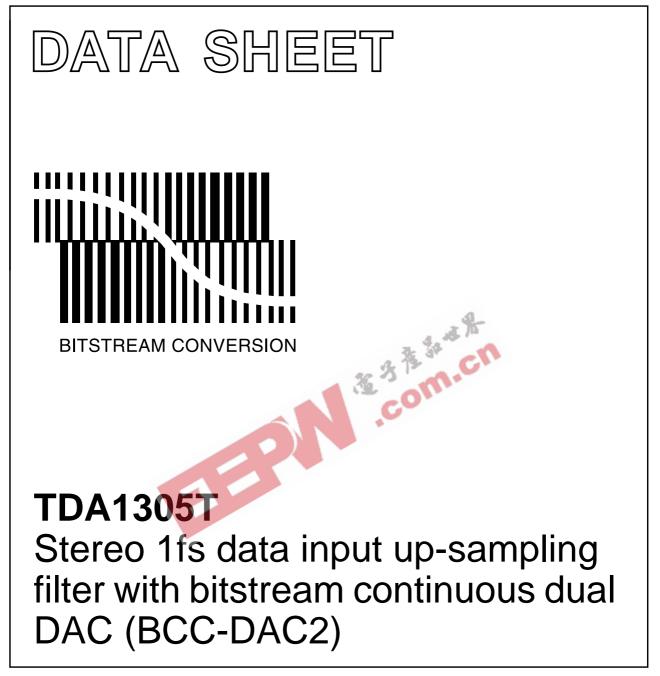
## INTEGRATED CIRCUITS



Preliminary specification Supersedes data of September 1994 File under Integrated Circuits, IC01 1995 Dec 08



### TDA1305T

#### FEATURES

- · Easy application
- 16f<sub>s</sub> Finite-duration Impulse-Response (FIR) filter incorporated
- Selectable system clock (f<sub>sys</sub>) 256f<sub>s</sub> or 384f<sub>s</sub>
- I<sup>2</sup>S-bus serial input format (at f<sub>sys</sub> = 256f<sub>s</sub>) or LSB fixed 16, 18 or 20 bits serial input mode (at f<sub>sys</sub> = 384f<sub>s</sub>)
- Slave-mode clock system
- Cascaded 4-stage digital filter incorporating 2-stage FIR filter, linear interpolator and sample-and-hold
- Smoothed transitions before and after muting (soft mute)
- Digital de-emphasis filter for three sampling rates of 32 kHz, 44.1 kHz and 48 kHz
- 12 dB attenuation via the attenuation input control
- Double speed mode
- 2nd order noise shaper
- 96 (f<sub>sys</sub> = 384f<sub>s</sub>) or 128 (f<sub>sys</sub> = 256f<sub>s</sub>) times oversampling in normal speed mode
- 48 (f<sub>sys</sub> = 384f<sub>s</sub>) or 64 (f<sub>sys</sub> = 256f<sub>s</sub>) times oversampling in double speed mode
- · Bitstream continuous calibration concept
- Small outline SO28 package
- Voltage output 1.5 V (RMS) at line drive level
- Low total harmonic distortion
- · No zero crossing distortion
- · Inherently monotonic
- No analog post filtering required
- Superior signal-to-noise ratio
- Wide dynamic range (18-bit)
- Single rail supply (3.4 to 5.5 V).

#### **GENERAL DESCRIPTION**

The TDA1305T is a new generation of filter-DAC which features a unique combination of bitstream and continuous calibration techniques. The converter functions as a



bitstream converter for low signals while large signals are generated using the dynamic continuous calibration technique, thus resulting in low power consumption, small chip size and easy application.

The TDA1305T is a dual CMOS DAC with up-sampling filter and noise shaper. The combination of high oversampling up to  $16f_s$ , 2nd order noise shaping and continuous calibration conversion ensures that only simple 1st order analog post filtering is required.

The TDA1305T supports the l<sup>2</sup>S-bus data input mode with word lengths of up to 20 bits (at  $f_{sys} = 256f_s$ ) and the LSB fixed serial data input format with word lengths of 16, 18 and 20 bits (at  $f_{sys} = 384f_s$ ). Four cascaded FIR filters increase the oversampling rate to 16 times. A sample-and-hold function increases the oversampling rate to 96 times ( $f_{sys} = 384f_s$ ) or 128 times ( $f_{sys} = 256f_s$ ). A 2nd order noise shaper converts this oversampled data to a bitstream for the 5-bit DACs.

The DACs are of the continuous calibration type and incorporate a special date coding. This ensures an extremely high signal-to-noise ratio, superior dynamic range and immunity to process variation and component ageing.

Two on-board operational amplifiers convert the digital-to-analog current to an output voltage. Externally connected capacitors perform the required 1st order filtering so that no further post filtering is required.

The unique combination of bitstream and continuous calibration techniques, together with a high degree of analog and digital integration, results in a single filter-DAC with 18-bit dynamic range, high linearity and simple low cost application.

