

## LINEAR INTEGRATED CIRCUITS

### PREAMPLIFIER WITH ALC FOR CASSETTE RECORDERS

- EXCELLENT VERSATILITY in USE ( $V_s$  from 4 to 20V)
- HIGH OPEN LOOP GAIN
- LOW DISTORTION
- LOW NOISE
- LARGE AUTOMATIC LEVEL CONTROL RANGE
- GOOD SUPPLY RIPPLE REJECTION
- STEREO MATCHING BETTER THAN 3 dB

The TDA 1054M is a monolithic integrated circuit in a 16-lead dual in-line plastic package. The functions incorporated are:

- Low noise preamplifier
- Automatic level control system (ALC)
- High gain equalization amplifier
- Supply voltage rejection facility (SVRF).

It is intended as preamplifier in cassette tape recorders and players, dictaphones, compressor and expander in industrial equipments, Hi-Fi preamplifiers and in wire diffusion receivers; for stereo applications the ALC matching is better than 3 dB.

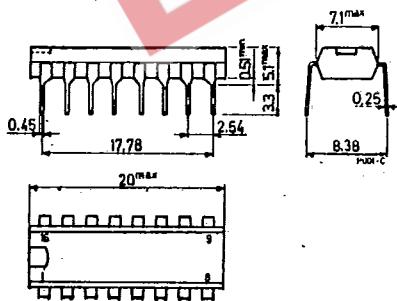
### ABSOLUTE MAXIMUM RATINGS

$V_s$	Supply voltage	20	V
$P_{tot}$	Total power dissipation at $T_{amb} \leq 50^\circ\text{C}$	500	mW
$T_{stg}, T_j$	Storage and junction temperature	-40 to 150	°C

ORDERING NUMBERS: TDA 1054M mono applications  
2 TDA 1054M stereo applications

### MECHANICAL DATA

Dimensions in mm



1034 D-07

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322

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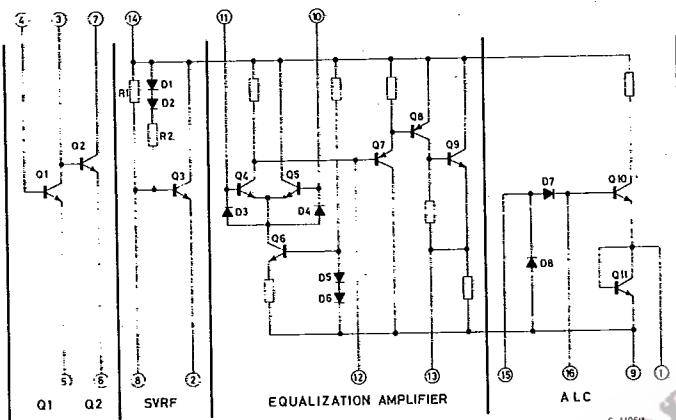
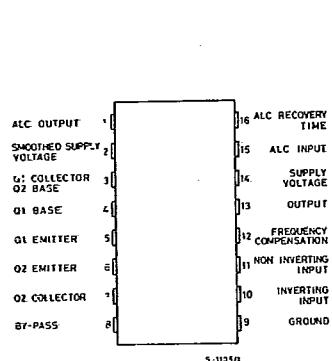
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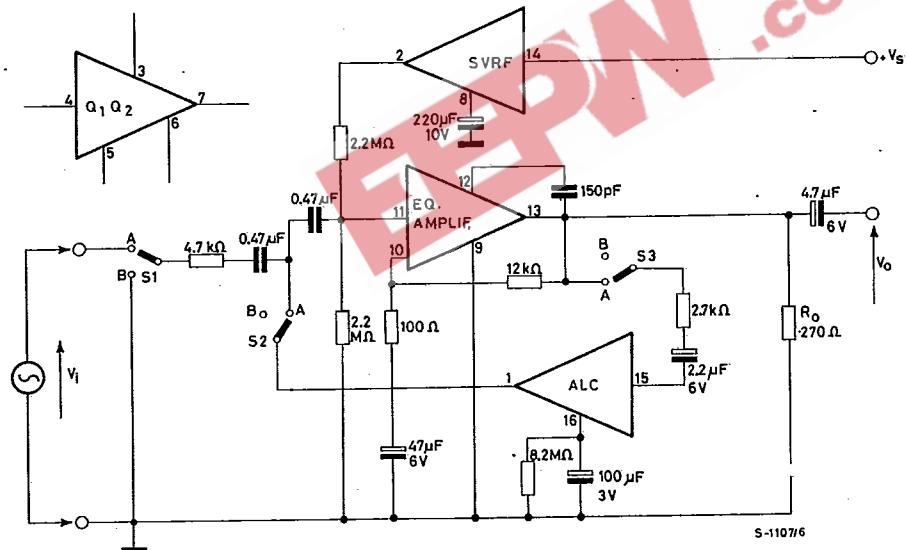
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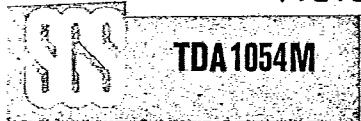
## CONNECTION AND SCHEMATIC DIAGRAMS

