



OPA241 OPA4241

OPA251 OPA2251 OPA4251

Single-Supply, *Micro*POWER OPERATIONAL AMPLIFIERS

OPA241 Family optimized for +5V supply. **OPA251 Family** optimized for ±15V supply.

FEATURES

- MicroPOWER: I_Q = 25μA
- SINGLE-SUPPLY OPERATION
- RAIL-TO-RAIL OUTPUT (within 50mV)
- WIDE SUPPLY RANGE
 Single Supply: +2.7V to +36V
 Dual Supply: ±1.35V to ±18V
- LOW OFFSET VOLTAGE: ±250µV max
- HIGH COMMON-MODE REJECTION: 124dB
- HIGH OPEN-LOOP GAIN: 128dB
- SINGLE, DUAL, AND QUAD

APPLICATIONS

- BATTERY OPERATED INSTRUMENTS
- PORTABLE DEVICES
- MEDICAL INSTRUMENTS
- TEST EQUIPMENT

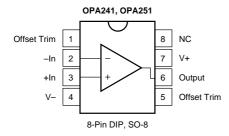
DESCRIPTION

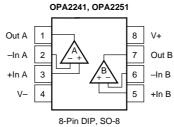
The OPA241 series and OPA251 series are specifically designed for battery powered, portable applications. In addition to very low power consumption (25 μ A), these amplifiers feature low offset voltage, rail-to-rail output swing, high common-mode rejection, and high open-loop gain.

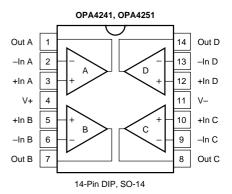
The OPA241 series is optimized for operation at low power supply voltage while the OPA251 series is optimized for high power supplies. Both can operate from either single ($\pm 2.7V$ to $\pm 36V$) or dual supplies ($\pm 1.35V$ to $\pm 18V$). The input common-mode voltage range extends 200mV below the negative supply—ideal for single-supply applications.

They are unity-gain stable and can drive large capacitive loads. Special design considerations assure that these products are easy to use. High performance is maintained as the amplifiers swing to their specified limits. Because the initial offset voltage ($\pm 250 \mu V$ max) is so low, user adjustment is usually not required. However, external trim pins are provided for special applications (single versions only).

The OPA241 and OPA251 (single versions) are available in standard 8-pin DIP and SO-8 surface-mount packages. The OPA2241 and OPA2251 (dual versions) come in 8-pin DIP and SO-8 surface-mount packages. The OPA4241 and OPA4251 (quad versions) are available in 14-pin DIP and SO-14 surface-mount packages. All are fully specified from -40°C to +85°C and operate from -55°C to +125°C.







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SPECIFICATIONS: $V_S = 2.7V$ to 5V

At T_A = +25°C, R_L = 100k Ω connected to $V_S/2$, unless otherwise noted. **Boldface** limits apply over the specified temperature range, $T_A = -40$ °C to +85°C.