#### **ORDERING INFORMATION**

		PACKAGE				
	NAME	DESCRIPTION	VERSION			
TDA1305T	SO28	plastic small outline package; 28 leads; body width 7.5 mm	SOT136-1			

### TDA1305T

#### QUICK REFERENCE DATA

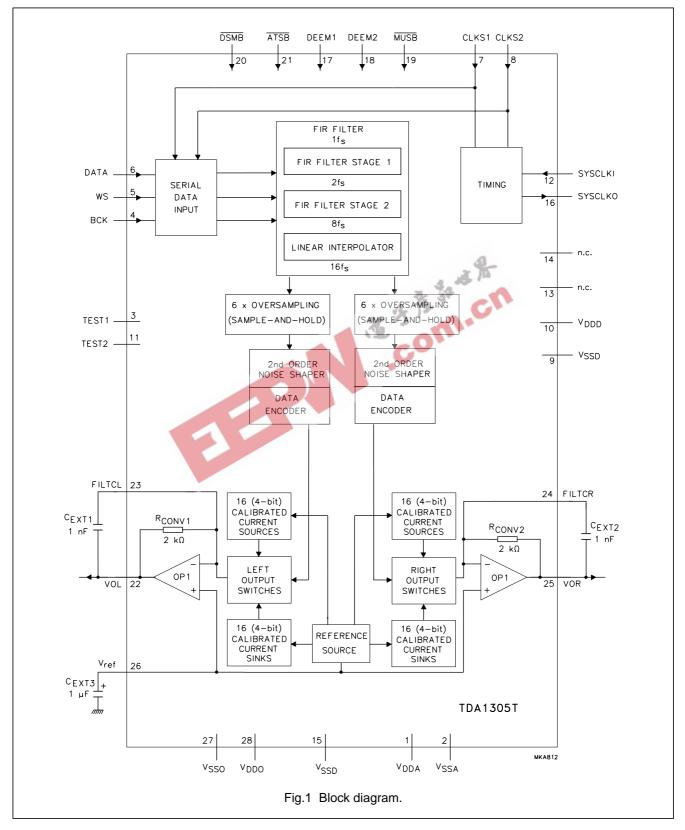
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>DDD</sub>	digital supply voltage	note 1	3.4	5.0	5.5	V
V <sub>DDA</sub>	analog supply voltage	note 1	3.4	5.0	5.5	V
V <sub>DDO</sub>	operational amplifier supply voltage	note 1	3.4	5.0	5.5	V
I <sub>DDD</sub>	digital supply current	V <sub>DDD</sub> = 5 V; at code 00000H	_	30	_	mA
I <sub>DDA</sub>	analog supply current	V <sub>DDA</sub> = 5 V; at code 00000H	-	5.5	8	mA
I <sub>DDO</sub>	operating amplifier supply current	V <sub>DDO</sub> = 5 V; at code 00000H	_	6.5	9	mA
V <sub>FS(rms)</sub>	full-scale output voltage (RMS value)	$V_{DDD} = V_{DDA} = V_{DDO} = 5 V$	1.425	1.5	1.575	V
(THD + N)/S	total harmonic distortion	at 0 dB signal level	-	-90	-81	dB
	plus noise-to-signal ratio		- A 14	0.003	0.009	%
		at –60 dB signal level	7. 34	-44	-40	dB
		30 33	P	0.63	0.1	%
		at –60 dB sign <mark>al level;</mark>	- 11-	-46	-	dB
		A-weighted	-	0.5	-	%
S/N	signal-to-noise ratio at bipolar zero	A-weighting; at code 00000H	100	108	_	dB
BR <sub>ns</sub>	input bit rate at data input	f <sub>s</sub> = 48 kHz; normal speed	-	-	3.072	Mbits
BR <sub>ds</sub>	input bit rate at data input	f <sub>s</sub> = 48 kHz; double speed	_	_	6.144	Mbits
f <sub>sys</sub>	system clock frequency		6.4	-	18.432	MHz
TC <sub>FS</sub>	full scale temperature coefficient at analog outputs (VOL and VOR)		-	±100 × 10 <sup>-6</sup>	-	
T <sub>amb</sub>	operating ambient temperature		-30	_	+85	°C