(top view)



## TEST CIRCUIT





## THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max 200	$^{\circ}\text{C/W}$
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## ELECTRICAL CHARACTERISTICS

(Refer to the test circuit,  $T_{amb} = 25^{\circ}\text{C}$ )

Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_s$ Supply voltage		4		20	V
$I_d$ Quiescent drain current	$V_s = 9\text{V}$ $R_L = \infty$ $S1 = S2 = S3 = B$		6		mA
$h_{FE}$ DC current gain	$I_C = 0.1\text{ mA}$ $V_{CE} = 5\text{V}$	300	600		—
$e_N$ Input noise voltage (Q1)	$I_C = 0.1\text{ mA}$ $V_{CE} = 5\text{V}$		2		$\frac{n\text{V}}{\sqrt{\text{Hz}}}$
$i_N$ Input noise current (Q1)	$f = 1\text{ kHz}$		0.5		$\frac{\text{pA}}{\sqrt{\text{Hz}}}$
NF Noise figure (Q1)	$I_C = 0.1\text{ mA}$ $V_{CE} = 5\text{V}$ $R_g = 4.7\text{ k}\Omega$ $B (-3\text{ dB}) = 20\text{ to }10,000\text{ Hz}$		0.5	4	dB
$G_V$ Open loop voltage gain (for equalization amplifier)	$V_s = 9\text{V}$ $f = 1\text{ kHz}$		60		dB
$V_o$ Output voltage with A.L.C.	$V_s = 9\text{V}$ $V_i = 100\text{mV}$ $f = 1\text{ kHz}$ $S1 = S2 = S3 = A$		1.1		V
R1 (for SVRF system)	see schematic diagram		7.5		$\text{k}\Omega$
R2 (for SVRF system)			120		$\Omega$
$e_N$ Input noise voltage (for equalization amplifier pin 11)	$V_s = 9\text{V}$ $R_g = 4.7\text{ k}\Omega$ $G_V = 40\text{ dB}$ $S1 = B$ $B (-3\text{ dB}) = 22\text{ Hz to }22\text{ KHz}$		1.3		$\mu\text{V}$
$V_{DR}$ Drop-out (between pins 14 and 2)	$V_s = 9\text{V}$ $I_d = 6\text{ mA}$		0.8		V

TDA1054M



Fig. 1 - Equivalent input spot voltage and noise current vs. bias current (input transistor  $Q_1$ )

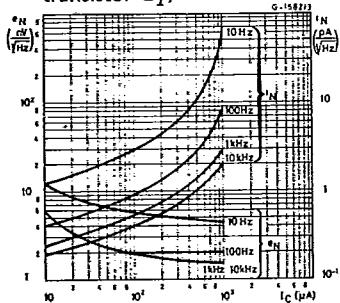


Fig. 2 - Equivalent input noise current vs. frequency (input transistor  $Q_1$ )

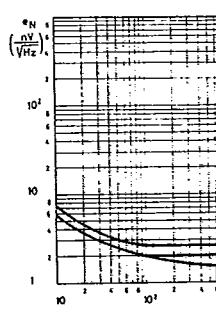


Fig. 3 - Equivalent input noise voltage vs. frequency (input transistor  $Q_1$ )