			OPA241UA, PA OPA2241UA, PA OPA4241UA, PA		OPA251UA, PA OPA2251UA, PA OPA4251UA, PA				
PARAMETER		CONDITION	MIN	TYP ⁽¹⁾	MAX	MIN	TYP(1)	MAX	UNITS
OFFSET VOLTAGE Input Offset Voltage $T_A = -40^{\circ}C \text{ to } +85^{\circ}C$ vs Temperature vs Power Supply $T_A = -40^{\circ}C \text{ to } +85^{\circ}C$ Channel Separation (dual, quad)	V _{OS} dV _{OS} /dT PSRR	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$ $V_S = 2.7\text{V to } 36\text{V}$ $V_S = 2.7\text{V to } 36\text{V}$		±50 ±100 ±0.4 3	±250 ± 400 30 30		±100 ±130 ±0.6 *	*	μV μV/°C μV/ν μV/ν μV/ν
INPUT BIAS CURRENT Input Bias Current ⁽²⁾ $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$ Input Offset Current $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	I _B			-4 ±0.1	-20 - 25 ±2 ± 2		*		nA nA nA nA
NOISE Input Voltage Noise, f = 0.1Hz to 1 Input Voltage Noise Density, f = 1k Current Noise Density, f = 1kHz				1 45 40			* * *		μVp-p nV/√Hz fA/√Hz
INPUT VOLTAGE RANGE Common-Mode Voltage Range Common-Mode Rejection Ratio $T_A = -40^{\circ}\text{C}$ to +85°C	V _{CM} CMRR	$V_{CM} = -0.2V$ to $(V+) -0.8V$ $V_{CM} = 0V$ to $(V+) -0.8V$	-0.2 80 80	106	(V+) -0.8		*		V dB dB
INPUT IMPEDANCE Differential Common-Mode			逐为	10 ⁷ 2 10 ⁹ 4	1.0		* *		Ω pF Ω pF
OPEN-LOOP GAIN Open-Loop Voltage Gain $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	A _{OL}	$\begin{split} R_L &= 100 k \Omega, \ V_O = (V-) + 100 mV \ to \ (V+) - 100 mV \\ R_L &= 100 k \Omega, \ V_O = (V-) + 100 mV \ to \ (V+) - 100 mV \\ R_L &= 10 k \Omega, \ V_O = (V-) + 200 mV \ to \ (V+) - 200 mV \\ R_L &= 10 k \Omega, \ V_O = (V-) + 200 mV \ to \ (V+) - 200 mV \end{split}$	10 0 1 00 100 100	120 120			*		dB dB dB dB
FREQUENCY RESPONSE Gain-Bandwidth Product Slew Rate Overload Recovery Time	GBW SR	$V_S = 5V$, $G = 1$ $V_{IN} \cdot G = V_S$		35 0.01 60			* * *		kHz V/μs μs
OUTPUT Voltage Output Swing from Rail ⁽³⁾ $T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	Vo	$R_L = 100k\Omega$ to $V_S/2$, $A_{OL} \ge 70dB$ $R_L = 100k\Omega$ to $V_S/2$, $A_{OL} \ge 100dB$ $R_L = 100k\Omega$ to $V_S/2$, $A_{OL} \ge 100dB$ $R_L = 10k\Omega$ to $V_S/2$, $A_{OL} \ge 100dB$		50 75 100	100 100 200		* *		mV mV mV
T _A = -40°C to +85°C Short-Circuit Current Single Versions Dual, Quad Versions Capacitive Load Drive	I _{SC}	$R_L = 10k\Omega$ to $V_S/2$, $A_{OL} \ge 100dB$	See	-24/+4 -30/+4 Typical C	200		* * *		mV mA mA
POWER SUPPLY Specified Voltage Range Operating Voltage Range Quiescent Current (per amplifier) T _A = -40°C to +85°C	V _S	$T_A = -40^{\circ}C$ to +85°C $I_O = 0$ $I_O = 0$	+2.7	+2.7 to +5 ±25		*	*	*	V V μΑ μΑ
TEMPERATURE RANGE Specified Range Operating Range Storage Range Thermal Resistance 8-Pin DIP	$ heta_{\sf JA}$		-40 -55 -55	100	+85 +125 +125	* * *	*	* *	°C/W
SO-8 Surface Mount 14-Pin DIP SO-14 Surface Mount				150 150 80 100			* * *		°C/W

^{*} Specifications the same as OPA241UA, PA.

NOTES: (1) V_S = +5V. (2) The negative sign indicates input bias current flows out of the input terminals. (3) Output voltage swings are measured between the output and power supply rails.

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SPECIFICATIONS: $V_S = \pm 15V$

At T_A = +25°C, R_L = 100k Ω connected to ground, unless otherwise noted. **Boldface** limits apply over the specified temperature range, T_A = -40°C to +85°C.