#### Note

1. All  $V_{\text{DD}}$  and  $V_{\text{SS}}$  pins must be connected to the same supply.

## TDA1305T

#### **BLOCK DIAGRAM**



#### PINNING

SYMBOL	PIN	DESCRIPTION			
V <sub>DDA</sub>	1	analog supply voltage			
V <sub>SSA</sub>	2	analog ground			
TEST1	3	test input; pin should be connected to ground (internal pull-down resistor)			
BCK	4	bit clock input			
WS	5	word select input			7
DATA	6	data input	V <sub>DDA</sub> 1	U	28 V <sub>DDO</sub>
CLKS1	7	clock selection 1 input			
CLKS2	8	clock selection 2 input	V <sub>SSA</sub> 2		27 V <sub>SSO</sub>
V <sub>SSD</sub>	9	digital ground	TEST1 3		26 V <sub>ref</sub>
V <sub>DDD</sub>	10	digital supply voltage		-	
TEST2	11	test input; pin should be connected to ground (internal pull-down resistor)	BCK 4 WS 5	St.	25 VOR 24 FILTCR
SYSCLKI	12	system clock input	DATA 6		23 FILTCL
n.c.	13	not connected (this pin should be left open-circuit)	CLKS1 7	TDA1305T	22 VOL
n.c.	14	not connected (this pin should be left open-circuit)	CLKS2 8 V <sub>SSD</sub> 9		21 ATSB 20 DSMB
V <sub>SSD</sub>	15	digital ground			
SYSCLKO	16	system clock output	VDDD 10		19 MUSB
DEEM1	17	de-emphasis on/off; f <sub>DEEM</sub> 32 kHz, 44 kHz and 48 kHz	TEST2 11		18 DEEM2
DEEM2	18	de-emphasis on/off; f <sub>DEEM</sub> 32 kHz, 44 kHz and 48 kHz	SYSCLKI 12 n.c. 13		17 DEEM1 16 SYSCLKO
MUSB	19	mute input (active LOW)			
DSMB	20	double-speed mode input (active LOW)	n.c. 14		15 V <sub>SSD</sub>
ATSB	21	12 dB attenuation input (active LOW)		МКА813	3
VOL	22	left channel output			
FILTCL	23	capacitor for left channel 1st order filter function should be connected between pins 22 and 23			
FILTCR	24	capacitor for right channel 1st order filter function should be connected between pins 25 and 24			
VOR	25	right channel output			
V <sub>ref</sub>	26	internal reference voltage for output channels (0.5V <sub>DD</sub> )			
V <sub>SSO</sub>	27	operational amplifier ground	Fig 2	2 Pin configura	ation
V <sub>DDO</sub>	28	operational amplifier supply voltage	1 19.2		

## TDA1305T

#### FUNCTIONAL DESCRIPTION

The TDA1305T CMOS digital-to-analog bitstream converter incorporates an up-sampling filter and noise shaper which increase the oversampling rate of 1fs input data to  $96f_s$  ( $f_{sys} = 192f_s$ ) or  $128f_s$  ( $f_{sys} = 256f_s$ ) in the normal speed mode. In the double speed mode the oversample rate of 1fs input data is increased to 48fs  $(f_{sys} = 384f_s)$  or  $64f_s$   $(f_{sys} = 256f_s)$ . This oversampling, together with the 5-bit DAC, enables the filtering required for waveform smoothing and out-of-band noise reduction to be achieved by simple 1st order analog post filtering.

#### System clock and data input format

The TDA1305T accommodates slave mode only, this means that in all applications the system devices must provide a system clock of 256 or  $384f_s$  ( $f_s = 32, 44.1$  or 48 kHz). The system frequency is selectable by means of pin CLKS1 and pin CLKS2. The SYSCLKO output (pin 16) provides the system clock for external use.

The TDA1305T supports the following data input modes:

- I<sup>2</sup>S-bus with data word lengths of up to 20 bits  $(at f_{sys} = 256 f_s).$
- LSB fixed serial format with data word lengths of 16, 18 and 20 bits (at  $f_{sys} = 384f_s$ ). As this format idles on the MSB it is necessary to know how many bits are being transmitted.

The input format is shown in Fig.3. Left and right data-channel words are time-multiplexed.

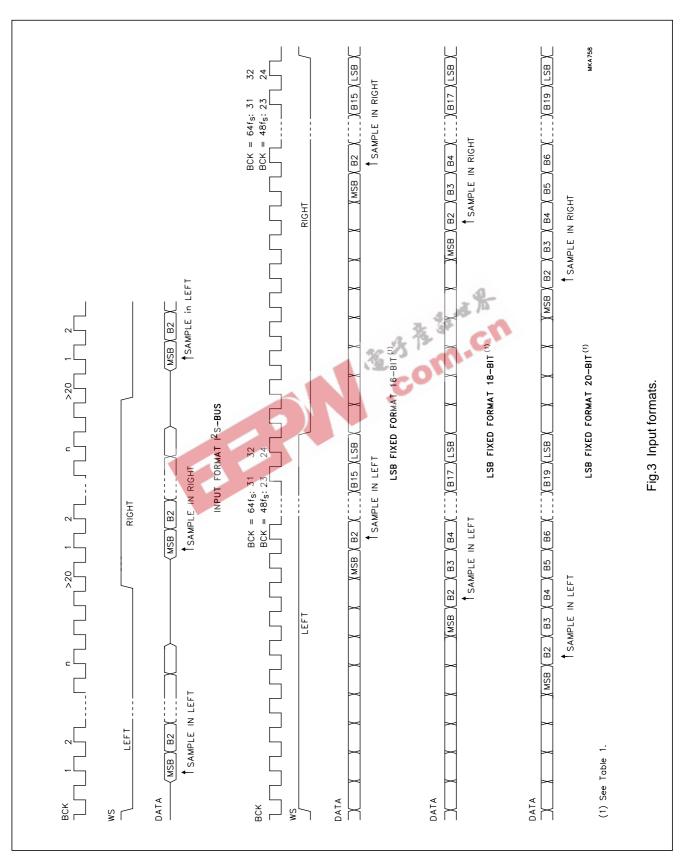
able 1 D	ata input for	mat and sys		8-		
TEST1	CLKS1	CLKS2	DATA INPUT FORMAT	SYSTEM CLOCK		SYSCLKO
0	0	0	I <sup>2</sup> S up to 20 bits	256fs	>20	256f <sub>s</sub>
0	0	1	LSB fixed 16 bits	384f <sub>s</sub>	24	384f <sub>s</sub>
0	1	0	LSB fixed 18 bits	384f <sub>s</sub>	24	384f <sub>s</sub>
0	1	1	LSB fixed 20 bits	384f <sub>s</sub>	24	384f <sub>s</sub>
1	0	0	reserved	-	_	_
1	0	1	LSB fixed 16 bits	384f <sub>s</sub>	32	384f <sub>s</sub>
1	1	0	LSB fixed 18 bits	384f <sub>s</sub>	32	384f <sub>s</sub>
1	1	1	LSB fixed 20 bits	384f <sub>s</sub>	32	384f <sub>s</sub>