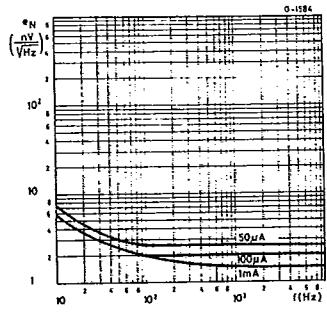


Fig. 4 - Noise figure vs. bias current (input transistor  $Q_1$ )

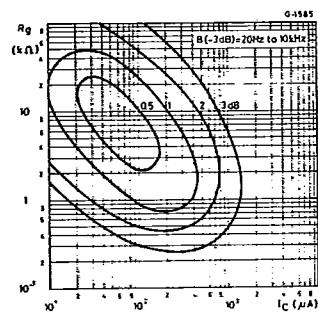


Fig. 5 - Optimum source resistance and minimum NF vs. bias current (input transistor  $Q_1$ )

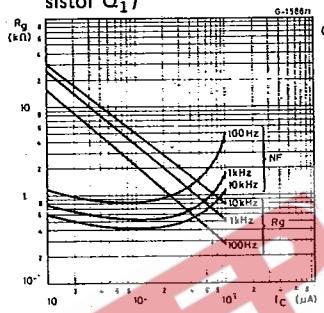


Fig. 6 - Current gain vs. collector current (input transistor  $Q_1$ )

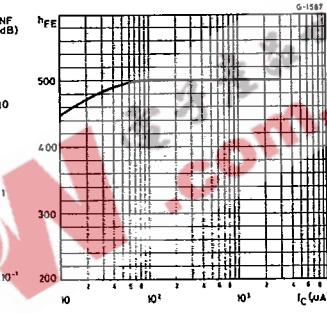


Fig. 7 - Open loop gain vs. frequency (equalization amplifier)

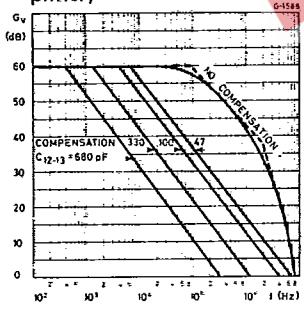
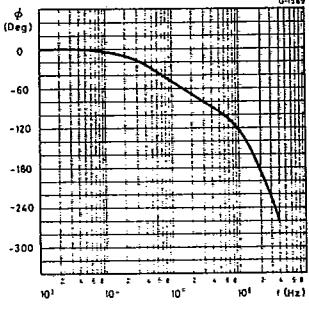
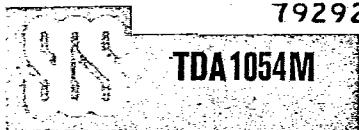


Fig. 8 - Open loop phase response vs. frequency (equalization amplifier)



