply over the specified temperature range, -40°C to +85°C, V<sub>S</sub> = 5V.

		OPA337NA, EA, UA, PA OPA2337EA, UA, PA OPA338NA, UA OPA2338EA, UA			
PARAMETER	CONDITION	MIN	ТҮР	MAX	UNITS
TEMPERATURE RANGE   Specified Range   Operating Range   Storage Range   Thermal Resistance   SOT23-5 Surface Mount   SOT23-8 Surface Mount   MSOP-8   SO-8 Surface Mount   8-Pin DIP		-40 -55 -55	200 200 150 150 100	+85 +125 +125	°C °C °C/W °C/W °C/W °C/W °C/W

NOTES: (1) V<sub>S</sub> = 5V. (2) Output voltage swings are measured between the output and negative and positive power supply rails.

#### **ABSOLUTE MAXIMUM RATINGS(1)**

Supply Voltage	5.5V
Input Voltage <sup>(2)</sup>	(V–) –0.5V to (V+) +0.5V
Input Current <sup>(2)</sup>	10mA
Output Short Circuit <sup>(3)</sup>	Continuous
Operating Temperature	–55°C to +125°C
Storage Temperature	–55°C to +125°C
Junction Temperature	150°C
Lead Temperature (soldering, 10s)	300°C

NOTES: (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum ratings for extended periods may degrade device reliability. (2) Input signal voltage is limited by internal diodes connected to power supplies. See text. (3) Short circuit to ground, one amplifier per package.

### ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

PRODUCT	DESCRIPTION	PACKAGE	PACKAGE DRAWING NUMBER	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER <sup>(1)</sup>	TRANSPORT MEDIA
OPA337 Series							
OPA337NA "	Single, G = 1 Stable	5-Lead SOT23-5 "	331 "	–40°C to +85°C "	C37 "	OPA337NA/250 OPA337NA/3K	Tape and Reel Tape and Reel
OPA337EA "	Single, G = 1 Stable	MSOP-8 "	337 "	–40°C to +85°C "	G37 "	OPA337EA/250 OPA337EA/2K5	Tape and Reel Tape and Reel
OPA337PA	Single, G = 1 Stable	8-Pin DIP	006	-40°C to +85°C	OPA337PA	OPA337PA	Rails
OPA337UA "	Single, G = 1 Stable	SO-8 Surface Mount	182 "	–40°C to +85°C "	OPA337UA "	OPA337UA OPA337UA/2K5	Rails Tape and Reel
OPA2337EA "	Dual, G = 1 Stable	8-Lead SOT23-8 "	348 "	–40°C to +85°C "	A7 "	OPA2337EA/250 OPA2337EA/3K	Tape and Reel Tape and Reel
OPA2337PA	Dual, G = 1 Stable	8-Pin DIP	006	–40°C to +85°C	OPA2337PA	OPA2337PA	Rails
OPA2337UA "	Dual, G = 1 Stable	SO-8 Surface Mount	182 "	–40°C to +85°C "	OPA2337UA "	OPA2337UA OPA2337UA/2K5	Rails Tape and Reel
OPA338 Series							
OPA338NA "	Single, $G \ge 5$ Stable	5-Lead SOT23-5 "	331 "	–40°C to +85°C "	A38 "	OPA338NA/250 OPA338NA/3K	Tape and Reel Tape and Reel
OPA338UA "	Single, $G \ge 5$ Stable	SO-8 Surface Mount	182 "	–40°C to +85°C "	OPA338UA "	OPA338UA OPA338UA/2K5	Rails Tape and Reel
OPA2338EA "	Dual, $G \ge 5$ Stable	8-Lead SOT23-8 "	348 "	–40°C to +85°C "	A8 "	OPA2338EA/250 OPA2338EA/3K	Tape and Reel Tape and Reel
OPA2338UA "	Dual, G $\ge$ 5 Stable	SO-8 Surface Mount	182 "	–40°C to +85°C "	OPA2338UA "	OPA2338UA OPA2338UA/2K5	Rails Tape and Reel