### Table 1 Data input format and ovatam alaak

#### Note

1. Number of clock pulses within half an audio sample.

### TDA1305T



## TDA1305T

#### Mute

Soft mute is controlled by the  $\overline{\text{MUSB}}$  at pin 9. When the input is active LOW the value of the samples is decreased smoothly to zero following a cosine curve. To step down the value of the data 32 coefficients are used, each one being used 31 times before stepping onto the next. When MUTE is released (pin 19 = HIGH), the samples are returned to the full level again following a cosine curve with the same coefficients being used in the reverse order. Mute is synchronized to prevent operation in the middle of a word.

#### **De-emphasis**

A digital de-emphasis is implemented for three sample rates (32, 44.1 and 48 kHz). By selecting DEEM1 and DEEM2 de-emphasis can be applied by means of a FIR filter. Time constants of the de-emphasis are 50  $\mu$ s and 15  $\mu$ s. De-emphasis is synchronized to prevent operation in the middle of a word. The de-emphasis deviation from ideal 50  $\mu$ s and 15  $\mu$ s de-emphasis is given in Table 4.

#### Table 2 De-emphasis.

DEEM1	DEEM2	CONDITION
0	0	de-emphasis disabled
0	1	de-emphasis for $f_s = 32 \text{ kHz}$
1	0	de-emphasis for $f_s = 4.1$ kHz
1	1	de-emphasis for f <sub>s</sub> = 48 kHz

#### Attenuation

Attenuation is controlled by the  $\overline{\text{ATSB}}$  input (pin 21). When the input is active LOW the sample is multiplied by a coefficient that provides 12 dB attenuation. If the input is HIGH the multiplication factor is 1. Attenuation is synchronized to prevent operation in the middle of a word.

#### **Double-speed mode**

Double speed is controlled by the DSMB input (pin 20). When the input is active LOW the device operates in the double-speed mode.

#### **Oversampling filter (normal-speed mode)**

In the normal-speed mode the oversampling filter consists of:

- A 91st order half-band low-pass FIR filter which increases the oversampling rate from 1 time to 2 times.
- A 23rd order quarter band low-pass FIR filter which increases the oversampling rate from 2 times to 8 times.
- A linear interpolation section which increases the oversampling rate to 16 times. This removes the spectral components around 8f<sub>s</sub>.
- A sample-and-hold section which provides another 6 times oversampling to 96 times. The zero-order hold characteristic of this sample-and-hold section plus the 1st order analog filtering remove the spectral components around 16f<sub>s</sub>.

Pass-band ripple and stop-band attenuation for normal-speed are given in Table 3.

#### Oversampling filter (double-speed mode)

In the double-speed mode the oversampling filter consists of:

- A 51st order half-band low-pass FIR filter which increases the oversampling rate from 1 time to 2 times.
- A 7th order half-band low-pass FIR filter which increases the oversampling rate from 2 times to 4 times.
- A linear interpolation section which increases the oversampling rate to 8 times. This removes the spectral components around 4f<sub>s</sub>.
- A sample-and-hold section which provides another 6 times oversampling to 48 times. The zero-order hold characteristic of this sample-and-hold section plus the 1st order analog filtering remove the spectral components around 8f<sub>s</sub>.

Pass-band ripple and stop-band attenuation for double-speed are given in Table 3.

### TDA1305T

### Noise shaper

In the normal speed mode the 2nd order digital noise shaper operates at 96f<sub>s</sub> ( $f_{sys} = 384f_s$ ) or  $128f_s$  ( $f_{sys} = 256f_s$ ). The digital noise shaper operates at  $48f_s$  ( $f_{sys} = 384f_s$ ) or  $64f_s$  ( $f_{sys} = 256f_s$ ) in double-speed mode. It shifts in-band quantization noise to frequencies well above the audio band. This noise shaping technique used in combination with a special data coding enables extremely high signal-to-noise ratios to be achieved. The noise shaper outputs a 5-bit pulse duration modulation (PDM) bitstream signal to the DAC.

#### Continuous calibration DAC

The dual 5-bit DAC uses the continuous calibration technique. This method, based on charge storage, involves exact duplication of a single reference current source. In the TDA1305T, 32 such current sources plus 1 spare source are continuously calibrated. The spare source is included to allow continuous converter operation.

The DAC receives a 5-bit data bitstream from the noise shaper. This data is then converted so that only small currents are switched to the output during digital silence (input 00000H). Using this technique extremely high signal-to-noise performance is achieved.

#### **Operational amplifiers**

High precision, low-noise amplifiers together with the internal conversion resistors  $R_{CONV1}$  and  $R_{CONV2}$  convert the converter output current to a voltage capable of driving a line output. This voltage is available at VOL and VOR (1.5 V RMS typical).