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### APPLICATION INFORMATION

Fig. 9 - Application circuit for battery/mains cassette player and recorder

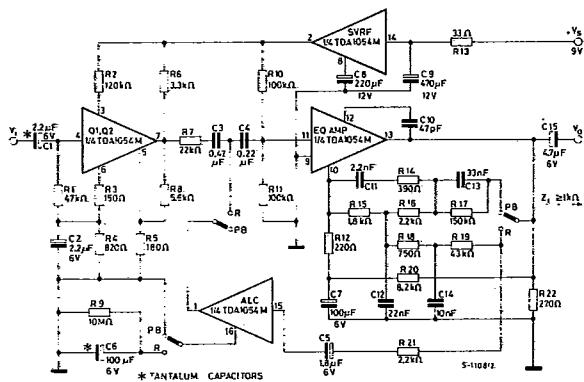
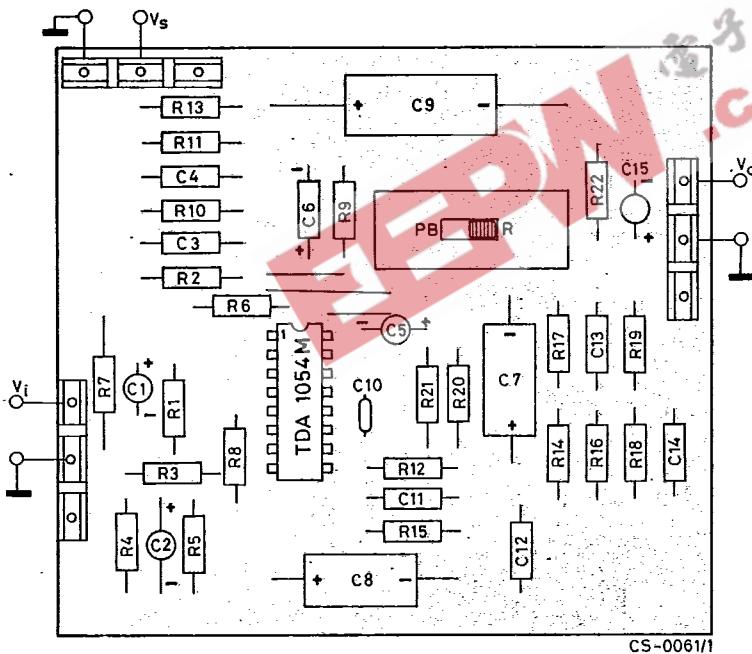
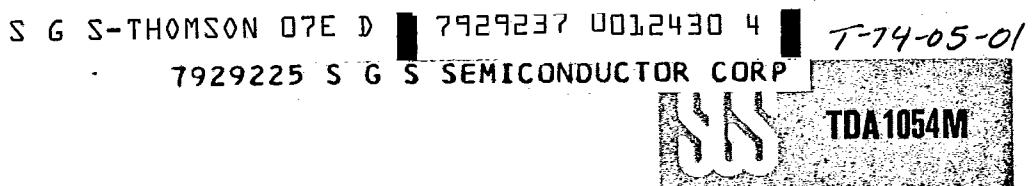


Fig. 10 - P.C. board and component layout for the circuit fig. 9 (1:1 scale)



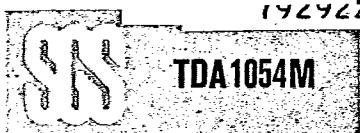
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**Typical performance of circuit in fig. 9**

( $T_{amb} = 25^\circ C$ ,  $V_s = 9V$ )

Parameter	Test conditions	Min.	Typ.	Max.	Unit
<b>PLAYBACK</b>					
$G_v$	Voltage gain (open loop) $f = 20$ to $20,000$ Hz		110		dB
$G_v$	Voltage gain (closed loop) $f = 1$ kHz		57		dB
$ Z_i $	Input impedance $f = 100$ Hz $f = 1$ kHz $f = 10$ kHz		10 41 43		k $\Omega$
$ Z_o $	Output impedance $f = 1$ kHz		12	35	$\Omega$
B	Frequency response	see fig. 12			
d	Distortion $V_o = 1V$ $f = 1$ kHz		0.1		%
Output background noise			1.3		mV
***	Output weighted background noise $Z_g = 300 \Omega + 120$ mH (DIN 45405)		1.3		mV
$S+N/N$	Signal to noise ratio $V_o = 1.3V$ $Z_g = 300 \Omega + 120$ mH		60		dB
SVR	Supply voltage ripple rejection at the output $f_{ripple} = 100$ Hz		30		dB
$t_{on}^{**}$	Switch-on time $V_o = 1V$		500		ms
<b>RECORDING</b>					
$G_v$	Voltage gain (open loop) $f = 20$ to $20,000$ Hz		110		dB
$G_v$	Voltage gain (closed loop) $f = 1$ kHz		70		dB
B	Frequency response	see fig. 14			
d*	Distortion without ALC $V_o = 1.1V$ $f = 1$ kHz		0.3		%
d	Distortion with ALC $V_o = 1.1V$ $f = 10$ kHz		0.4		%
ALC	Automatic level control range (for 3 dB of output voltage variation) $V_i \leq 40$ mV $f = 10$ kHz		54		dB
$V_o$	Output voltage before clipping without ALC $f = 1$ kHz		2.3		V
$V_o$	Output voltage with ALC $V_i = 30$ mV $f = 10$ kHz		1.1		V