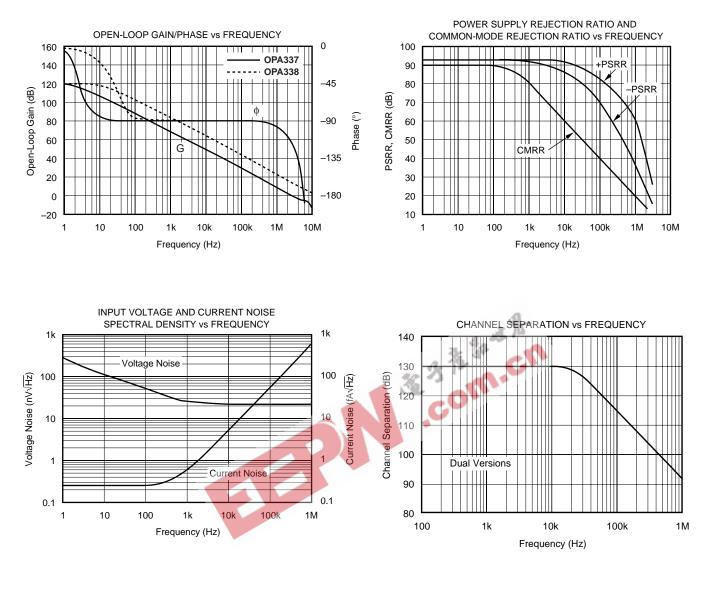
NOTES: (1) Models with a slash (/) are available only in Tape and Reel in the quantities indicated (e.g., /2K5 indicates 2500 devices per reel). Ordering 2500 pieces of "OPA2337UA/2K5" will get a single 2500-piece Tape and Reel.

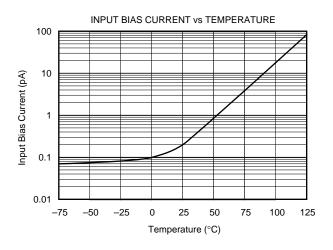
#### PACKAGE/ORDERING INFORMATION

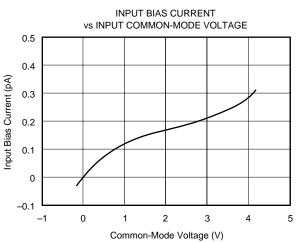


## **TYPICAL PERFORMANCE CURVES**

At T<sub>A</sub> = +25°C, V<sub>S</sub> = +5V, and R<sub>L</sub> = 25k $\Omega$  connected to V<sub>S</sub>/2, unless otherwise noted.