			OPA241UA, PA OPA2241UA, PA OPA4241UA, PA			OPA251UA, PA OPA2251UA, PA OPA4251UA, PA			
PARAMETER		CONDITION	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
OFFSET VOLTAGE Input Offset Voltage T _A = -40°C to +85°C vs Temperature vs Power Supply T _A = -40°C to +85°C Channel Separation (dual, quad)	V _{OS} dV _{OS} /dT PSRR	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$ $V_S = \pm 1.35\text{V to } \pm 18\text{V}$ $V_S = \pm 1.35\text{V to } \pm 18\text{V}$		±100 ±150 ±0.6 *	* *		±50 ±100 ±0.5 3	±250 ± 300 30 30	μV μV μV/°C μV/V μV/V
INPUT BIAS CURRENT Input Bias Current ⁽¹⁾ $T_A = -40^{\circ}\text{C}$ to +85°C Input Offset Current $T_A = -40^{\circ}\text{C}$ to +85°C	I _B			*			-4 ±0.1	-20 - 25 ±2 ±2	nA nA nA nA
NOISE Input Voltage Noise, f = 0.1Hz to 1 Input Voltage Noise Density, f = 1 Current Noise Density, f = 1kHz				* * *			1 45 40		μVp-p nV/√Hz fA/√Hz
INPUT VOLTAGE RANGE Common-Mode Voltage Range Common-Mode Rejection Ratio $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	V _{CM} CMRR	$V_{CM} = -15.2V$ to 14.2V $V_{CM} = -15V$ to 14.2V	1 4 4 A	*		(V-) -0.2 100 100	124	(V+) -0.8	V dB dB
INPUT IMPEDANCE Differential Common-Mode		3 3	OLL	*			10 ⁷ 2 10 ⁹ 4		$\Omega \parallel pF$ $\Omega \parallel pF$
OPEN-LOOP GAIN Open-Loop Voltage Gain $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	A _{OL}	$\begin{split} R_L &= 100 k \Omega, \ V_O = -14.75 V \ to \ +14.75 V \\ R_L &= 100 k \Omega, \ V_O = -14.75 V \ to \ +14.75 V \\ R_L &= 20 k \Omega, \ V_O = -14.7 V \ to \ +14.7 V \\ R_L &= 20 k \Omega, \ V_O = -14.7 V \ to \ +14.7 V \end{split}$		*		100 100 100 100	128 128		dB dB dB dB
FREQUENCY RESPONSE Gain-Bandwidth Product Slew Rate Overload Recovery Time	GBW SR	G = 1 V _{IN} • G = V _S		* * *			35 0.01 60		kHz V/μs μs
OUTPUT Voltage Output Swing from Rail ⁽²⁾ $T_A = -40^{\circ}C$ to $+85^{\circ}C$	Vo	$\begin{aligned} R_L &= 100k\Omega, \ A_{OL} \geq 70dB \\ R_L &= 100k\Omega, \ A_{OL} \geq 100dB \\ R_L &= 100k\Omega, \ A_{OL} \geq 100dB \\ R_L &= 20k\Omega, \ A_{OL} \geq 100dB \end{aligned}$		* *			50 75 100	250 250 300	mV mV mV
T _A = -40°C to +85°C Short-Circuit Current Single Versions Dual Versions Capacitive Load Drive	I _{sc}	$R_L = 20k\Omega$, $A_{OL} \ge 100dB$		* * *			-21/+4 -50/+4 Typical C	300	mV mA mA
POWER SUPPLY Specified Voltage Range Operating Voltage Range Quiescent Current (per amplifier) T _A = -40°C to +85°C	C _{LOAD} V _S	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$ $I_O = 0$ $I_O = 0$	*	*	*	±1.35	±15 ±27	±18 ±38 ±45	V V μΑ μΑ
TEMPERATURE RANGE Specified Range Operating Range Storage Range Thermal Resistance	$ heta_{\sf JA}$		* *		* *	-40 -55 -55	1.5-	+85 +125 +125	°C °C °C
8-Pin DIP SO-8 Surface Mount 14-Pin DIP SO-14 Surface Mount				* * *			100 150 80 100		°C/W °C/W °C/W

^{*} Specifications the same as OPA251UA, PA.

NOTES: (1) The negative sign indicates input bias current flows out of the input terminals. (2) Output voltage swings are measured between the output and power supply rails.