Typical performance of circuit in fig. 9 (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_l^{**}$ Limiting time (see fig. 11)	$\Delta V_I = +40 \text{ dB}$ $f = 1 \text{ kHz}$		75		ms
$t_{set}^{**}$ Level setting time (see fig. 11)			300		ms
$t_{rec}^{**}$ Recovery time (see fig. 11)	$\Delta V_I = -40 \text{ dB}$ $f = 1 \text{ kHz}$		150		s
$t_{on}^{**}$ Switch-on time	$V_o = 1.1V$		500		ms
$S+N/N^{****}$ Signal to noise ratio with ALC	$V_o = 1.1V$ $R_g = 470 \Omega$		64		dB

\* Measured with selective voltmeter

\*\* This value depends on external network

\*\*\* When the DIN 45511 norm for frequency response is not mandatory the equalization peak at 10 kHz can be avoided — so halving the output noise

\*\*\*\* Weighted noise measurement (DIN 45405)

Fig. 11 - Limiting, level setting, recovery time

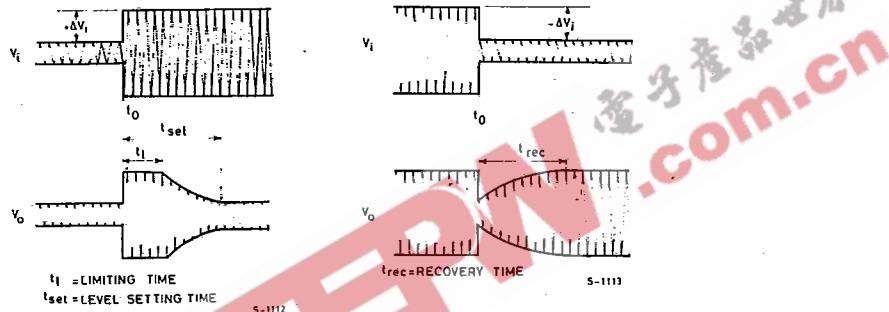


Fig. 12 - Relative frequency response for the circuit in fig. 9 (playback)

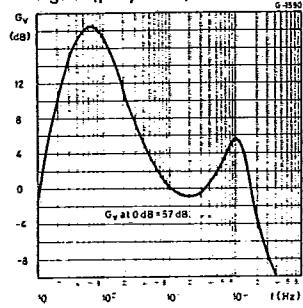


Fig. 13 - Distortion vs. frequency for the circuit in fig. 9 (playback)

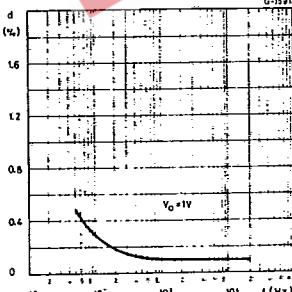
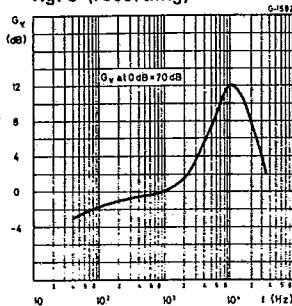


Fig. 14 - Relative frequency response for the circuit in fig. 9 (recording)



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Fig. 15 - Output voltage variation and distortion with ALC vs. input voltage for the circuit in fig. 9 (recording)

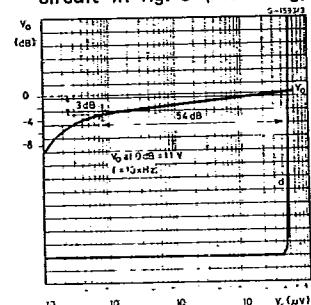


Fig. 16 - Distortion vs. frequency with ALC for the circuit in fig. 9 (recording)