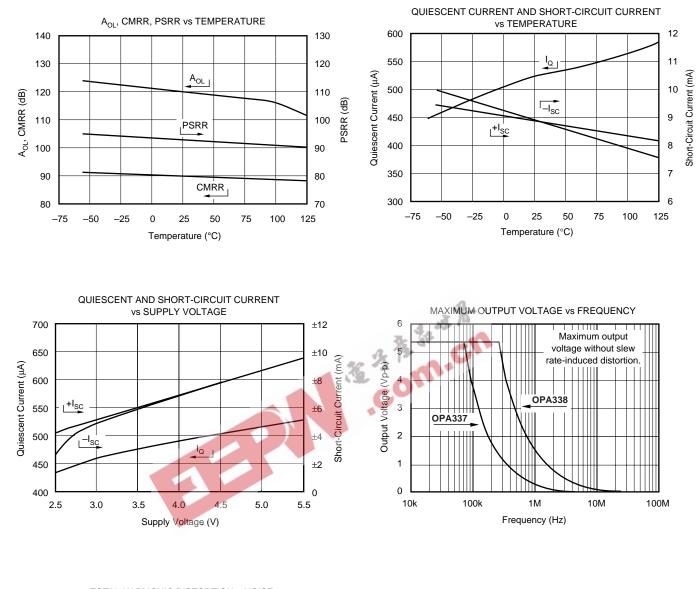


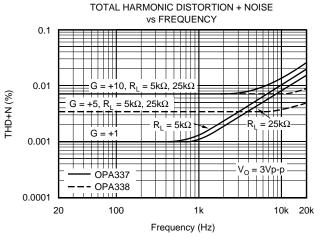


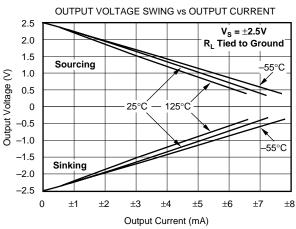
OPA337, OPA2337 OPA338, OPA2338

### TYPICAL PERFORMANCE CURVES (Cont.)

At  $T_A = +25^{\circ}C$ ,  $V_S = +5V$ , and  $R_L = 25k\Omega$  connected to  $V_S/2$ , unless otherwise noted.







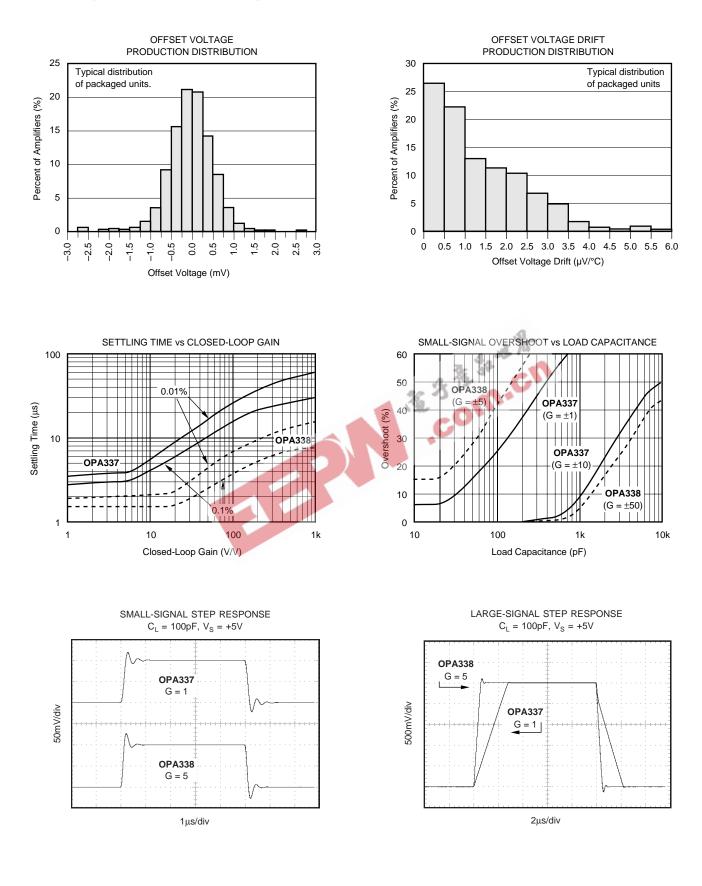




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### **TYPICAL PERFORMANCE CURVES (Cont.)**

At  $T_A = +25^{\circ}C$ ,  $V_S = +5V$ , and  $R_L = 25k\Omega$  connected to  $V_S/2$ , unless otherwise noted.





## **APPLICATIONS INFORMATION**

The OPA337 series and OPA338 series are fabricated on a state-of-the-art CMOS process. The OPA337 series is unitygain stable. The OPA338 series is optimized for gains greater than or equal to five. Both are suitable for a wide range of general purpose applications. Power supply pins should be bypassed with  $0.01\mu$ F ceramic capacitors.

#### **OPERATING VOLTAGE**

The OPA337 series and OPA338 series can operate from a +2.5V to +5.5V single supply with excellent performance. Unlike most op amps which are specified at only one supply voltage, these op amps are specified for real-world applications; a single limit applies throughout the +2.7V to +5.5V supply range. This allows a designer to have the same assured performance at any supply voltage within the specified voltage range. Most behavior remains unchanged throughout the full operating voltage range. Parameters which vary significantly with operating voltage are shown in typical performance curves.