Connecting external capacitors CEXT1 and CEXT2 between FILTCL and VOL and between FILTCR and VOR respectively provides the required 1st order post filtering for the left and right channels (see Fig.1). The combinations of  $R_{CONV1}$  with CEXT1 and  $R_{CONV2}$  with CEXT2 determine the 1st order fall-off frequencies.

#### Internal reference circuitry

Internal reference circuitry ensures that the output voltage signal is proportional to the supply voltage, thereby maintaining maximum dynamic range for supply voltages from 3.4 to 5.5 V and making the circuit also suitable for battery-powered applications.

#### LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V <sub>DDD</sub>	digital supply voltage		-	7.0	V
V <sub>DDA</sub>	analog supply voltage		-	7.0	V
V <sub>DDO</sub>	operational amplifier supply voltage		-	7.0	V
T <sub>xtal</sub>	maximum crystal temperature		-	+150	°C
T <sub>stg</sub>	storage temperature		-65	+150	°C
T <sub>amb</sub>	ambient operating temperature		-30	+85	°C
V <sub>es</sub>	electrostatic handling	note 1	-2000	+2000	V
		note 2	-200	+200	V

#### Notes

1. Human body model; C = 100 pF, R = 1500  $\Omega$ , V = 2000 V, 3 pulses positive and 3 pulses negative.

2. Machine model; C = 200 pF, R = 10  $\Omega$ , L = 0.5  $\mu$ H.

#### THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	VALUE	UNIT
R <sub>th j-a</sub>	thermal resistance from junction to ambient in free air	75	K/W

Stereo 1fs data input up-sampling filter with	
bitstream continuous dual DAC (BCC-DAC2)	

#### QUALITY SPECIFICATION

In accordance with "SNW-FQ-611E". The number of this quality specification can be found in the "Quality Reference Handbook". The handbook can be ordered using the code 9398 510 63011.

#### DIGITAL CHARACTERISTICS

 $V_{DD}$  = 3.4 to 5.5 V;  $V_{SS}$  = 0 V;  $T_{amb}$  = -40 to +85 °C; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply	1		•	•		•
V <sub>DDD</sub>	digital supply voltage	note 1	3.4	5.0	5.5	V
I <sub>DDD</sub>	digital supply current	V <sub>DDD</sub> = 5 V; at code 00000H	_	30	40	mA
V <sub>DDA</sub>	analog supply voltage	note 1	3.4	5.0	5.5	V
I <sub>DDA</sub>	analog supply current	V <sub>DDA</sub> = 5 V; at code 00000H	-	5.5	8	mA
V <sub>DDO</sub>	operational amplifier supply voltage	note 1	3.4	5.0	5.5	V
I <sub>DDO</sub>	operational amplifier supply current	$V_{DDO} = 5 V;$ at code 00000H	372	6.5	9	mA
RR	ripple rejection to V <sub>DDA</sub>	note 2	-0'	25	-	dB
System clo	ck input			•		
f <sub>sys</sub>	system frequency	$f_{sys} = 384 f_s$	9.6	16.93	18.4	MHz
		$f_{sys} = 256 f_s$	6.4	11.29	12.28	MHz
V <sub>IL</sub>	LOW level input voltage	note 3	-0.5	-	0.2V <sub>DD</sub>	V
V <sub>IH</sub>	HIGH level input voltage	note 3	0.8V <sub>DD</sub>	-	V <sub>DD</sub> + 0.5	V
I <sub>LI</sub>	input leakage current	note 4	_	-	10	μA
C <sub>i</sub>	input capacitance		_	-	10	pF
T <sub>cy</sub>	clock cycle time	$f_{sys} = 384 f_s$	104	59.1	54.2	ns
		$f_{sys} = 256 f_s$	156	88.6	81.3	ns
Digital inpu	ts; WS, BCK, DATA, <mark>DSMB</mark> , I	MUSB, DEEM1, DEEM2	, ATSB, CLK	S1, CLKS2,	TEST1 and	TEST2
V <sub>IL</sub>	LOW level input voltage	note 3	-0.5	-	0.3V <sub>DD</sub>	V
V <sub>IH</sub>	HIGH level input voltage	note 3	0.7V <sub>DD</sub>	-	V <sub>DD</sub> + 0.5	V
I <sub>LI</sub>	input leakage current	note 4	-	-	10	μA
Ci	input capacitance		-	-	10	pF
Digital outp	out; CDEC					•
V <sub>OL</sub>	LOW level output voltage	I <sub>OL</sub> = 0.4 mA	0	-	0.5	V
V <sub>OH</sub>	HIGH level output voltage	I <sub>OH</sub> = -0.2 mA	V <sub>DD</sub> - 0.5	_	V <sub>DD</sub>	V
t <sub>r</sub>	output rise time	note 5	_	-	20	ns
t <sub>f</sub>	output fall time	note 5	_	_	20	ns
CL	load capacitance		_	_	30	pF