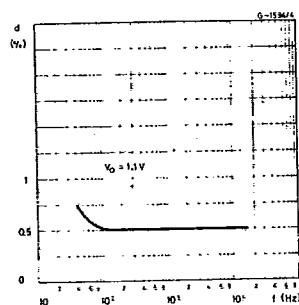


Fig. 17 - Limiting and level setting time vs. input signal variation

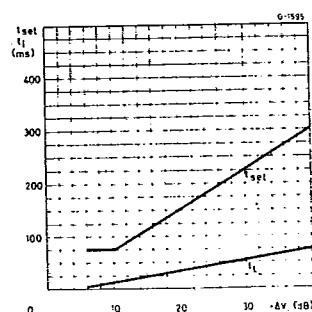
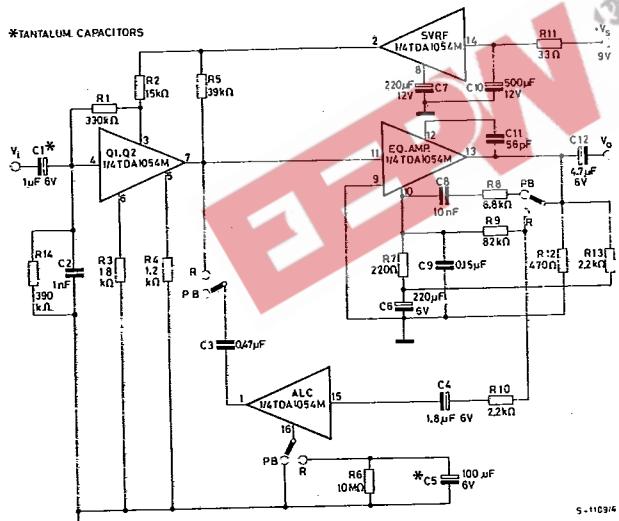


Fig. 18 - Low cost application circuit



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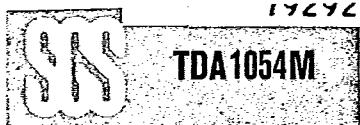
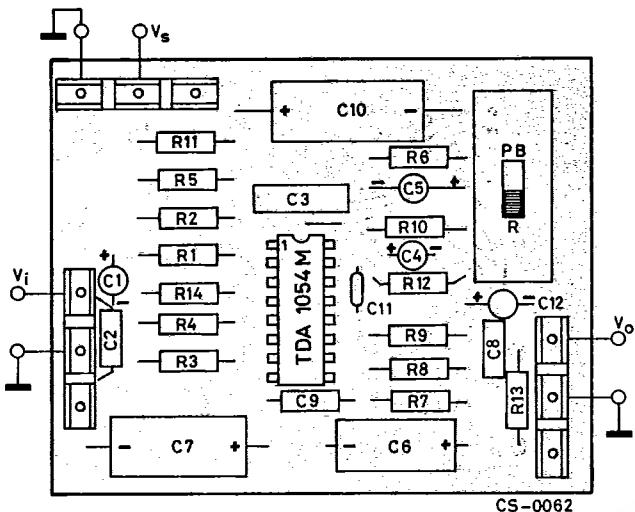


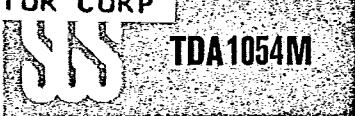
Fig. 19 - P.C. board and component layout for the circuit in fig. 18 (1:1 scale)



#### Typical performance of circuit in fig. 18

( $T_{amb} = 25^\circ\text{C}$ ,  $V_s = 9\text{V}$ )

Parameter	Test conditions	Min.	Typ.	Max.	Unit
<b>PLAYBACK</b>					
$V_s$	Supply voltage		5		12 V
$I_d$	Quiescent drain current		18		mA
$G_v$	Voltage gain (closed loop)	$f = 1\text{ kHz}$	54		dB
B	Frequency response	$f = 100\text{ Hz}$ $f = 1\text{ kHz}$ $f = 6\text{ kHz}$ $f = 10\text{ kHz}$ $f = 60\text{ kHz}$	12 0 5 11 10		dB
d	Distortion	$V_o = 1\text{V}$ $f = 1\text{ kHz}$	0.6		%
$e_N$	Output weighted background noise	$Z_g = 300\ \Omega + 120\text{ mH}$ (DIN 45405)	1.3		mV



Typical performance of circuit in fig. 18 (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
<b>RECORDING</b>					
G <sub>V</sub>	Voltage gain (closed loop)	f = 1 kHz		70	dB
B	Frequency response	f = 140 Hz f = 1 kHz f = 10 kHz		-3 0 4	dB dB dB
d	Distortion	V <sub>O</sub> = 1.1V f = 10 kHz		0.7	%
ALC	Range for 3 dB of output voltage variation	V <sub>I</sub> ≤ 40 mV f = 10 kHz		54	dB

Fig. 20 - Complete cassette player and recorder