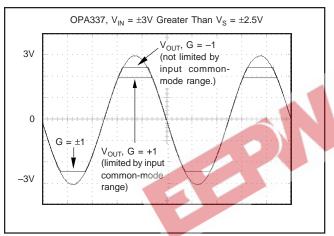


FIGURE 1. OPA337—No Phase Inversion with Inputs Greater than the Power Supply Voltage.

#### **INPUT VOLTAGE**

The input common-mode range extends from (V-) - 0.2V to (V+) - 1.2V. For normal operation, inputs should be limited to this range. The absolute maximum input voltage is 500mV beyond the supplies. Inputs greater than the input common-mode range but less than maximum input voltage, while not valid, will not cause any damage to the op amp. Furthermore, if input current is limited the inputs may go beyond the power supplies without phase inversion (Figure 1) unlike some other op amps.

Normally, input currents are 0.2pA. However, large inputs (greater than 500mV beyond the supply rails) can cause excessive current to flow in or out of the input pins. Therefore, as well as keeping the input voltage below the maximum rating, it is also important to limit the input current to less than 10mA. This is easily accomplished with an input resistor as shown in Figure 2.

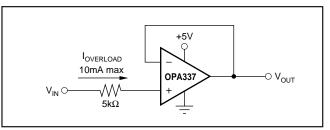


FIGURE 2. Input Current Protection for Voltages Exceeding the Supply Voltage.

#### **USING THE OPA338 IN LOW GAINS**

The OPA338 series is optimized for gains greater than or equal to five. It has significantly wider bandwidth (12.5MHz) and faster slew rate ( $4.6V/\mu s$ ) when compared to the OPA337 series. The OPA338 series can be used in lower gain configurations at low frequencies while maintaining its high slew rate with the proper compensation.

Figure 3 shows the OPA338 in a unity-gain buffer configuration. At dc, the compensation capacitor  $C_1$  is effectively "open" resulting in 100% feedback (closed-loop gain = 1). As frequency increases,  $C_1$  becomes lower impedance and closed-loop gain increases, eventually becoming  $1 + R_2/R_1$ (in this case five, which is equal to the minimum gain required for stability).

The required compensation capacitor value can be determined from the following equation:

$$C_1 = 1/(2\pi f_C R_1)$$

Since  $f_C$  may shift with process variations, it is recommended that a value less than  $f_C$  be used for determining  $C_1$ . With  $f_C = 1$ MHz and  $R_1 = 2.5k\Omega$ , the compensation capacitor is about 68pF.

The selection of the compensation capacitor  $C_1$  is important. A proper value ensures that the closed-loop circuit gain is greater than or equal to five at high frequencies. Referring to the "Open-Loop Gain vs Frequency" plot in the Typical Performance Curves section, the OPA338 gain line (dashed in the curve) has a constant slope (-20dB/decade) up to approximately 3MHz. This frequency is referred to as  $f_C$ . Beyond  $f_C$  the slope of the curve increases, suggesting that closed-loop gains less than 5 are not appropriate.

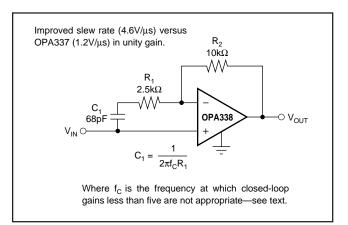


FIGURE 3. Compensation of OPA338 for Unity-Gain Buffer.

OPA337, OPA2337 OPA338, OPA2338



Figure 4 shows a compensation technique using an inverting configuration. The low frequency gain is set by the resistor ratio while the high frequency gain is set by the capacitor ratio. As with the noninverting circuit, for frequencies above  $f_C$  the gain must be greater than the recommended minimum stable gain for the op amp.

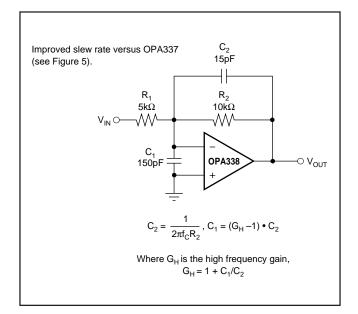


FIGURE 4. Inverting Compensation Circuit of OPA338 for Low Gain.