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Serial input	t data timing (see Fig.4)					
f <sub>BCK</sub>	bit-clock input (data input	$f_{sys} = 384 f_s$	_	48fs	_	MHz
	rate) frequency	$f_{sys} = 256 f_s$	-	64f <sub>s</sub>	_	MHz
f <sub>WS</sub>	word select input frequency	normal speed	25	44.1	48	kHz
		double speed	50	88.2	96	kHz
t <sub>r</sub>	rise time		-	_	20	ns
t <sub>f</sub>	fall time		-	_	20	ns
t <sub>H</sub>	bit clock time HIGH		55	_	-	ns
tL	bit clock time LOW		55	_	_	ns
t <sub>su</sub>	data set-up time		40	-	-	ns
t <sub>h</sub>	data hold time		10	_	-	ns
t <sub>suWS</sub>	word select set-up time		40	-	-	ns
t <sub>hWS</sub>	word select hold time		10	4	_	ns

- All V<sub>DD</sub> and V<sub>SS</sub> pins must be connected externally to the same supply.
   V<sub>ripple</sub> = 1% of supply voltage: from the same supply. 2.  $V_{ripple} = 1\%$  of supply voltage;  $f_{ripple} = 100$  Hz. Ripple rejection RR to  $V_{DDA}$  is dependent on the value of the external capacitor ( $C_{EXT3}$  in Fig.1) connected to  $V_{ref}$ . The value here assumes that  $C_{EXT3} = 1 \ \mu$ F.
- 3. Minimum VIL and maximum VIH are peak values to allow for transients.
- 4.  $I_{Llmni}$  measured at  $V_1 = 0$  V;  $I_{Llmax}$  measured at  $V_1 = 5.5$  V.
- 5. Reference levels = 10% and 90%.

#### ANALOG CHARACTERISTICS

 $V_{DD} = V_{DDA} = V_{DDO} = 5 \text{ V}; V_{SS} = 0 \text{ V}; T_{amb} = 25 \text{ °C}; unless otherwise specified.}$ 

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Reference v	alues			•		•
V <sub>ref</sub>	reference voltage level		2.45	2.5	2.55	V
R <sub>CONV</sub>	current-to-voltage conversion resistor		1.6	2.2	2.8	kΩ
Analog outp	outs					
RES	resolution		_	-	18	bit
V <sub>FS(rms)</sub>	full-scale output voltage (pins 23 and 25) (RMS value)		1.425	1.5	1.575	V
V <sub>OFF</sub>	output voltage DC offset with respect to reference voltage level V <sub>ref</sub>		-80	-65	-50	mV
TC <sub>FS</sub>	full scale temperature coefficient		-	$\pm 100 \times 10^{-6}$	-	

### TDA1305T

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
(THD + N)/S	total harmonic distortion plus	at 0 dB input level;	-	-90	-81	dB
	noise-to-signal ratio	note 1	-	0.003	0.009	%
		at –60 dB input level;	-	-44	-40	dB
		note 2	-	0.63	1.0	%
		at –60 dB input level;	-	-46	-	dB
		A-weighted; note 3	_	0.5	_	%
		at 0 dB input level; (20 Hz to 20 kHz); note 4	-	-90	-81	dB
			-	0.003	0.003	%
S/N	signal-to-noise ratio at bipolar zero	A weighted; at code (00000H)	100	108	-	dB
$\alpha_{cs}$	channel separation		85	100	_	dB
δVo	unbalance between outputs		-	0.2	0.3	dB
z <sub>o</sub>	dynamic output impedance		-	10	_	Ω
RL	output load resistance		3	5-1"	-	kΩ
CL	output load capacitance	4	-12 -		200	pF

#### Notes

- 1. Measured with a 1 kHz, 0 dB, 18-bit sine wave generated at a sampling rate of 48 kHz. The (THD + N)/S measured over a bandwidth of 20 Hz to 20 kHz.
- 2. Measured with a 1 kHz, -60 dB, 18-bit sine wave generated at a sampling rate of 48 kHz. The (THD + N)/S measured over a bandwidth of 20 Hz to 20 kHz. For 16-bit input signals, the performance is limited to the theoretical maximum.
- 3. Measured with a 1 kHz, -60 dB, 18-bit sine wave generated at a sampling rate of 48 kHz. The (THD + N)/S measured over a bandwidth of 20 Hz to 20 kHz and filtered with a A-weighted characteristic. For 16-bit input signals, the performance is limited to the theoretical maximum.
- 4. Measured with a sine wave from 20 Hz to 20 kHz generated at a sampling rate of 48 kHz. The (THD + N)/S measured over a bandwidth of 20 Hz to 20 kHz.

#### **TEST AND APPLICATION INFORMATION**

#### Filter characteristics (theoretical values)

Table 3 Normal speed filter characteristics.

ITEM	SAMPLE FREQUENCY	RANGE	CONDITIONS	CHARACTERISTICS
Pass band	44.1 kHz	0 to 20 kHz		0 ±0.025 dB
	32 kHz	14.5 to 15 kHz		–0.15 dB (min.)
Stop band	44.1 kHz	24.1 to 150 kHz	typical	-60 dB (max.)
			worst case	–57 dB (max.)
		150 kHz to infinity	typical	–57 dB (max.)
			worst case	–47 dB (max.)
	32 kHz	17 to 17.5 kHz		-40 dB (max.)