ABSOLUTE MAXIMUM RATINGS(1)

Supply Voltage, V+ to V	36V
Input Voltage ⁽²⁾	
Output Short Circuit to Ground(3)	Continuous
Operating Temperature	55°C to +125°C
Storage Temperature	55°C to +125°C
Junction Temperature	150°C
Lead Temperature (soldering, 10s)	300°C

NOTES: (1) Stresses above these ratings may cause permanent damage. (2) Input terminals are diode-clamped to the power supply rails. Input signals that can swing more that 0.5V beyond the supply rails should be currentlimited to 5mA or less. (3) One amplifier per package.

ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

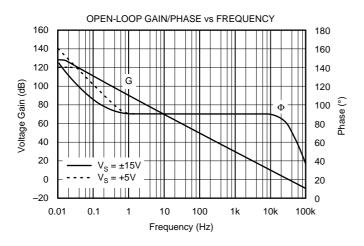
PACKAGE/ORDERING INFORMATION

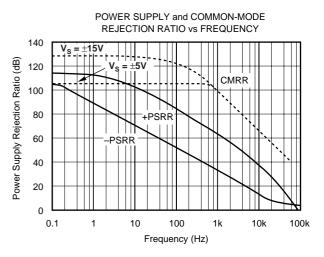
PRODUCT	SPECIFIED VOLTAGE	OPERATING VOLTAGE RANGE	PACKAGE	PACKAGE DRAWING NUMBER ⁽¹⁾	SPECIFICATION TEMPERATURE RANGE
OPA241 SERIES					
Single OPA241PA OPA241UA	2.7V to 5V 2.7V to 5V	2.7V to 36V 2.7V to 36V	8-Pin DIP SO-8 Surface Mount	006 182	-40°C to +85°C -40°C to +85°C
Dual OPA2241PA OPA2241UA	2.7V to 5V 2.7V to 5V	2.7V to 36V 2.7V to 36V	8-Pin DIP SO-8 Surface Mount	006 1 <mark>8</mark> 2	-40°C to +85°C -40°C to +85°C
Quad OPA4241PA OPA4241UA	2.7V to 5V 2.7V to 5V	2.7V to 36V 2.7V to 36V	14-Pin DIP SO-14 Surface Mount	010 235	-40°C to +85°C -40°C to +85°C
OPA251 SERIES					
Single OPA251PA OPA251UA	±15V ±15V	2.7V to 36V 2.7V to 36V	8-Pin DIP SO-8 Surface Mount	006 182	-40°C to +85°C -40°C to +85°C
Dual OPA2251PA OPA2251UA	±15V ±15V	2.7V to 36V 2.7V to 36V	8-Pin DIP SO-8 Surface Mount	006 182	-40°C to +85°C -40°C to +85°C
Quad OPA4251PA OPA4251UA	±15V ±15V	2.7V to 36V 2.7V to 36V	14-Pin DIP SO-14 Surface Mount	010 235	-40°C to +85°C -40°C to +85°C

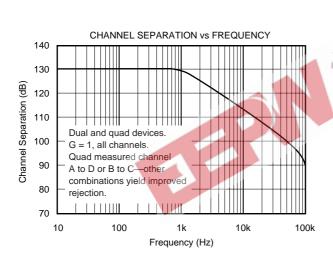
NOTE: (1) For detailed drawing and dimension table, please see end of data sheet, or Appendix C of Burr-Brown IC Data Book.

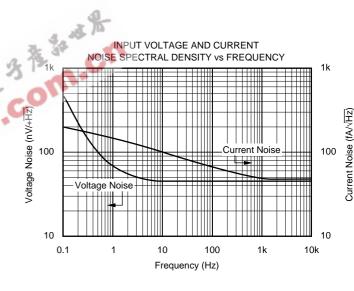
TYPICAL PERFORMANCE CURVES

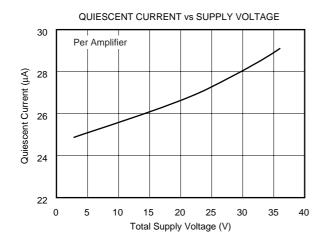
At $T_A = +25^{\circ}C$, and $R_L = 100k\Omega$ connected to $V_S/2$ (ground for $V_S = \pm 15V$), unless otherwise noted.