Resistors  $R_1$  and  $R_2$  are chosen to set the desired dc signal gain. Then the value for  $C_2$  is determined as follows:

$$C_2 = 1/(2\pi f_C R_2)$$

 $C_1$  is determined from the desired high frequency gain ( $G_H$ ):

$$C_1 = (G_H - 1) \bullet$$

For a desired dc gain of 2 and high frequency gain of 10, the following resistor and capacitor values result:

$R_1 = 10k\Omega$	$C_1 = 150 pF$
$R_2 = 5k\Omega$	$C_2 = 15 pF$

The capacitor values shown are the nearest standard values. Capacitor values may need to be adjusted slightly to optimize performance. For more detailed information, consult the OPA686 product data sheet.

Figure 5 shows the large-signal transient response using the circuit given in Figure 4. As shown, the OPA338 is stable in low gain applications and provides improved slew rate performance when compared to the OPA337.

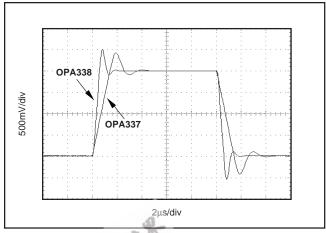


FIGURE 5. G = 2, Slew-Rate Comparison of OPA338 and OPA337.

## TYPICAL APPLICATION

Figure 6 shows the OPA2337 in a typical application. The ADS7822 is a 12-bit, micro-power sampling analog-todigital converter available in the tiny MSOP-8 package. As with the OPA2337, it operates with a supply voltage as low as +2.7V. When used with the miniature SOT23-8 package of the OPA2337, the circuit is ideal for spacelimited and low power applications. In addition, OPA2337's high input impedance allows large value resistors to be used which results in small physical capacitors, further reducing circuit size. For further information, consult the ADS7822 product data sheet.



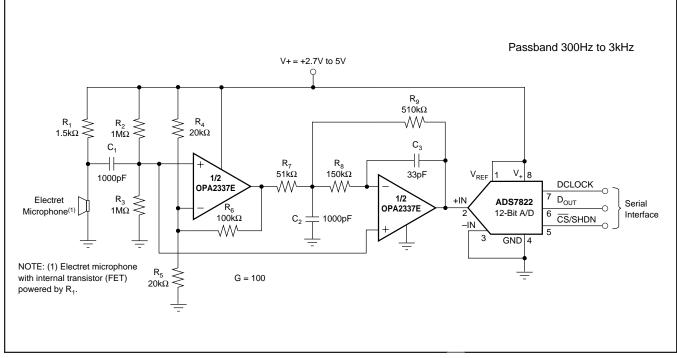


FIGURE 6. Low Power, Single-Supply, Speech Bandpass Filtered Data Acquisition System.

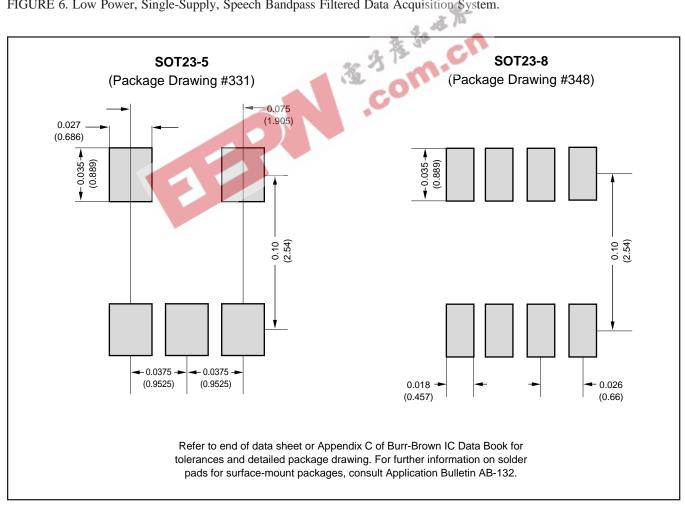


FIGURE 7. Recommended SOT23-5 and SOT23-8 Solder Footprints.