### **TDA1305T**

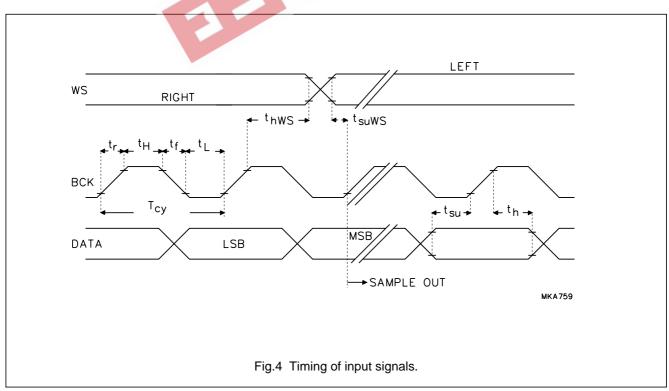
#### De-emphasis filter characteristics (theoretical values)

Table 4 De-emphasis deviation from ideal 50 µs to 15 µs de-emphasis network.

ITEM	SAMPLE FREQUENCY	RANGE	CHARACTERISTICS
Gain deviation	44.1 and 48 kHz	0 to 18 kHz	0 ±0.05 dB
		18 to 20 kHz	0.12 dB (max.)
	32 kHz	0 to 13 kHz	0 ±0.06 dB
		13 to 15 kHz	0.22 dB (max.)
Phase deviation	44.1 and 48 kHz	0 to 15 kHz	10 deg (max.)
		15 to 20 kHz	15 deg (max.)
	32 kHz	0 to 9 kHz	10 deg (max.)
		9 to 15 kHz	16 deg (max.)

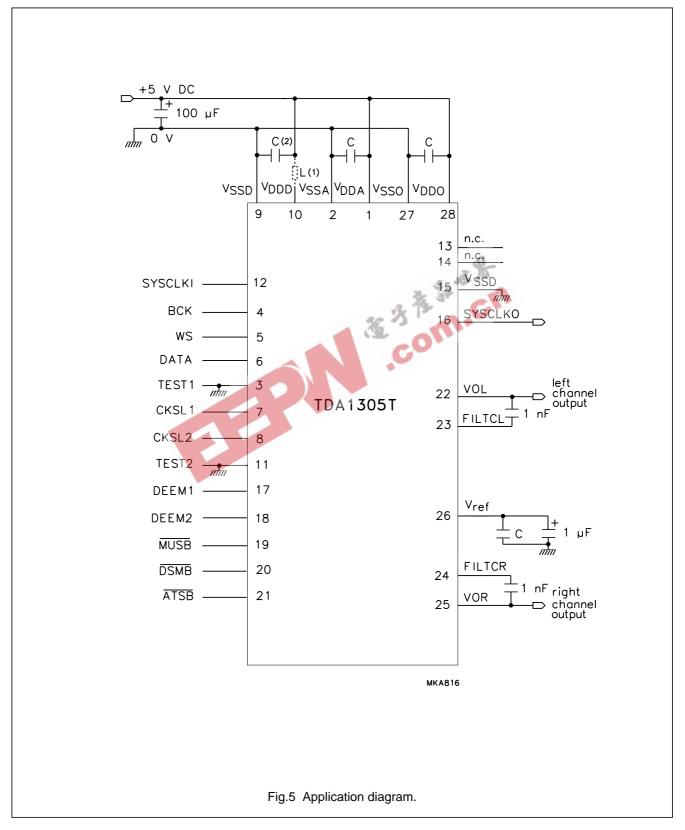
#### **Double-speed characteristics**

Table 5 Double-speed filter characteristics.				
RANGE	CONDITIONS	CHARACTERISTICS		
Pass band 0 to 17 kHz		0 ±0.075 dB		
17 to 20 kHz	3.	–0.3 dB (min.)		
24.1 to 150 kHz	typical	-47 dB (max.)		
	worst case	-45 dB (max.)		
150 kHz to infinite	typical	–33 dB (max.)		
	worst case	–25 dB (max.)		
	RANGE         0 to 17 kHz         17 to 20 kHz         24.1 to 150 kHz	RANGECONDITIONS0 to 17 kHz		



## Stereo 1fs data input up-sampling filter with bitstream continuous dual DAC (BCC-DAC2)

#### **APPLICATION INFORMATION**



#### Preliminary specification

## Stereo 1fs data input up-sampling filter with bitstream continuous dual DAC (BCC-DAC2)

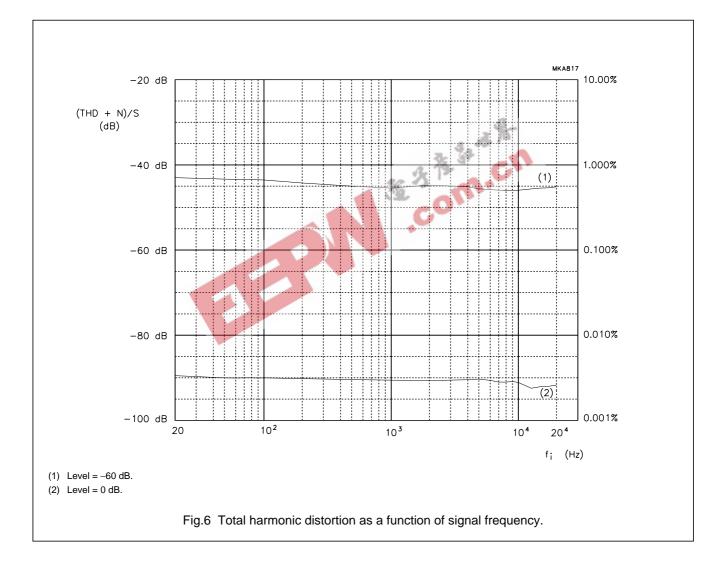
### TDA1305T

A typical application diagram is illustrated in Fig.5. The left and right channel outputs can drive a line output directly. The series inductor (L) in the digital supply line, though not strictly necessary, helps to reduce crosstalk between the digital and analog circuits.