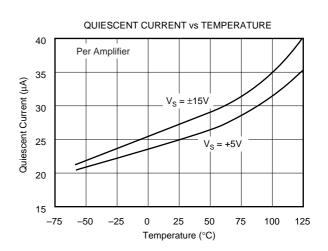






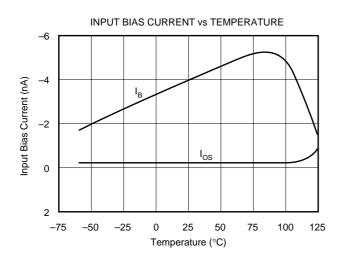


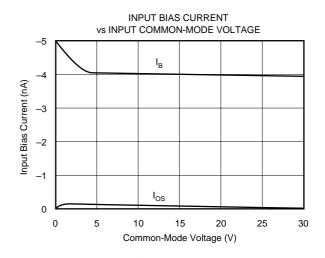


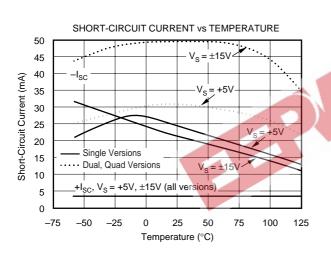


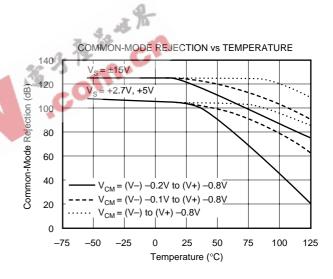
TYPICAL PERFORMANCE CURVES (CONT)

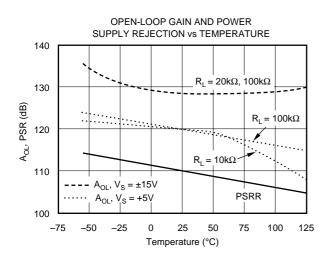
At T_A = +25°C, and R_L = 100k Ω connected to $V_S/2$ (ground for V_S = ±15V), unless otherwise noted.

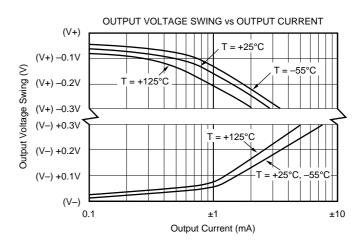






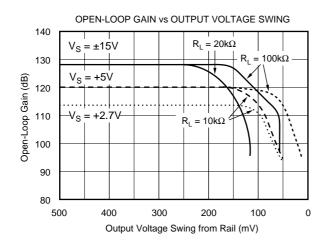


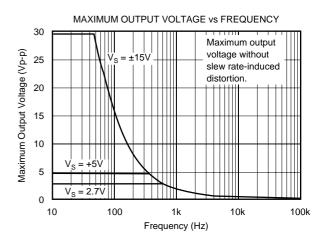


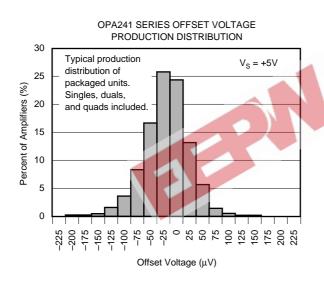


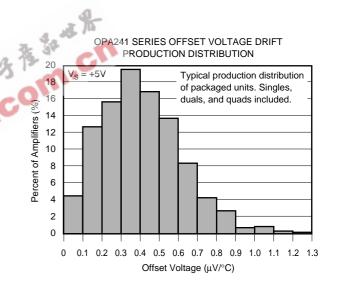
TYPICAL PERFORMANCE CURVES (CONT)

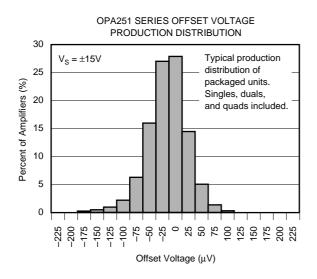
At T_A = +25°C, and R_L = 100k Ω connected to $V_S/2$ (ground for V_S = ±15V), unless otherwise noted.

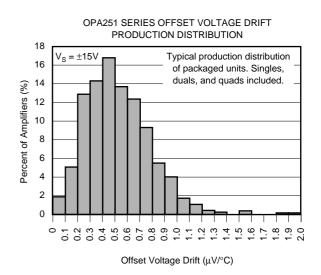






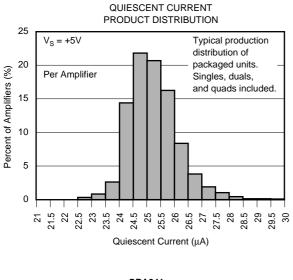


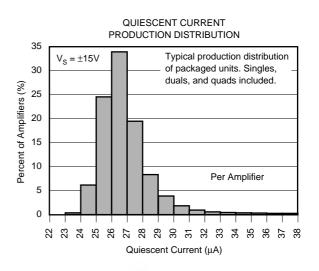


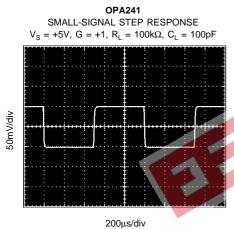


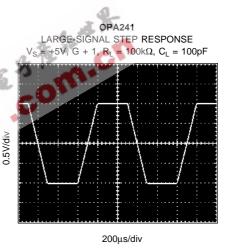
TYPICAL PERFORMANCE CURVES (CONT)

At T_A = +25°C, and R_L = 100k Ω connected to $V_S/2$ (ground for V_S ±15V), unless otherwise noted.

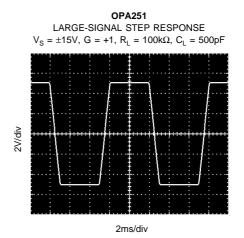








OPA251 SMALL-SIGNAL STEP RESPONSE $V_S=\pm 15V,\,G=+1,\,R_L=100 k\Omega,\,C_L=500 pF$



APPLICATIONS INFORMATION

The OPA241 and OPA251 series are unity-gain stable and suitable for a wide range of general purpose applications. Power supply pins should be bypassed with $0.01\mu F$ ceramic capacitors.

OPERATING VOLTAGE

The OPA241 series is laser-trimmed for low offset voltage and drift at low supply voltage ($V_S = +5V$). The OPA251 series is trimmed for $\pm 15V$ operation. Both products operate over the full voltage range (+2.7V to +36V or $\pm 1.35V$ to $\pm 18V$) with some compromises in offset voltage and drift performance. However, all other parameters have similar performance. Key parameters are guaranteed over the specified temperature range, $-40^{\circ}C$ to $+85^{\circ}C$. Most behavior remains unchanged throughout the full operating voltage range. Parameters which vary significantly with operating voltage or temperature are shown in typical performance curves.

OFFSET VOLTAGE TRIM

As mentioned previously, offset voltage of the OPA241 series is laser-trimmed at +5V. The OPA251 series is trimmed at ±15V. Because the initial offset is so low, user adjustment is usually not required. However, the OPA241 and OPA251 (single op amp versions) provide offset voltage trim connections on pins 1 and 5. Offset voltage can be adjusted by connecting a potentiometer as shown in Figure 1. This adjustment should be used only to null the offset of the op amp, not to adjust system offset or offset produced by the signal source. Nulling offset could degrade the offset drift behavior of the op amp. While it is not possible to predict the exact change in drift, the effect is usually small.

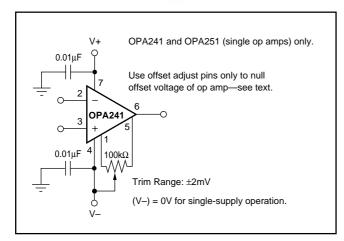


FIGURE 1. OPA241 and OPA251 Offset Voltage Trim Circuit.