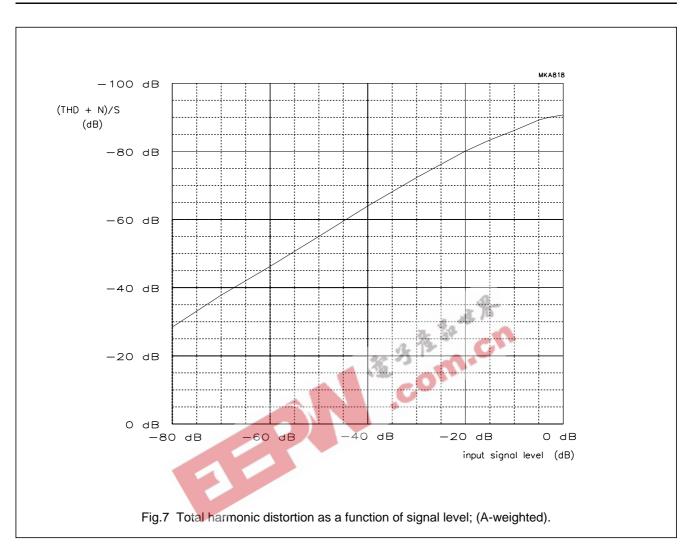
In Fig.6 measurements were taken with an 18-bit sine wave generated at a sampling rate of 48 kHz. The (THD + N)/S was measured over a bandwidth of 20 Hz to 20 kHz. The graph was constructed from average

measurement values of a small amount of engineering samples. No guarantee for typical values is implied.

In Fig.6 measurements were taken with an 18-bit sine wave generated at a sampling rate of 48 kHz. The (THD + N)/S was measured over a bandwidth of 20 Hz to 20 kHz and filtered with A-weighted characteristics. The graph was constructed from average measurement values of a small amount of engineering samples. No guarantee for typical values is implied.



## Stereo 1fs data input up-sampling filter with bitstream continuous dual DAC (BCC-DAC2)



95-01-24

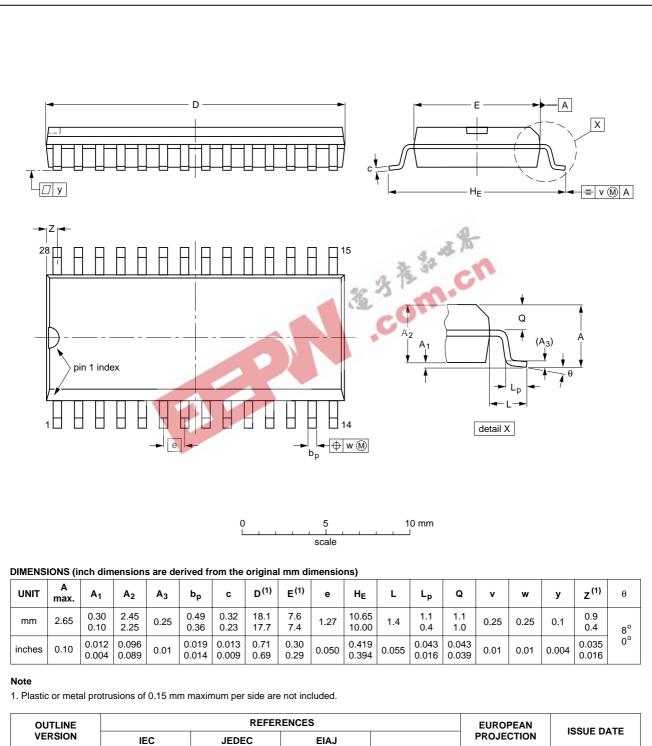
97-05-22

 $\square$ 

## Stereo 1fs data input up-sampling filter with bitstream continuous dual DAC (BCC-DAC2)

#### PACKAGE OUTLINE

#### SO28: plastic small outline package; 28 leads; body width 7.5 mm



SOT136-1

075E06

MS-013AE

### TDA1305T

#### SOLDERING

#### Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

#### **Reflow soldering**

Reflow soldering techniques are suitable for all SO packages.

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

#### Wave soldering

Wave soldering techniques can be used for all SO packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream end.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

#### **Repairing soldered joints**

Fix the component by first soldering two diagonallyopposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

### TDA1305T

#### DEFINITIONS

Data sheet status		
Objective specification	This data sheet contains target or goal specifications for product development.	
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.	
Product specification	This data sheet contains final product specifications.	
Limiting values		
more of the limiting values of the device at these or at	accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or may cause permanent damage to the device. These are stress ratings only and operation any other conditions above those given in the Characteristics sections of the specification limiting values for extended periods may affect device reliability.	
Application information		

Where application information is given, it is advisory and does not form part of the specification.

#### LIFE SUPPORT APPLICATIONS

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