CAPACITIVE LOAD AND STABILITY

The OPA241 series and OPA251 series can drive a wide range of capacitive loads. However, all op amps under certain conditions may be unstable. Op amp configuration, gain, and load value are just a few of the factors to consider when determining stability.

Figures 2 and 3 show the regions where the OPA241 series and OPA251 series have the potential for instability. As shown, the unity gain configuration with low supplies is the most susceptible to the effects of capacitive load. With $V_S = \pm 5V$, $G = \pm 1$, and $I_{OUT} = 0$, operation remains stable with load capacitance up to approximately 200pF. Increasing supply voltage, output current, and/or gain significantly improves capacitive load drive. For example, increasing the supplies to $\pm 15V$ and gain to 10 allows approximately 2700pF to be driven.

One method of improving capacitive load drive in the unity gain configuration is to insert a resistor inside the feedback loop as shown in Figure 4. This reduces ringing with large capacitive loads while maintaining dc accuracy. For example, with $V_S=\pm 1.35 V$ and $R_S=5 k\Omega$, the OPA241 series and OPA251 series perform well with capacitive loads in excess of 1000pF. Without the series resistor, capacitive load drive is typically 200pF for these conditions. However, this method will result in a slight reduction of output voltage swing.

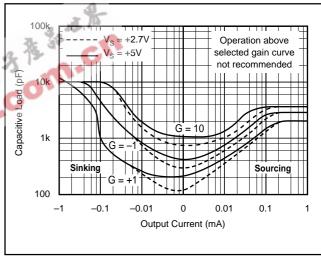


FIGURE 2. Stability—Capacitive Load versus Output Current for Low Supply Voltage.

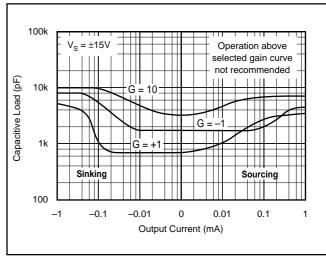


FIGURE 3. Stability—Capacitive Load versus Output Current for ± 15 V Supplies.



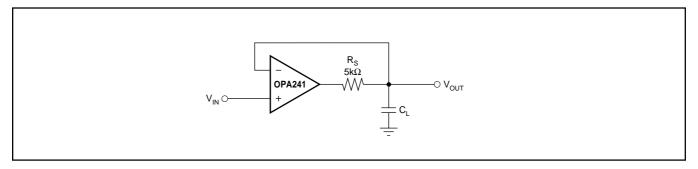


FIGURE 4. Series Resistor in Unity Gain Configuration Improves Capacitive Load Drive.

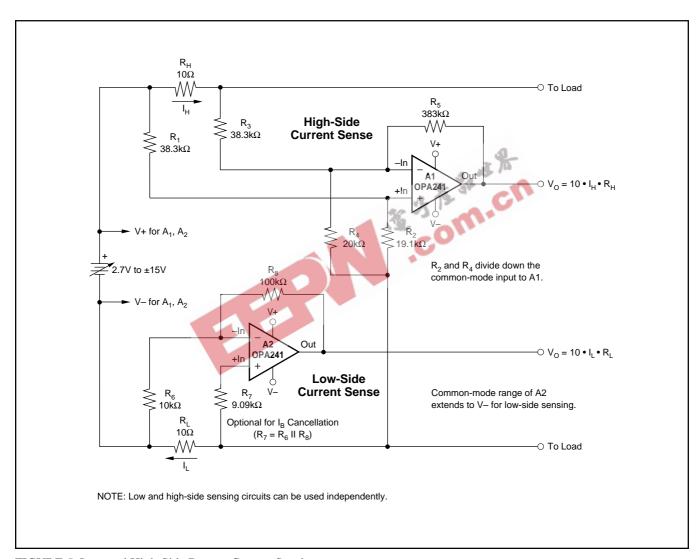


FIGURE 5. Low and High-Side Battery Current Sensing.